

A Novel Low-bias Charge Concept for HBT/BJT Models Including Heterobandgap and Temperature Effects –Part I: Theory and Model Implementation

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Outline

- BJT/HBT transfer current and (hole) charge
- the principal low bias charge formula
- overview of the AC linked model approaches
- the generic low bias charge concept
- AC linked & bandgap feature temperature dependences
- model implementation

Note: fully rigorous treatments are included in [1]



Transfer Current and Charge

$$I_{T} = q \cdot A_{E} \cdot V_{T} \cdot \frac{\exp\left(\frac{vbiei}{V_{T}}\right) - \exp\left(\frac{vbici}{V_{T}}\right)}{\int_{xE}^{xC} \frac{p(x)}{\mu_{n}(x) \cdot n_{i}^{2}(x)} \cdot dx}$$

$$Q_{pT} = q \cdot A_E \cdot \int_{xE}^{xC} \frac{\mu_n(x_R) \cdot n_i^2(x_R)}{\mu_n(x) \cdot n_i^2(x)} \cdot p(x) \cdot dx$$

$$c10 = q^2 \cdot V_T \cdot A_E^2 \cdot \mu_n(x_R) \cdot n_i^2(x_R)$$

$$exp\left(\frac{vbiei}{V_T}\right) - exp\left(\frac{vbici}{V_T}\right)$$

$$Q_{pT}$$

 $I_{T} = q \cdot A_{E} \cdot V_{T} \cdot \frac{\exp\left(\frac{vbiei}{V_{T}}\right) - \exp\left(\frac{vbiei}{V_{T}}\right)}{\int_{xE}^{xC} \frac{p(x)}{\mu_{n}(x) \cdot n_{i}^{2}(x)}}$ Derived by Moll-Ross [2] for low injection, generalized to all injection conditions by Gummel [3] as ICCR (each at Bell, but [2] not cited in [3]!)

> Extending by constants in x_R the Hicum form is obtained with the Q_{pT} hole charge in denominator

Space charge neutrality for low injection implies $p(x) \approx N_A(x) = N(x)$

Low bias hole charge: (see figure on slide #5)

$$Q_{pT_low} = q \cdot A_E \cdot \int_{xdEb}^{xdCb} \frac{\mu_n(x_R) \cdot n_i^2(x_R)}{\mu_n(x) \cdot n_i^2(x)} \cdot N(x) \cdot dx$$



Low bias hole charge

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The effective doping density has been suggested in [4] as

$$N_{eff}(x) = \frac{\mu_n(x_R) \cdot n_i^2(x_R)}{\mu_n(x) \cdot n_i^2(x)} \cdot N(x) = m(x, T) \cdot N(x)$$

The Moll-Ross-Gummel or MRG function is introduced by

$$m(x,T) = m_{\mu}(x) \cdot \exp\left[-\frac{T_0}{T} \cdot \frac{\Delta V_G(x) - \Delta V_G(x_R)}{V_{T0}}\right] \qquad m_{\mu}(x) = \frac{\mu_n(x_R, T_0)}{\mu_n(x, T_0)}$$

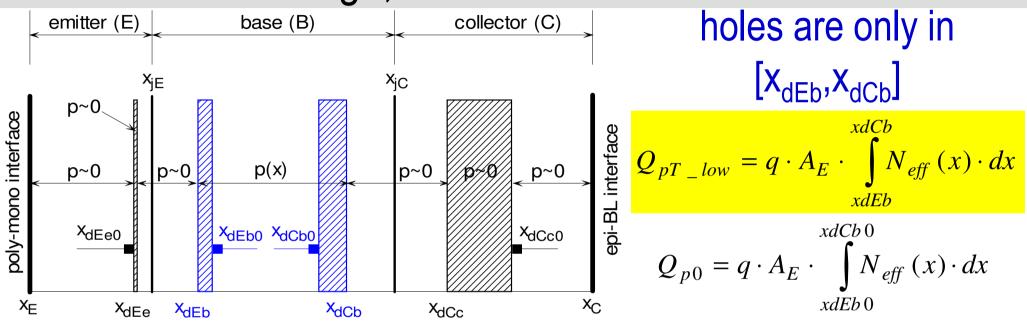
Variable bandgap in the base makes the MRG function temperature dependent.

