

4th European HICUM Workshop

HICUM Model and Scaling Laws Issues

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HICUM Model and Scaling Laws Issues

→ HICUM Model Issues

- Temperature dependence of the base current.
- Temperature dependence of the neutral emitter storage time.
- Temperature dependence of the zero bias hole charge.
- Temperature dependence of the low-current forward transit time.
- Bump on $I_C(V_{CE})$ characteristics in saturation region.
- Effect of the reverse transit time T_R .
- HICUM implementation issues in circuit simulators.
- Status of HICUM Level 0.

→ HICUM Scaling Laws (or Parameter Extraction) Issues

- Split of the BE junction capacitance.
- Split of the BC junction capacitance.
- Weak avalanche current.
- Forward transit time for small devices.

HICUM Model Issues

→ Temperature Dependence of the Base Current.

- The HICUM temperature dependence of the base saturation current I_{BS} is derived from the temperature dependence of the saturation current I_S and of the forward current gain B_F as follows

$$I_{BS}(T) = \frac{I_S(T)}{B_F(T)}$$

In HICUM (version 2.1) the temperature dependence of I_S is similar to the one of the SGP model with $X_{TI} = 3$.

On the other hand, the temperature dependence of B_F is linear with the temperature

$$B_F(T) = B_F(T_0) \cdot (1 + A_{LB} \cdot (T - T_0))$$

This formulation is **too simple** and **not physical**, as shown in the next section.

Finally, the temperature dependence of the base saturation current can be written (non ideality factor assumed to be equal to 1)

$$I_{BS}(T) = I_{BS}(T_0) \cdot \left(\frac{T}{T_0}\right)^{A_{CT}} \cdot \exp\left\{\left(\frac{V_{GB}}{V_{T0}} - T \cdot A_{LB}\right) \cdot \left(1 - \frac{T_0}{T}\right)\right\} \quad \text{with } A_{CT} = 3 \text{ in HICUM version 2.1.}$$

- More physics based formulation of the forward current gain temperature dependence

As the forward current gain is proportional to $e^{-\frac{\Delta_{VG}}{V_T}}$, where Δ_{VG} is the difference between the bangap voltage in the base and in the emitter, $\Delta_{VG} = V_{GB} - V_{GE}$, its temperature dependence can be written

$$B_F(T) = B_F(T_0) \cdot \exp\left\{\frac{\Delta_{VG}}{V_{T0}} \cdot \left(1 - \frac{T_0}{T}\right)\right\} \text{ or } B_F(T) = B_F(T_0) \cdot \exp\left\{A_{LB} \cdot \left(1 - \frac{T_0}{T}\right)\right\} \text{ with } A_{LB} = \frac{\Delta_{VG}}{V_{T0}}$$

Therefore, the temperature dependence of the base saturation current can be written (non ideality factor assumed to be equal to 1)

$$I_{BS}(T) = I_{BS}(T_0) \cdot \left(\frac{T}{T_0}\right)^{A_{CT}} \cdot \exp\left\{\left(\frac{V_{GB}}{V_{T0}} - A_{LB}\right) \cdot \left(1 - \frac{T_0}{T}\right)\right\}$$

We can notice that this formulation is similar to the HICUM model, with A_{LB} instead of $T \cdot A_{LB}$.

✓ Suggested model

- In order to have a more physics based temperature dependence of the base current, and to take into account both the temperature dependence of the base and emitter Gummel numbers

$$B_F(T) = \frac{N_{GE}(T)}{N_{GB}(T)}$$

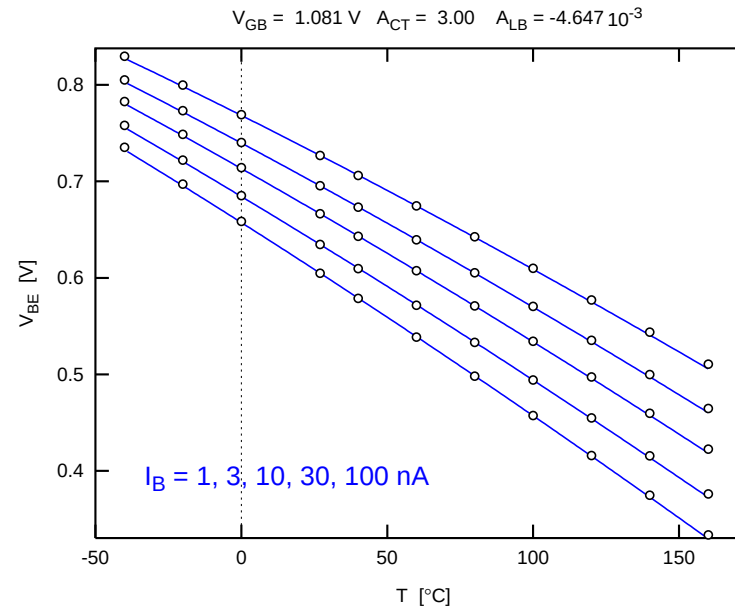
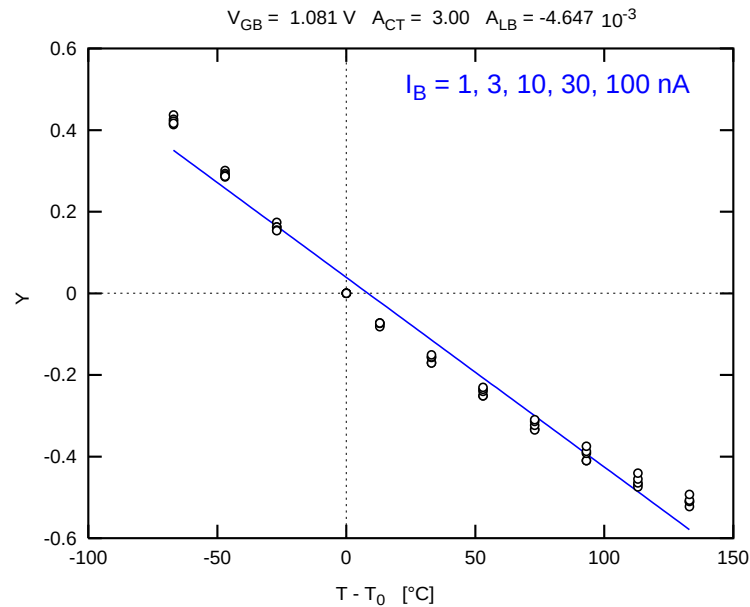
we suggest to use the following model which is in fact already implemented in other BJT models like VBIC or MEXTRAM

