Two-/three-dimensional GICCR Master equation for Si/SiGe bipolar transistors

M. Schroter^{1),2)} and H. Tran²⁾

¹⁾ECE Dept., University of California San Diego, La Jolla, CA, USA ²⁾Chair for Electron Devices and Integr. Circuits, Univ. of Technol. Dresden, Germany mschroter@ieee.org

http://www.iee.et.tu-dresden.de/iee/eb/eb_homee.html

HICUM Workshop, Grenoble (France), 6./7. June 2005

Contents

- Introduction
- Investigated device structure and process technology
- Theory and derivation
- Model equations
- Conclusions

Multi-dimensional GICCR Introduction

Introduction

Bipolar transistor applications

- high-frequency/high-speed operation
 - Bluetooth, 802.11, WLAN (e.g. 60GHz), UWB (impulse) radio, Free Space Optics
 - OC 192/768...
 - · linear circuits, ...

Constraints

- some applications are close to the technology limit ⇒ careful circuit optimization
- cost reduction (mask, re-spins..., yield prediction)
 - ⇒ need for accurate compact models for SiGe HBTs
- presently: covered by advanced models such as HICUM, MEXTRAM, VBIC
- future: increasing importance of accurate models (e.g., geometry scaling ...)
 - ⇒ **goal**: physics-based model concept for future technologies

Multi-dimensional GICCR Introduction

Transfer current theory

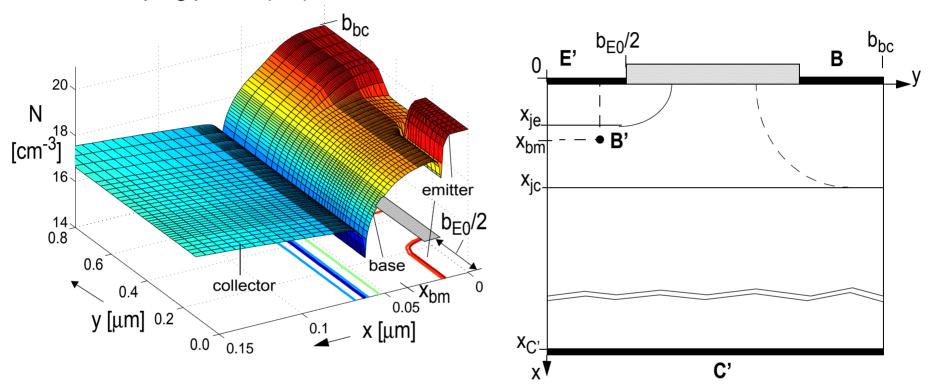
- classic approach: differential equation in base region
 - 1D
 - different formulation for each bias and spatial region (no single-piece expression)
 - important effects (such as Early- and high-current effect) are difficult to include
- Integral Charge-Control Relation (ICCR):
 - natural to BJT action, elegant single-piece solution over relevant bias region
 - requires smooth charge model
 - limited to transistors without bandgap variation, 1D structure
- Generalized ICCR (GICCR)
 - extension of ICCR
 - applicable to HBTs, but still 1D

⇒ This work: multi-dimensional GICCR formulation

Investigated device structure and process technology

vertical doping profile (1D) under emitter

schematic cross-section of SiGe HBT



- profile corresp. to high-speed transistor version with peak $f_T \approx 100 GHz$ @ V_{BC} = 0V
- example here: "conventional" profile (other profiles have been investigated, too)
- 2D device simulation: unit emitter length $I_F = 1 \mu m$ (no new information by 3D simul.)

Theory and derivation

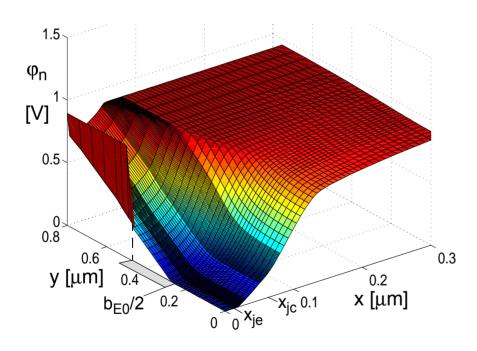
- assumptions:
 - · negligible volume recombination
 - zero time derivative (for practical applications: quasi-static operation)
 - tunneling and thermionic emission across the junctions are neglected
- starting point: x-component of electron current density, $J_{nx} = -q\mu_n n \frac{d\varphi_n}{dx}$
- yields (same as 1D derivation): $I_T h(x, y) p(x, y) = -c_0 \frac{d[\exp(-\varphi_n/V_T)]}{dx} \exp\left(\frac{V_{B'E'}}{V_T}\right)$
- with constant $c_0 = qV_T A_{E0} \mu_{n0} n_{i0}^2$ and weighting function