An analytic technique – not used so far - will be adopted for model construction enabling a consistent discussion of the bias and temperature controlled operation modes



Low bias hole charge, cont.'d

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At low bias in the forward active mode the hole density is negligible in the emitter and collector *n*-type quasineutral regions. Shaded rectangles represent the electrostatic (ES) charge of the doping centers when the depletion layer boundaries are displaced from their equilibrium (zero subscripted) positions.



The principal low bias charge formula

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Apply a small forward bias increment to the BE junction

$$\frac{\partial Q_{pT_low}}{\partial vbiei} = -q \cdot A_E \cdot \frac{\partial x_{dEb}}{\partial vbiei} \cdot N_{eff}(x_{dEb}) = -m(x_{dEb}) \cdot \frac{A_E \cdot \Delta x_{dEb} \cdot \left[-q \cdot N_A(x_{dEb}) \right]}{\Delta vbiei} = m(x_{dEb}) \cdot \frac{\Delta \tilde{Q}_{jEi}}{\Delta vbiei}$$

$$\frac{\partial Q_{pT_low}}{\partial vbiei} = m(x_{dEb}(vbiei)) \cdot \tilde{C}_{jEi}(vbiei) \sim \text{denotes AC related quantities}$$

Similarly for the BC side

 $\frac{\partial Q_{pT_low}}{\partial vbici} = m(x_{dCb}(vbici)) \cdot \tilde{C}_{jCi}(vbici)$ pair of PDEs has the solution

As long as the depletion approximation is accepted the two forms are equivalent

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Junction related charges: terminology and implication in analog

$$\tilde{Q}_{jEi} = q \cdot A_E \int_{xdEb}^{xdEb} N(x) \cdot dx = \int_{o}^{vbiei} \tilde{C}_{jEi}(u) \cdot du$$

Emitter AC (ES) charge

$$\widetilde{Q}_{jCi} = q \cdot A_E \int_{xdCb}^{xdCb} N(x) \cdot dx = \int_{0}^{vbici} \widetilde{C}_{jCi}(u) \cdot du$$

Collector AC (ES) charge

$$\overline{Q}_{jE} = q \cdot A_E \int_{\substack{xdEb \ xdCb}}^{xdEb} N_{eff}(x) \cdot dx = \int_{o}^{vbiei} m(x_{dEb}(u)) \cdot \tilde{C}_{jEi}(u) \cdot du$$
 Emitter MRG charge
$$\overline{Q}_{jC} = q \cdot A_E \int_{xdCb}^{xdCb} N_{eff}(x) \cdot dx = \int_{vbici}^{vbici} m(x_{dCb}(u)) \cdot \tilde{C}_{jCi}(u) \cdot du$$
 Collector MRG charge

The MRG charges are the weighted integrals of the capacitances.

A linear MRG to AC charge relationship assumed in all present models is not a mandatory approximation!



Approximation of the MRG charges by averages

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Second mean value theorem of integrals yields

$$\overline{m}_{J} = \frac{\overline{Q}_{jJ}}{\overline{Q}_{jJi}} = \int_{o}^{vbiJi} m(x_{dJb}(u)) \cdot \widetilde{C}_{jJi}(u) \cdot du / \int_{o}^{vbiJi} \widetilde{C}_{jJi}(u) \cdot du \quad J = E, C$$

$$Q_{pT_low} = Q_{p0} + \overline{m}_E \cdot \tilde{Q}_{jEi}(vbiei) + \overline{m}_C \cdot \tilde{Q}_{jCi}(vbici)$$

Normalized charge function at *nominal temperature*

$$\Phi_{JT0} = \frac{1}{1 - zJi} \cdot \left[1 - \left(1 - \frac{vbiJi}{vdJi_0} \right)^{1 - zJi} \right]$$