$$B_F(T) = B_F(T_0) \cdot \left(\frac{T}{T_0}\right)^{A_{BT}} \cdot \exp\left\{A_{LB} \cdot \left(1 - \frac{T_0}{T}\right)\right\}$$

- The temperature dependence of the base saturation current can, in this case, be written (non ideality factor assumed to be equal to 1)

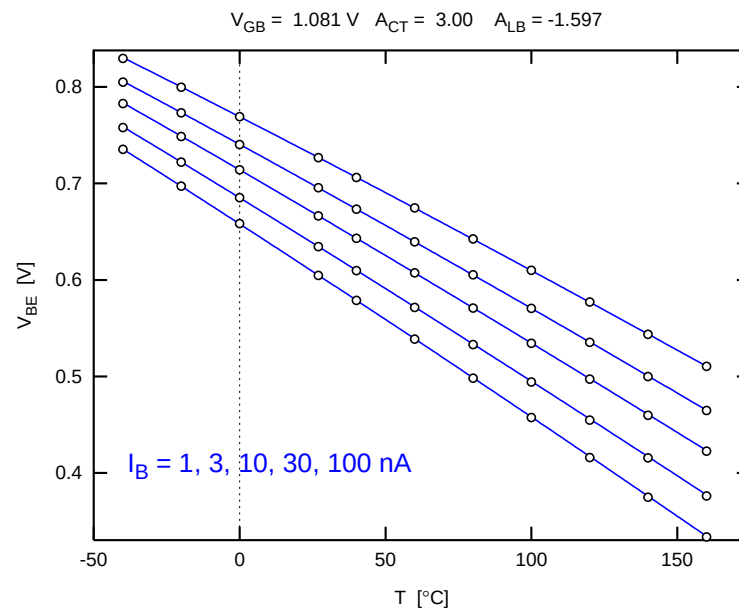
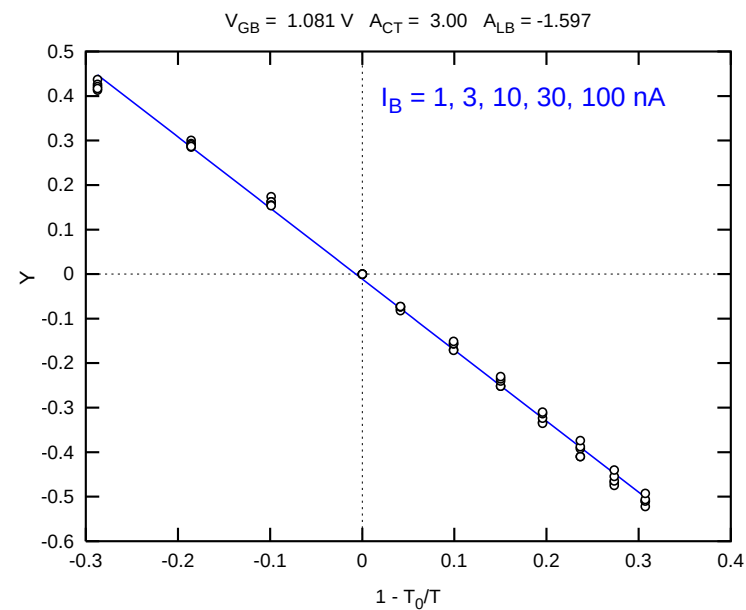
$$I_{BS}(T) = I_{BS}(T_0) \cdot \left(\frac{T}{T_0}\right)^{A_{CT} - A_{BT}} \cdot \exp\left\{\left(\frac{V_{GB}}{V_{T0}} - A_{LB}\right) \cdot \left(1 - \frac{T_0}{T}\right)\right\}$$

Validation on experimental results



HICUM Version 2.1

Deviation greater than 5mV
at extreme temperatures



Proposed Model with $A_{BT} = 0$

Deviation lower than 1.5mV
in all the temperature range

→ Temperature Dependence of the Neutral Emitter Storage Time.