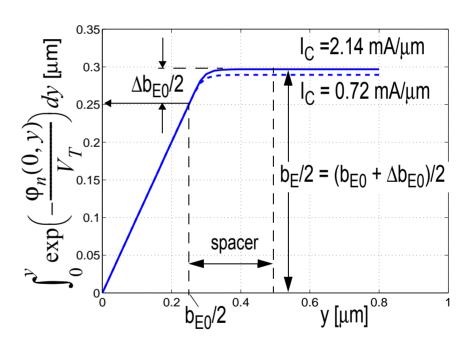
$$h(x,y) = \underbrace{\frac{-J_{nx}(x,y)}{I_{T}/A_{E0}}}_{\text{h}_{J}(x,y)} \underbrace{\frac{\mu_{n0}n_{i0}^{2}}{\mu_{n}n_{i}^{2}}}_{\text{h}_{g}(x,y)} \exp\left(\frac{V_{B'E'} - \phi_{p}(x,y)}{V_{T}}\right)$$

Theory (cont'd)

• 2D integration of r.h.s.:

$$\int_{0}^{b_{bc}} \int_{\exp\left(-\frac{\Phi_{n}(x_{C'}, y)}{V_{T}}\right)}^{\exp\left(-\frac{\Phi_{n}(x_{C'}, y)}{V_{T}}\right)} d\left[\exp\left(-\frac{\Phi_{n}}{V_{T}}\right)\right] dy = \int_{0}^{b_{bc}} \exp\left(-\frac{\Phi_{n}(x_{C'}, y)}{V_{T}}\right) dy - \int_{0}^{b_{bc}} \exp\left(-\frac{\Phi_{n}(0, y)}{V_{T}}\right) dy$$





$$\int_{0}^{b_{bc}} [...] dy = \exp\left(-\frac{V_{CE}}{V_{T}}\right) b_{bc} - \left(\frac{b_{E0}}{2} + \frac{\Delta b_{E0}}{2}\right)$$

$$\int_{0}^{b_{bc}} [\dots] dy = \exp\left(-\frac{V_{CE}}{V_{T}}\right) b_{bc} - \left(\frac{b_{E0}}{2} + \frac{\Delta b_{E0}}{2}\right) \quad \text{with} \quad \Delta b_{E0} = 2 \int_{b_{E0}/2}^{b_{bc}} \exp\left(-\frac{\varphi_{n}(0, y)}{V_{T}}\right) dy$$

Theory (cont'd)

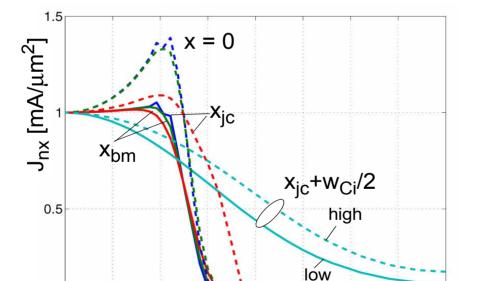
• 2D integration of l.h.s.: $I_T \int_0^{b_{bc}} \int_0^{x_{C'}} h_g(x, y) h_J(x, y) h_e(x, y) p(x, y) dx dy$

0.7

0.8

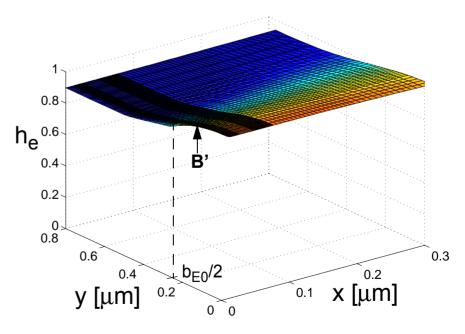
• $h_g = (\mu_{n0} n_{i0}^2) / (\mu_n n_i^2)$: impact in y-direction negligible \Rightarrow treat as in 1D case

$$h_J = \frac{-J_{nx}(x, y)}{I_T / A_{E0}}$$



^{0.3} y [μm] ^{0.5}

$$h_e = \exp\left(\frac{V_{B'E'} - \varphi_p(x, y)}{V_T}\right)$$



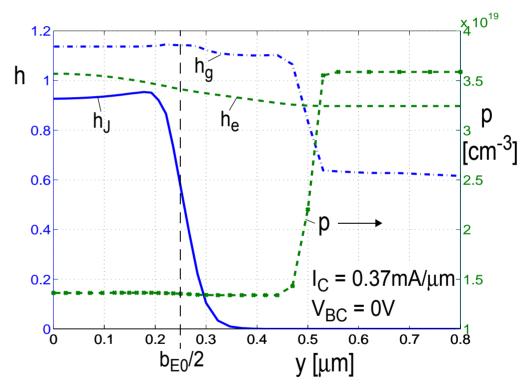
© MS

0.1

Theory (cont'd)