GPM, VBIC and MEXTRAM use normalized formulations

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AC linked approaches

Transition voltages: $v_E(vbiei) = \tilde{v} dei_0 \cdot \tilde{\Phi}_{ET0}$ $v_C(vbici) = \tilde{v} dci_0 \cdot \tilde{\Phi}_{CT0}$

Early voltages: $ver = \frac{Q_{p0}}{\overline{m}_E \cdot \tilde{C}_{jEi0}(T_0)} \quad vef = \frac{Q_{p0}}{\overline{m}_C \cdot \tilde{C}_{jCi0}(T_0)}$

Normalized charge: $Q_{pT_low} = Q_{p0} \cdot q_{pT}$ $q_{pT} = 1 + \frac{v_E(vbiei)}{ver} + \frac{v_C(vbici)}{vef}$

Hicum looks different: $Q_{pT_low} = Q_{p0} + hjei \cdot \tilde{Q}_{jEi}(vbiei) + hjci \cdot \tilde{Q}_{jCi}(vbici)$

$$hj_{Ji} = q \cdot A_E \int_{xdJb}^{xdJb \, 0} m(x) \cdot N(x) \cdot dx / q \cdot A_E \int_{xdJb}^{xdJb \, 0} N(x) \cdot dx = \frac{\overline{Q}_{jJ}}{\widetilde{Q}_{jJi}} = \overline{m}_J$$

From Q_{pT_low} aspect GPM==VBIC==MEXTRAM ==HICUM It is justified to term these approaches "AC linked"

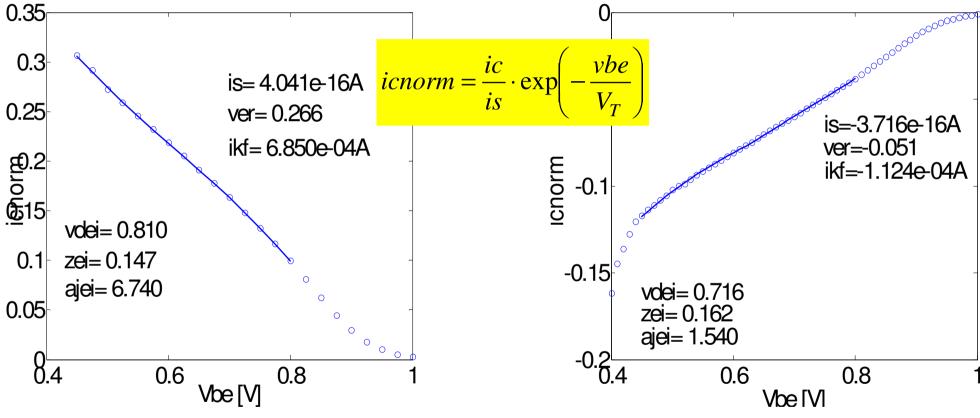
since ES capacitances & charges are measured by AC methods



Problems with AC linked approaches, all models

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Negative IS, VER were reported in [4] for a 250GHz device



70GHz device, falling curve is, ver can be extracted

250 GHz device, rising curve is, ver can not be extracted



The generic low bias charge approach, primitives leap ahead in analog No theoretical coercion exists which should demand an AC charge term in the related MRG charge expression

- Maxwell equation $div(\mathbf{D}) = \sigma$ linking electrical displacement to space charge is valid for any kinds of spatial charge distributions
- 1D integral form –Gauss' theorem implies depletion capacitance and charge relations
- Semi-empirical capacitance relations do not restrict the shape of the charge distribution i.e. shape of the doping concentration N(x)
- Consequently C-Q type functions can be constructed to any N(x) incl. the effective doping concentration Neff(x) as well [4]

$$\frac{Q_{pT}}{c10} = \frac{1}{I_T} \left[\exp\left(\frac{vbiei}{V_T}\right) - \exp\left(\frac{vbici}{V_T}\right) \right] \quad \text{parameters incorporated in } Q_{pT} \text{ can be extracted from the known RHS}$$