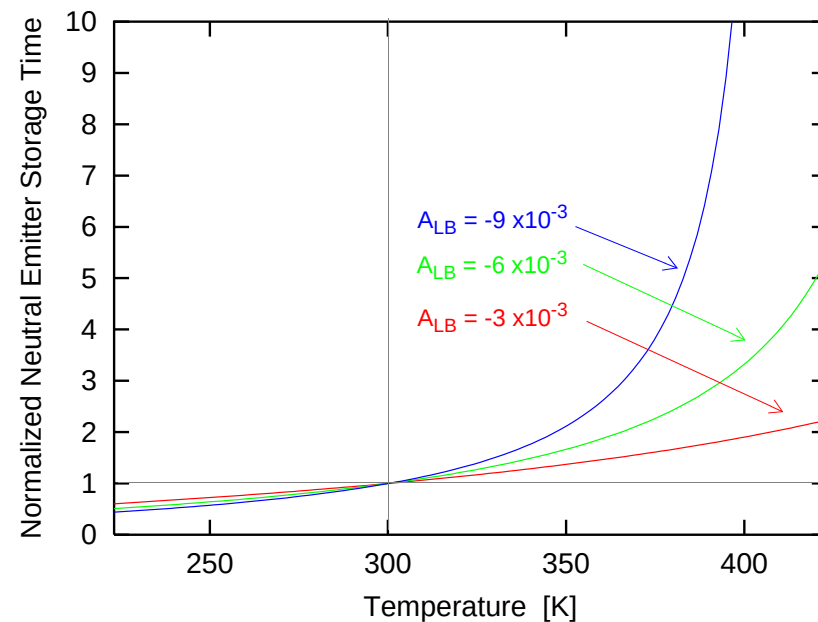
- The temperature dependence of the neutral emitter storage time T_{EF0} is given by

$$T_{EF0}(T) = T_{EF0}(T_0) \cdot \frac{T/T_0}{1 + (A_{LB} \cdot (T - T_0))} \quad \text{There is a numerical issue when } A_{LB} \cdot (T - T_0) = -1.$$

To overcome this issue, in HICUM the following expression is used

$$T_{EF0}(T) = T_{EF0}(T_0) \cdot \frac{T/T_0}{1 + A_{LB} \cdot (T - T_0) + \frac{\sqrt{(1 + A_{LB} \cdot (T - T_0))^2 + 10^{-6}}}{2}}$$

The smoothing function in the denominator avoids numerical problem at extreme temperatures (very low or high, depending on the sign of A_{LB} and on its value)

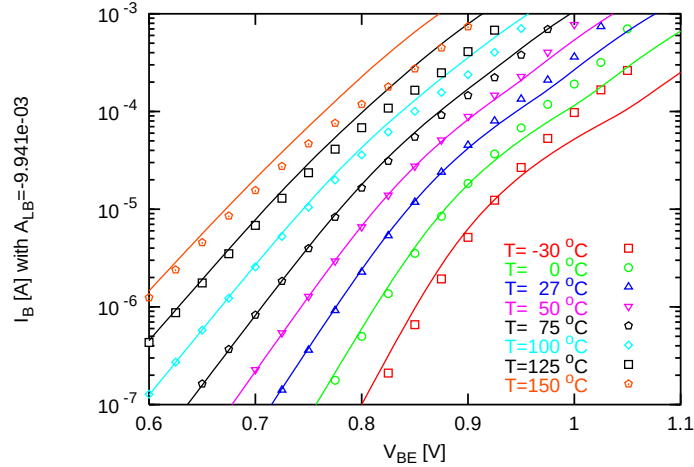


But, when the sign of $1 + A_{LB} \cdot (T - T_0)$ changes, the decrease of the smoothing function is too abrupt and T_{EF0} increases too rapidly. Consequently f_T decreases too much in comparison with measurements.

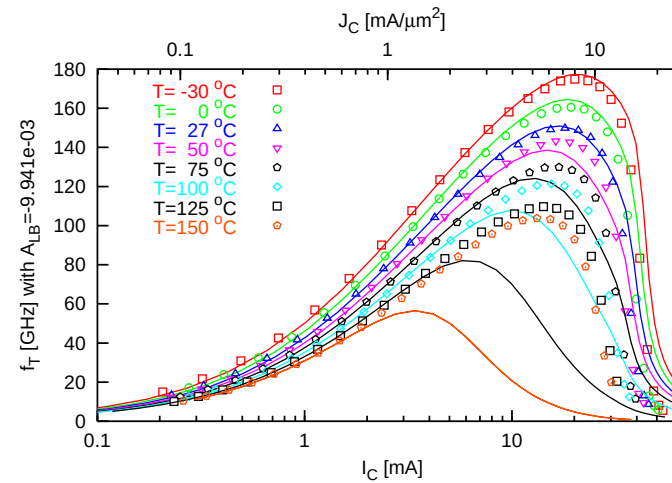
To solve this important issue, it is necessary to modify this smoothing function or to find a more physical law for the temperature dependence of T_{EF0} and perhaps, decouple the temperature dependence of T_{EF0} from the temperature dependence of the current gain.