- relation of weighting functions to hole density distribution
 - lateral spatial dependence of h_J , h_e , h_g , and p along $x = x_{bm}$

- hg
 - constant vs y \Rightarrow h_g \approx h_g(x,0)
- h_J
 - constant vs y under most of window (can be affected by emitter current crowding)
 - drops rapidly to zero in spacer region due to current spreading
- he
 - slightly dependent on y due to voltage drop across base resistance components



- ⇒ h_J determines the lateral distribution of h
- ⇒ hole charge contribution only from region with (vertical) current flow

Theory (cont'd)

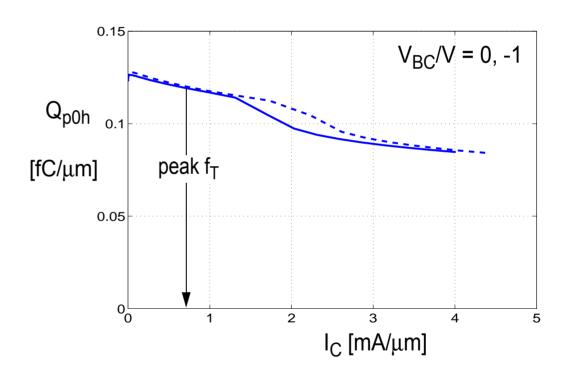
evaluation of hole density related integrals

- split p into zero-bias and bias dependent component: $p = p_0 + \Delta p$
- zero-bias weighted hole charge

$$Q_{p0h} = q2l_{E0} \int_{0}^{b_{bc}} \int_{0}^{x_{C'}} h p_0 dx dy$$

- slightly bias dependent due to "shape" change of h components
- bias dependent weigthed hole charge

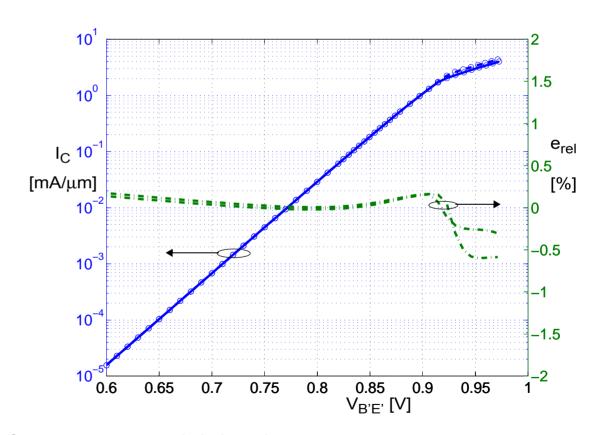
$$\Delta Q_{ph} = q2l_{E0} \int_0^{b_{bc}} \int_0^{x_{C'}} h \ \Delta p \ dxdy$$



Master equation

$$I_{T} = qb_{E}l_{E0}c_{0} \frac{\exp\left(\frac{V_{B'E'}}{V_{T}}\right) - \exp\left(\frac{V_{B'C'}}{V_{T}}\right)\frac{2b_{bc}}{b_{E}}}{Q_{p0h} + \Delta Q_{ph}}$$

- evaluation with numerical values for charges and h
- excellent agreement over whole bias region
- slight deviation (0.5%) from device simulation due to numerical integration



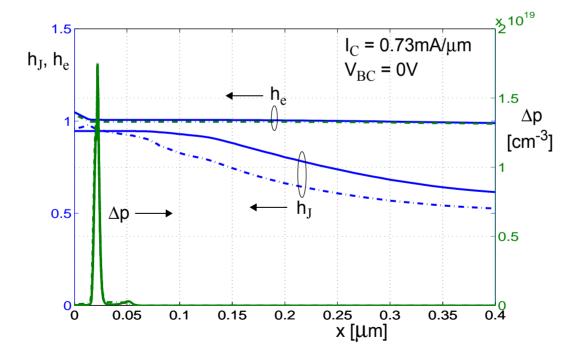
⇒ suitable for compact model development

Multi-dimensional GICCR Model equations

Model equations

• need to represent ΔQ_{ph} by measurable expression, such as ΔQ_{p} components

- partition ΔQ_p into space-charge and minority components: $\Delta Q_p = Q_{jE} + Q_{jC} + Q_m$
- split weighting function into "mostly"
 - vertical component $h_q = h_q(x,0) \implies$ treat like 1D case
 - lateral component $h_2 = h_J h_e = h_2(x_{bm}, y)$