The generic low bias charge approach, formulation analog

$$Q_{pT_low} = Q_{p0} + q \cdot A_E \cdot \int_{xdEb}^{xdEb0} N_{eff}(x) \cdot dx + q \cdot A_E \cdot \int_{xdCb0}^{xdCb} N_{eff}(x) \cdot dx$$

$$Q_{pT_low} = Q_{p0} + \overline{Q}_{jE} + \overline{Q}_{jC}$$

- MRG charge parameters are extracted from the DC data
- Moderate V_B shift by Ge content directly addressed at extraction
- MRG "depletion boundaries" differ from AC ones but this is indifferent from DC point of view (positions do not control anything)

Each junction shall be attributed a pair of independent MRG and AC charge component to, the former responsible for the DC, the latter for the AC behaviour



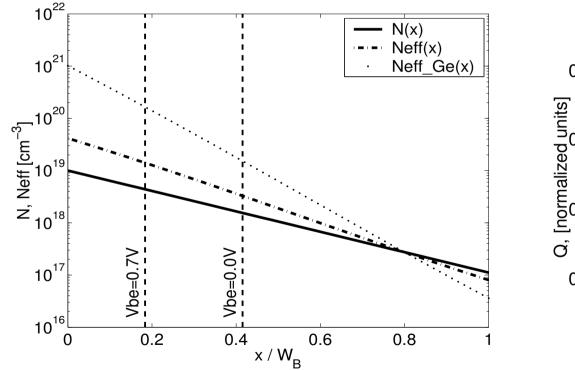
The *generic* low bias charge approach, example

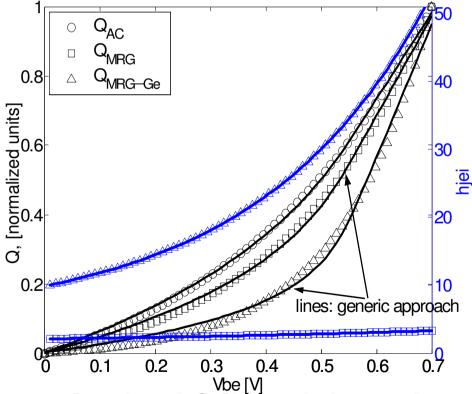
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Abrupt exponential EB junction was numerically computed with

$$N_{D0} = 2.10^{20} cm^{-3} N_A(x) = N_{A0} \cdot \exp(-\eta \cdot x / W_B) N_{A0} = 1.10^{19} cm^{-3} \eta = 4.5 W_B = 50 nm$$

 $x_R = 0.8 W_B \Delta V_G / V_{T0} = 4$, linear ramp





Charge curves on RHS plot should overlap by AC linked theory!



Temperature dependence: homobandgap

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$$C_{j}(T) = C_{j0}(T_{0}) \cdot \left(\frac{v_{tot}}{vd_{0}}\right)^{-z} \quad Q_{j}(T) = Q_{jnom} \cdot \Phi_{T} \quad \Phi_{T} = \frac{1}{1-z} \left[\left(\frac{vd}{vd_{0}}\right)^{1-z} - \left(\frac{v_{tot}}{vd_{0}}\right)^{1-z}\right]$$

$$v_{tot} = vd - v$$
 $Q_{jnom} = C_{j0}(T_0) \cdot vd_0$ to be used below without verification

Total derivative w.r.t. temperature: $\frac{dQ_{pT_low}}{dT} = \frac{dQ_{jE}}{\Delta T} + \frac{dQ_{jC}}{\Delta T}$

$$\frac{dQ_{pT_low}}{dT} = \frac{\partial Q_{jE}}{\partial T} + \frac{\partial Q_{jC}}{\partial T}$$

The emitter part from the AC linked approach reads

$$\frac{\partial Q_{jE}}{\partial T} = -q \cdot A_E \cdot \frac{\partial x_{dEb0}}{\partial T} \cdot N_{eff} (x_{dEb0}) + \overline{m}_E \cdot \int_{o}^{vbiei} \frac{\partial \widetilde{C}_{jEi}(u)}{\partial T} \cdot du$$