- Today, the only way to correct this issue is to alter the value of A_{LB} . In this case the temperature dependence of f_T is correct but not the one of the base current.

I_B at $V_{BC}=0$ for a $0.3\mu\text{m} \times 14.92\mu\text{m}$ CBEBC device in a $[-30^\circ\text{C}; 150^\circ\text{C}]$ temperature range

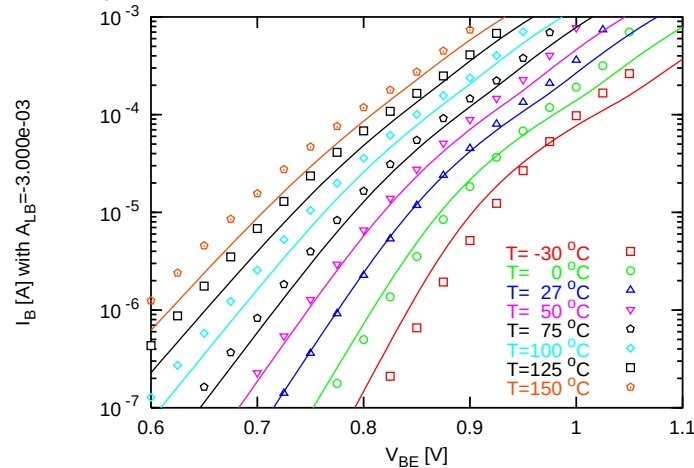


$f_T(I_C)$ for a $0.3\mu\text{m} \times 14.92\mu\text{m}$ CBEBC device at $V_{BC}=0\text{V}$

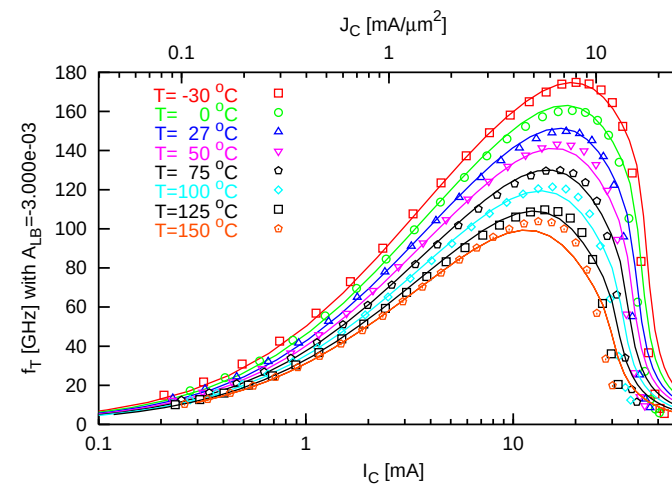


$$A_{LB} = -9.94 \times 10^{-3}$$

I_B at $V_{BC}=0$ for a $0.3\mu\text{m} \times 14.92\mu\text{m}$ CBEBC device in a $[-30^\circ\text{C}; 150^\circ\text{C}]$ temperature range



$f_T(I_C)$ for a $0.3\mu\text{m} \times 14.92\mu\text{m}$ CBEBC device at $V_{BC}=0\text{V}$



$$A_{LB} = -3.00 \times 10^{-3}$$

→ Temperature Dependence of the Zero Bias Hole Charge.