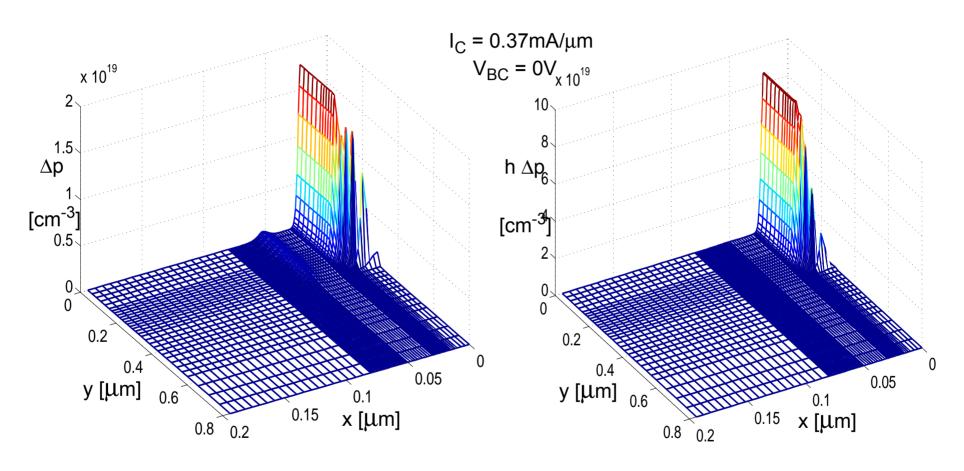
solid lines: y = 0

dashed lines: $y = b_{E0}/2$

Multi-dimensional GICCR Model equations

Appendix: Theory (cont'd)

3D distribution of Δp and h Δp in the BE junction region



⇒ suppression of part of the *perimeter* BE depletion charge by h (h_J)

Model equations (cont'd)

• example:

BC space charge
$$Q_{jCh} = 2l_{E0} \int_0^{b_{bc}} h_2 \left(q \int_{w_{bc}} h_g \Delta p dx \right) dy \cong 2l_{E0} \bar{h}_{gjC} \int_0^{b_{bc}} h_2 \bar{Q}_{jC} dy = \bar{h}_{jC} Q_{jC}$$

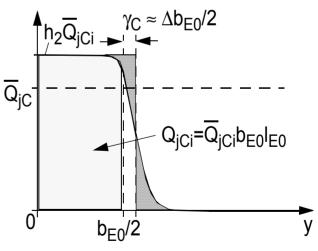
with

- $Q_{\mbox{\scriptsize iC}}$ as total (internal and external) BC space charge measurable at the terminals
- \overline{Q}_{jC} as charge per area (equals under emitter window the *internal* charge density \overline{Q}_{jCi})
- $\overline{h}_{iC} = \overline{h}_{giC}\overline{h}_2$ with bias independent \overline{h}_2 as first-order approximation
- · at low injection:

$$\begin{split} Q_{jCh} &= \bar{h}_{jC}Q_{jC} \approx \bar{h}_{gjC}Q_{jCi} \left(1 + \frac{2}{b_{E0}} \int_{b_{E0}/2}^{b_{bc}} h_2 dy\right) \\ &\approx \bar{h}_{gjC}Q_{jCi} \left(1 + \frac{\Delta b_{E0}}{b_{E0}}\right) \end{split}$$

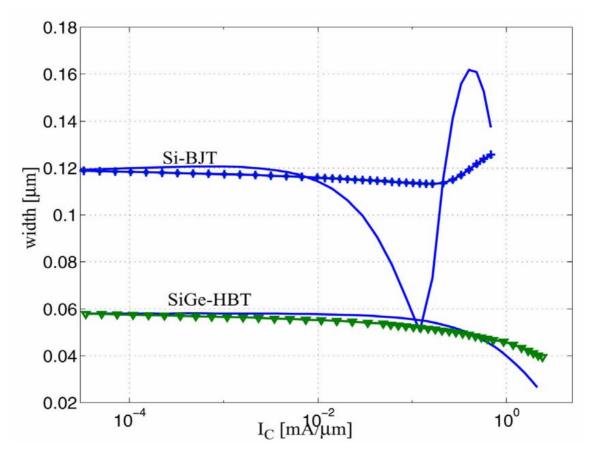
similarly:

$$Q_{p0h} \approx \bar{h}_{g0} Q_{p0i} \bigg(1 + \frac{\Delta b_{E0}}{b_{E0}} \bigg) \ , \ Q_{jEh} \ = \ \bar{h}_{gjE} Q_{jEi} \bigg(1 + \frac{\Delta b_{E0}}{b_{E0}} \bigg)$$