Integrand:

$$\frac{\partial \tilde{C}_{jEi}}{\partial T} = \frac{\partial \tilde{C}_{jEi}}{\partial vbiei} \cdot \frac{\partial vbiei}{\partial v_{tot}} \cdot \frac{\partial v_{tot}}{\partial T} = -\frac{\partial \tilde{C}_{jEi}}{\partial vbiei} \cdot \frac{d(vdei)}{dT}$$



Temperature dependence: homobandgap, cont.'d leap ahead in analog

First term:

$$\begin{split} -q \cdot A_{E} \cdot \frac{\partial x_{dEb\,0}}{\partial T} \cdot N_{eff} \left(x_{dEb\,0} \right) &= m(x_{dEb\,0}) \cdot \frac{\partial \widetilde{Q}_{jE\,0}}{\partial v biei} \cdot \frac{\partial v biei}{\partial v_{tot}} \cdot \frac{\partial v_{tot}}{\partial T} = \\ -m(x_{dEb\,0}) \cdot \widetilde{C}_{jEi\,0} \cdot \frac{d \left(v dei \right)}{dT} \end{split}$$

Substitution reveals a new model parameter and a PDE w.r.t. vdei

$$\delta_E = \frac{m(x_{dEb\,0})}{\overline{m}_E}$$

$$\delta_{E} = \frac{m(x_{dEb\,0})}{\overline{m}_{E}} \qquad \qquad \frac{\partial Q_{jE}}{\partial (vdei)} = -\overline{m}_{E} \cdot (\delta_{E} \cdot \widetilde{C}_{jEi\,0} + \widetilde{C}_{jEi\,0} - \widetilde{C}_{jEi\,0})$$

Similarly for the collector side

$$\delta_C = \frac{m(x_{dCb\,0})}{\overline{m}_C}$$

$$\delta_{C} = \frac{m(x_{dCb\,0})}{\overline{m}_{C}} \qquad \qquad \frac{\partial Q_{jC}}{\partial (vdci)} = -\overline{m}_{C} \cdot (\delta_{C} \cdot \widetilde{C}_{jCi\,0} + \widetilde{C}_{jCi\,0} + \widetilde{C}_{jCi\,0})$$

Solution of the pair of PDEs must contain a bias and temperature independent constant Q_{p0} redefined to

"quiescent hole charge"



Temperature dependence: final result

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Integrals can be performed w.r.t. the built in voltages yielding

$$Q_{pT_low} = Q_{p0} + \overline{m}_E \cdot \widetilde{Q}_{jEnom} \cdot \widetilde{\Gamma}_{ET} + \overline{m}_C \cdot \widetilde{Q}_{jCnom} \cdot \widetilde{\Gamma}_{CT} \qquad \Gamma_{JT} = \frac{\delta_J}{1 - zji} \cdot \left| 1 - \left(\frac{vdji_T}{vdji_0} \right)^{1 - zji} \right| + \Phi_{JT}$$

$$\Gamma_{JT} = \frac{\delta_J}{1 - zji} \cdot \left[1 - \left(\frac{vdji_T}{vdji_0} \right)^{1 - zji} \right] + \Phi_{JT}$$

$$\operatorname{Hicum/L2v24:} \ Q_{pT_low} = Q_{p0} \cdot \left[2 - \left(\frac{vdei_T}{vdei_0} \right)^{zei} \right] + hjei \cdot \tilde{Q}_{jEnom} \cdot \tilde{\Phi}_{ET} + hjci \cdot \tilde{Q}_{jCnom} \cdot \tilde{\Phi}_{CT}$$

Neither the E nor the C terms agree. Inserting an additional segment in the undepleted base region, Q_{p0} should change but its temperature dependence should not since the added segment is *T*-independent

The *generic* approach yields with $\overline{m}_E = \overline{m}_C = 1$

$$Q_{pT_low} = Q_{p0} + \overline{Q}_{jEnom} \cdot \overline{\Gamma}_{ET} + \overline{Q}_{jCnom} \cdot \overline{\Gamma}_{CT}$$