- In the formulation of the temperature dependence of Q_{P0} given by

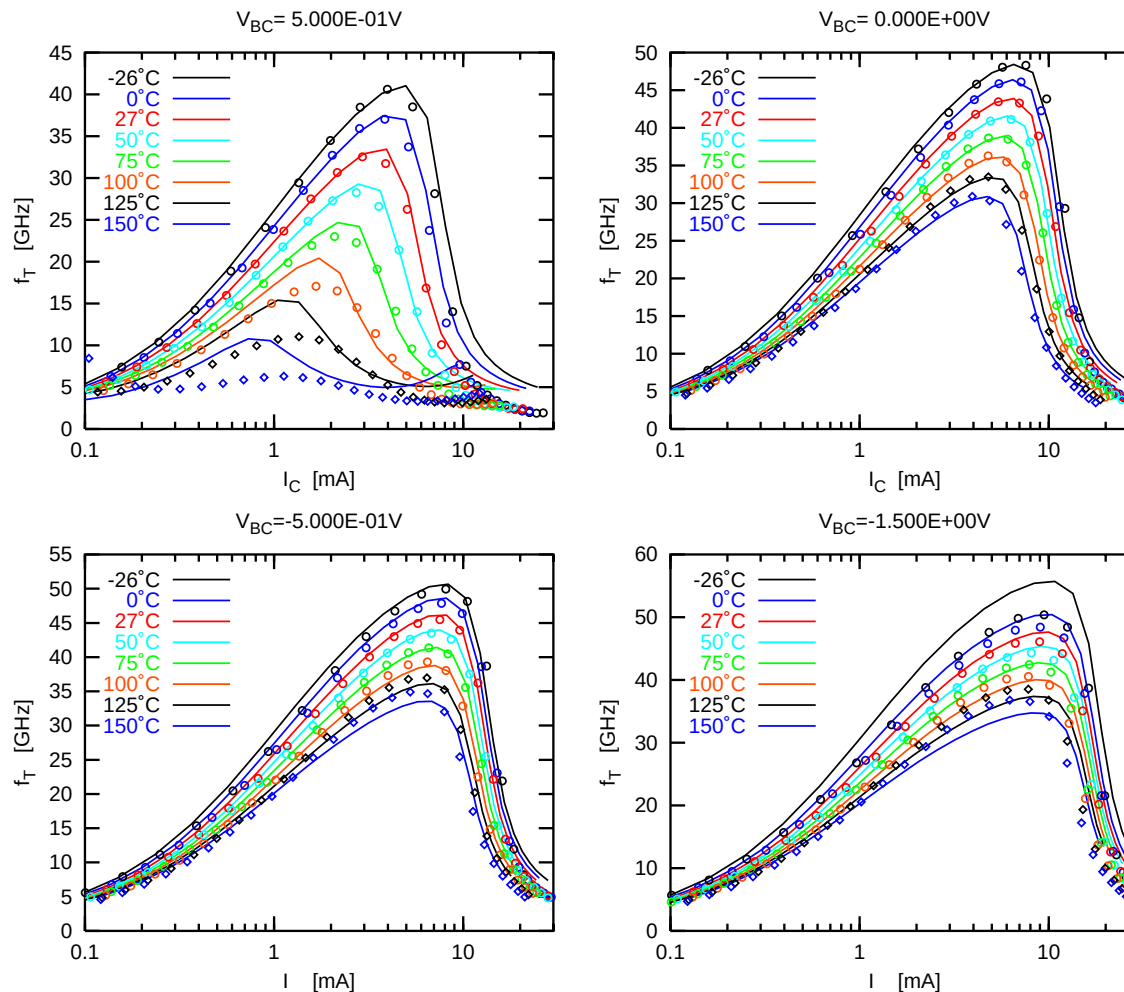
$$Q_{P0}(T) = Q_{P0}(T_0) \cdot \left[1 + \frac{Z_{EI}}{2} \cdot \left\{ 1 - \frac{V_{DEI}(T)}{V_{DEI}(T_0)} \right\} \right]$$

the factor **2** is arbitrary.

A model parameter could be preferable, otherwise there is **no way** to alter the temperature dependence of Q_{P0} .

→ Temperature Dependence of the Low-Current Forward Transit Time.

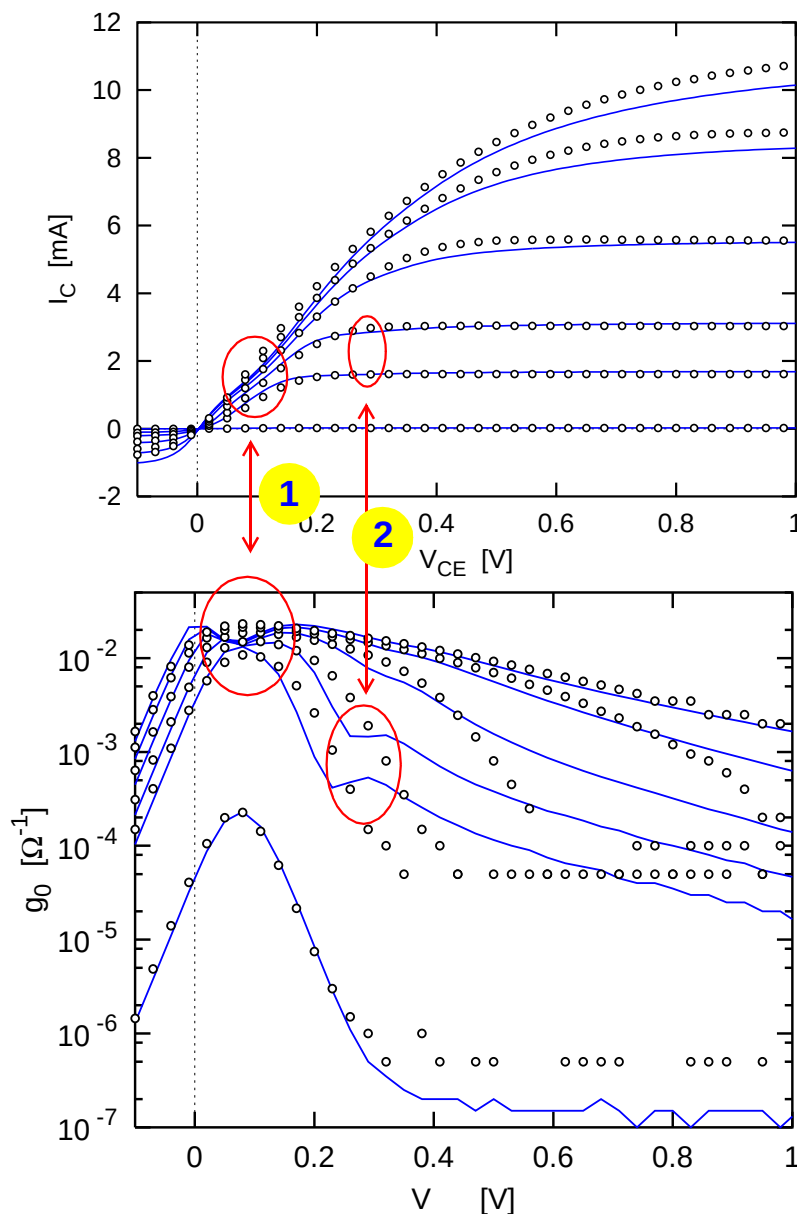
- In HICUM, only the low-current forward transit time at $V_{BC}=0V$, T_0 , is temperature dependent. Experimentally, we notice a variation of D_{T0H} and T_{BVL} with the temperature. Is it an extraction issue or a limitation of HICUM?. To be investigated.