Model equations (cont'd)

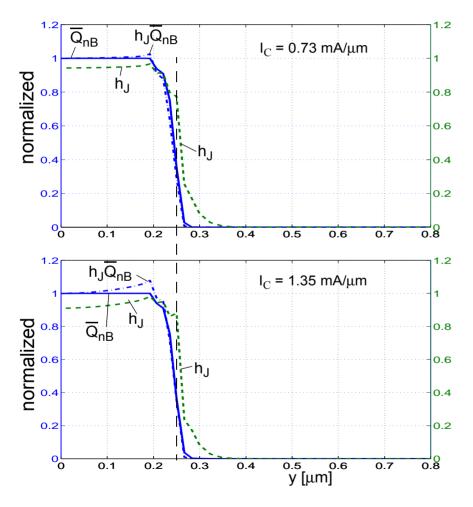
comparison between $\Delta b_{E0}/2$ and γ_{C}



- very close at low current densities
- deviations at higher current densities caused by different way of accounting for r_B

Model equations (cont'd)

minority charge



- base region: $Q_m = Q_{nB}$
- <u>internal electron charge density</u>: $\overline{Q}_{nBi} = \overline{Q}_{nB}(y=0) \implies Q_{nBi} \approx \overline{Q}_{nBi}b_{E0}l_{E0}$
- normalized electron charge density $\overline{Q}_{nB}(y)/\overline{Q}_{nBi} \propto h_2(y)$

$$\Rightarrow h_2 (\overline{Q}_{nB}(y)/\overline{Q}_{nBi}) \propto h_2^2(y)$$

$$\Rightarrow \int_0^{b_{bc}} h_2 \overline{Q}_{nB} \ dy \approx \overline{Q}_{nBi} \frac{b_{E0}}{2}$$

• first-order approximation:

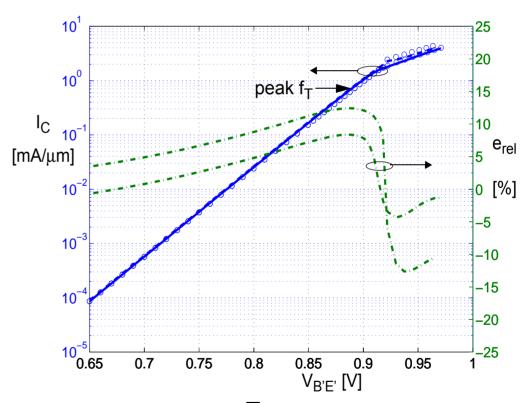
$$Q_{mh} \approx \bar{h}_{mE} Q_{pE} + \bar{h}_{mB} Q_{nB} + \bar{h}_{mC} Q_{pC}$$

with \overline{h}_{m} as $\emph{spatial}$ and \emph{bias} averages

Model equations: comparison to reference

$$\bullet \text{ model equation: } I_T = qb_E l_{E0} c_0 \frac{\exp\Bigl(\frac{V_{B'E'}}{V_T}\Bigr) - \exp\Bigl(\frac{V_{B'C'}}{V_T}\Bigr) \frac{2b_{bc}}{b_E}}{Q_{p0h} + \bar{h}_{jE}Q_{jE} + \bar{h}_{jC}Q_{jC} + Q_{mh}}$$

- numerical values for charges
- (bias) averaged weighting factors
- good agreement in relevant bias region
 - SiGe HBTs: deviation caused mostly by change of h_{.l}*h_e "form factor"
- improvements:
 - make h_x bias dependent
 - normalize to \overline{h}_{mB} to reduce parameter number and simplify modeling of weighting factors



 \Rightarrow existing compact models all use \overline{h}_x

Conclusions

- The derivation of a multi-dimensional (generalized) Integral Charge-Control Relation for the transfer current of BJTs/HBTs has been presented
- The resulting "master equation" has been verified by device simulation
- The "master equation" can serve for deriving simplified formulations for compact models
 - keep control over errors
 - well-defined evaluation of the impact of assumptions
 - clear meaning of variables and parameters (particularly as function of lateral dimensions)
- A first-order approximation of the master equation has been derived
 - · equivalent to formulation used in existing model
 - · demonstration of achievable accuracy

Acknowledgments

- financial support from
 - German Ministry of Research and Technology (SFB358, DETAILS)

- Atmel Germany, Heilbronn
- JazzSemi, Newport Beach (USA)