Model parameters δ_J change accordingly but all functional forms stay the same



Temperature dependence: heterobandgap

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Temperature variation of the quiescent hole charge:

$$\frac{\partial Q_{p0}}{\partial T} = q \cdot A_E \cdot \int_{xdEb0}^{xdCb0} \frac{\partial \ln(m(x))}{\partial T} \cdot N_{eff}(x) \cdot dx = \frac{d}{dT} \left(-\frac{T_0}{T} \right) \cdot q \cdot A_E \cdot \int_{xdEb0}^{xdCb0} \frac{\Delta V_G(x) - \Delta V_G(x_R)}{V_{T0}} \cdot N_{eff}(x) \cdot dx$$

Approximately 7mV bandgap reduction can be achieved in Si by adding 1% Ge contaminant. A typical Ge ramp of 15% yields approximately $\frac{\Delta V_G}{V_{T0}} = 4$ what is small to $N_{eff}(x)$ allowing a separation

$$\zeta_{G0} = \int_{xdEb0}^{xdCb0} \left[\frac{\Delta V_G(x) - \Delta V_G(x_R)}{V_{T0}} \right] \cdot N_{eff}(x) \cdot dx / \int_{xdEb0}^{xdCb0} N_{eff}(x) \cdot dx$$

Resulting solution of the PDE $\frac{\partial Q_{p0}}{\partial T} = \frac{d}{dT} \left(-\frac{T_0}{T} \right) \cdot \zeta_{G0} \cdot Q_{p0}$ reads

$$Q_{p0}^{G} = Q_{p0} \cdot \exp\left[\left(1 - \frac{T_0}{T}\right) \cdot \zeta_{G0}\right]$$



Temperature dependence: heterobandgap, cont. 'd' ap ahead in analog

Temperature variation of the emitter MRG charge is approximately:

$$\frac{\partial \overline{Q}_{jE}}{\partial T} \approx q \cdot A_E \int_{xdEh}^{xdEb \ 0} \frac{\partial \ln(m(x))}{\partial T} \cdot N_{eff}(x) \cdot dx$$

As before, it yields for the *E* and *C* nominal MRG charges

$$\boxed{Q_{jEnom}^G = \overline{Q}_{jEnom} \cdot \exp \left[\left(1 - \frac{T_0}{T} \right) \cdot \zeta_{GE} \right]} \qquad \boxed{Q_{jCnom}^G = \overline{Q}_{jCnom} \cdot \exp \left[\left(1 - \frac{T_0}{T} \right) \cdot \zeta_{GC} \right]}$$

The integration limits x_{dEb} and x_{dCb} also change in the PDEs what have been neglected here as second orders. It is proven in [1] though that the above temperature variations are exact for \overline{C}_{iE0} and \overline{C}_{iC0}

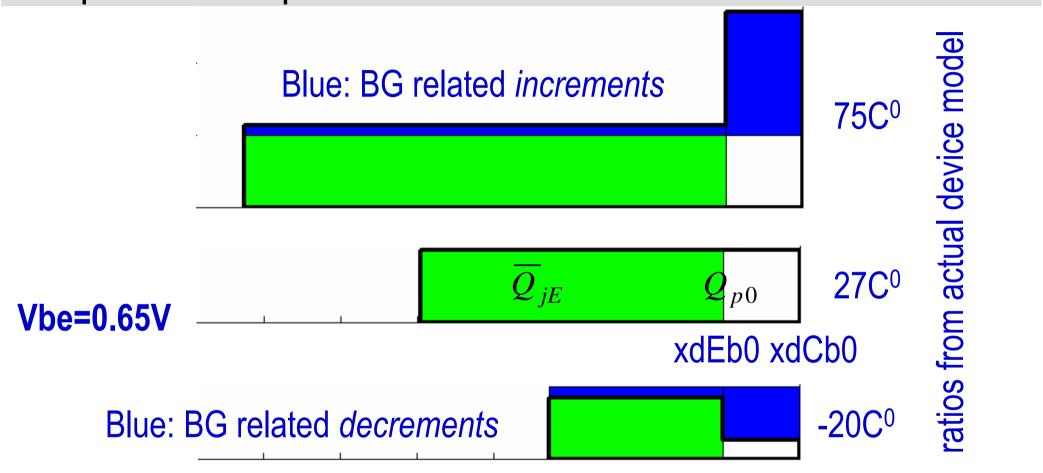
$$Q_{pT_low}^G = Q_{p0}^G + Q_{jEnom}^G \cdot \overline{\Gamma}_{ET} + Q_{jCnom}^G \cdot \overline{\Gamma}_{CT}$$