Self-heating included

← Model limitation or extraction issue?

→ Bump on $I_C(V_{CE})$ Characteristics in Saturation Region.



1

'Bump' in hard-saturation region, at medium and high current densities.

- Cannot be corrected by increasing V_{DCI} or/and V_{DCX} .
- The cause of this 'bump' is not clearly identified. Is it due to a wrong combination of model parameters, or is it a limitation of the HICUM model for linking the quasi-saturation and saturation region?
- To be analysed, which parameters are responsible of this 'bump'.
- Could be affect the convergence of the model?.

2

'Discontinuity' in quasi-saturation region, at medium current densities.

- Can be corrected by increasing V_{DCI} or/and V_{DCX} .
- Nevertheless, an improvement of the formulation of the BC junction capacitance in forward mode would be better.
- Is it planned to modify the model equations?.

→ Effect of the Reverse Transit Time T_R

- Strange behaviour of the output characteristics in the saturation region when T_R parameter is specified:
 - $T_R = 0$ no effect on the output characteristics (it is natural)
 - If T_R is different of zero, **important** and **same** effects on the output characteristics whatever the value of T_R (1 fs, 1ps, 1 ns).

• Is it normal?

• Can we really use T_R in HICUM?

- Some EDA vendors have modified the expression of the BC diffusion capacitance as follows

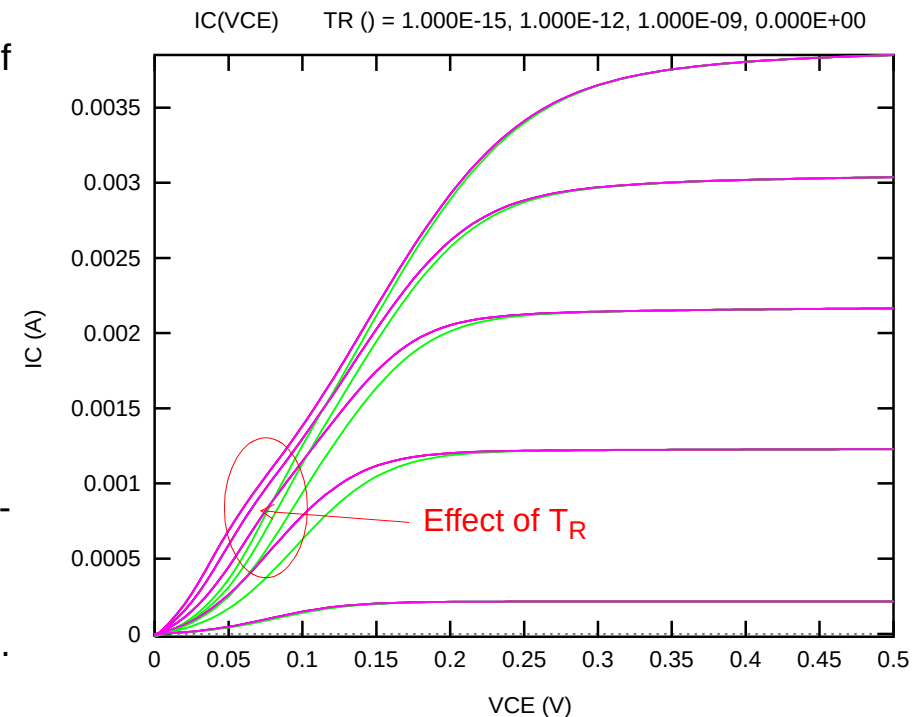
$$C_{DBC} = T_R \cdot \left. \frac{dI_{TR}}{dV_{CE}} \right|_{V_{BE}} \quad \text{instead of}$$

$$C_{DBC} = T_R \cdot \left. \frac{dI_{TR}}{dV_{CE}} \right|_{V_{BE}} + T_F \cdot \left. \frac{dI_{TF}}{dV_{CE}} \right|_{V_{BE}} \quad (\text{HICUM 2.1})$$

• Is it justified?

• Is this modification have been done by **all** EDA vendors?

• HICUM code and documentation **have to be updated**.



→ HICUM Implementation Issues in Circuits Simulators

- For the first time (Q1 2004) we have delivered to design kit model libraries based on HICUM. The main supported circuit simulators being: ADS, APLAC, ELDO, HSPICE, SPECTRE...
- The feedback of internal and external customers were not good:
 - Too CPU time increase in comparison with STBJT model (SPICE Gummel-Poon model like including base push-out effect and BC breakdown...).
 - Important convergence issues in **transient analysis** and **large signal simulations**, especially with APLAC, ELDO and SPECTRE.
- Some CAD vendors provided *patches* solving these issues:
 - Correction of the *incomplete* or *wrong* derivatives (temperature, NQS effects...)
 - Modification of the HICUM equations (BC diffusion capacitance: T_F contribution removed, ...)
- Convergence and CPU time were improved (but not totally solved, simple CML ring oscillator still does not converge in some simulators), and now we ask the following questions:
 - Where is the **reference code** of HICUM?
 - Are we certain that all convergence issues are completely solved?
 - Are we certain that all EDA vendors have made the **same corrections** (model equations, derivatives,...) and implemented the **same code** without new bugs?
- In order to answer all these questions, and to avoid these kind of issues in future (bad reputation for HICUM: very accurate model but unusable), we request to M. Schroter **to become again the owner of the code**, that is to say, we request a new HICUM version (version 2.1 dated December 2000 is now obsolete) that solve all these existing issues (code formulation and derivatives...).
- This work could be partially founded by the CMC and by companies supporting HICUM.

→ Status of HICUM Level 0.

- Maturity of the model, accuracy in comparison with HICUM Level 2.
- Status of the code: CPU time, convergence in comparison with HICUM Level 2.
- Implementation in circuit simulators: which ones, when?.
- ST request: **to add the parasitic PNP substrate** inside the model, otherwise it is not possible to accurately fit the output characteristics in the saturation region.
A solution could be the use of a sub-circuit, but lower CPU time (more node to be solve) and possible convergence issue (all parameters of the parasitic PNP not defined).

HICUM Scaling Laws Issues

→ Split of the BE junction capacitance

- Depending on the process and on the architecture of the device, the scaling laws used in TRADICA could give negative peripheral BE junction capacitance, it is the case when the γ (ratio of the peripheral specific junction capacitance to the area specific junction capacitance) **associated to the junction capacitance** is lower than the γ (ratio of the peripheral collector current density to the area collector current density) **associated to the collector current**. **How to solve this issues?.**

→ Split of the BC junction capacitance

- How to split accurately the BC junction capacitance between the internal and external part?
The fit obtained on the Y_{12} parameter demonstrates that this partitioning is not always enough accurate **(the internal part of the BC junction capacitance is often underestimated)**.

→ BC weak avalanche parameters

- In HICUM the weak avalanche current is given by

$$I_{AVL} = I_{TF} \cdot F_{AVL} \cdot (V_{DCI} - V_{BCi}) \cdot \exp \left\{ -\frac{Q_{AVL}}{C_{JCI} \cdot (V_{DCI} - V_{BCi})} \right\}$$

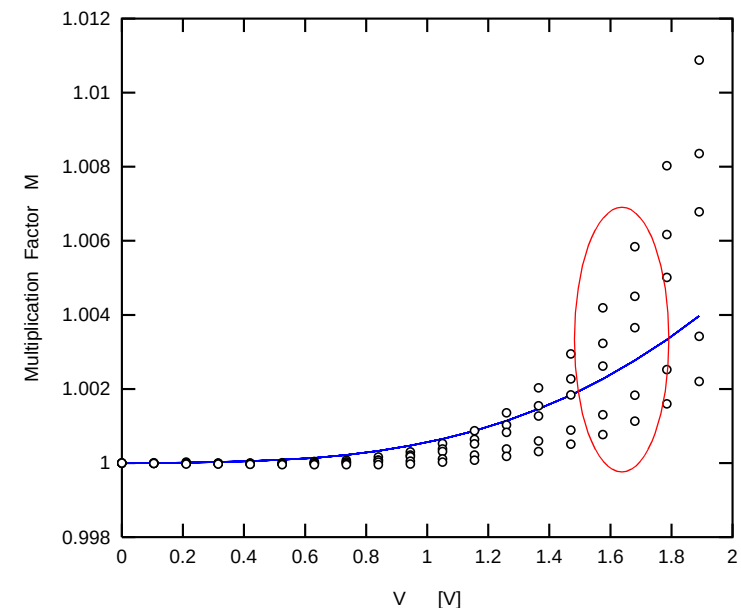
If it is assumed that C_{JCI} , Q_{AVL} and I_{TF} are only proportional to the effective emitter area, I_{AVL} is also proportional only to the effective emitter area.

- From $I_C = I_{C0} + I_{AVL} = M \cdot I_{C0}$ we can deduce that the multiplication factor M is geometry independent

$$M = 1 + \frac{I_{AVL}}{I_{C0}} = \underbrace{F_{AVL} \cdot (V_{DCI} - V_{BCi}) \cdot \exp \left\{ -\frac{Q_{AVL}}{C_{JCI} \cdot (V_{DCI} - V_{BCi})} \right\}}_{\text{geometry independent}}$$

- But experimentally, it is not the case. 

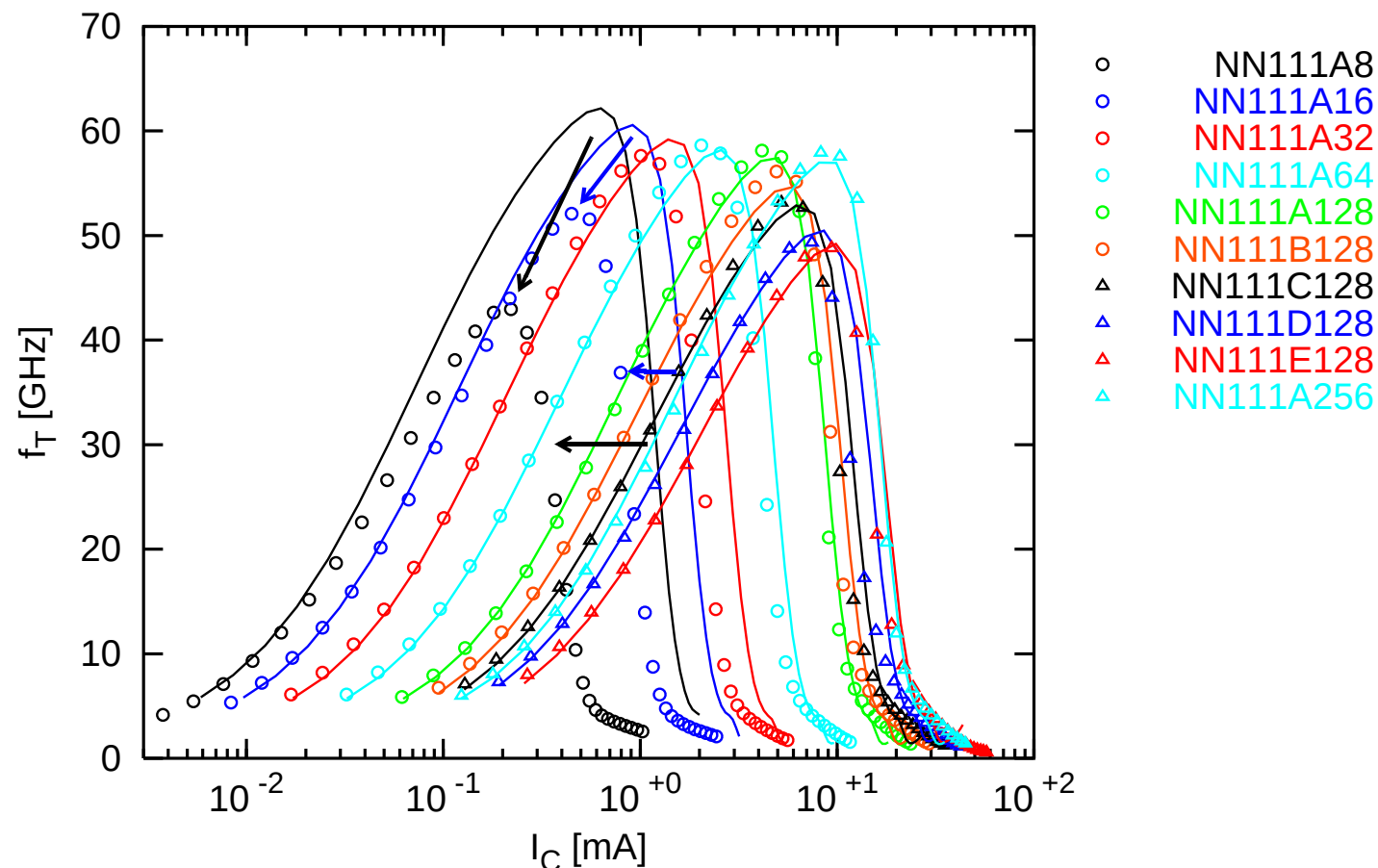
- How to solve this scaling issue of the avalanche current (TRADICA)?



Multiplication factor for transistors with different emitter length and width

→ Transit time scaling laws

- Depending on the process, it is difficult to accurately fit the f_T characteristics for devices with small emitter length. Is it an extraction issue (I_{CK} overestimated for small devices) or the scaling law for R_{C10} has to be improved (one δ_C for each direction)?.
- Is it a problem of non-scalable process?.
- Is there a solution in TRADICA to solve this issue?. How?.



Summary of Main ST Requirements

→ New HICUM Level 2 version (2.2)

- Update of the model equations and of the code corresponding to the corrections done by EDA vendors for solving all convergence issues.
- Reference code needed.

→ New HICUM Level 2 version (2.3)

- Correction of the temperature scaling laws (especially for I_B and T_{EF0}).
- Other physical effects to be improved?...

→ Level 0 status and implementation in circuits simulators

→ Improvement of HICUM Scaling laws and guidelines for the extraction of some specific parameters