This general result has been achieved w/o assuming any specific Ge profile as opposed to [6]



Temperature dependence: illustration

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Net Q_{pT_low} resides in black frames (Q_{jC} was negligible this case) Junction and BG related temperature variations are "orthogonal"



Temperature dependence: low bias transit time and Qin analog

Bandgap induced temperature variation affects the low bias transit time as well. With approximation for large drift factors

$$t_0 = \frac{W_B^2}{\mu_{n0} \cdot V_{T0}} \cdot \left(\frac{T_0}{T}\right)^{1-\zeta_n} \cdot f(\eta_{eff}) \approx \frac{W_B^2}{\mu_{n0} \cdot V_{T0}} \cdot \left(\frac{T_0}{T}\right)^{1-\zeta_n} \cdot \frac{1}{\eta_{eff}}$$

The effective drift factor at a linear Ge ramp reads

$$\eta_{\it eff} = \eta + \frac{T_0}{T} \cdot \frac{\Delta V_G}{V_{T0}}$$

This adds a mild further variation to the homobandgap case what can be conveniently absorbed by the existing quadratic temperature mapping

$$t_0(T) = t_0(T_0) \cdot \left[1 + alt \, 0 \cdot (T - T_0) + kt \, 0 \cdot (T - T_0)^2 \right]$$

No change in t_0 and Q_{f0} is required



Temperature dependence: weight of E minority charge analog

$$hfe(T) = \frac{\int_{x_E}^{x_{jE}} m_E(x,T) \cdot p_E(x) \cdot dx}{\int_{x_E}^{x_{jE}} p_E(x) \cdot dx} \approx \frac{\mu_n(x_R)}{\mu_n(\xi_E)} \cdot \exp\left[\frac{T_0}{T} \cdot \frac{\Delta V_G(x_R) - \Delta V_{GE}}{V_{T0}}\right] \quad x_E < \xi_E < x_{jE} \quad \Delta V_{GE} = BGN_E$$

$$tef \ 0(T) = tef \ 0(T_0) \cdot \left(\frac{T_0}{T}\right)^{-zetatef} \cdot \exp\left[-\frac{vgb - vge}{V_{T0}} \cdot \left(1 - \frac{T_0}{T}\right)\right] \ \text{std. temperature dependence}$$

$$\mathit{hfe}(T) \cdot \Delta Q_{\mathit{Ef}}(T) = \mathit{hfe}(T) \cdot \mathit{tef} \ 0(T) \cdot (\mathsf{other \ terms}) \\ \propto \left(\frac{T_0}{T}\right)^{-\mathit{zetatef}} \cdot \exp \left[-\frac{\mathit{vgb} - \mathit{vge}_{\mathit{eff}}}{V_{T0}} \cdot \left(1 - \frac{T_0}{T}\right)\right]$$

$$tef \ 0(T) = tef \ 0(T_0) \cdot \left(\frac{T_0}{T}\right)^{-zetatef} \cdot \exp\left[-\frac{vgb - vge_{eff}}{V_{T0}} \cdot \left(1 - \frac{T_0}{T}\right)\right]$$
 modified vge

hfe stays constant with its T-dependence absorbed by tefO(T)



Temperature dependence: weight of C minority charge analog

$$hfc(T) = \frac{\int_{x_{jC}}^{x_{C}} m_{C}(x,T) \cdot p_{C}(x) \cdot dx}{\int_{x_{jC}}^{x_{C}} p_{C}(x) \cdot dx} \approx \frac{\mu_{n}(x_{R})}{\mu_{n}(\xi_{C})} \cdot \exp\left[\frac{T_{0}}{T} \cdot \frac{\Delta V_{G}(x_{R})}{V_{T0}}\right] \quad x_{jC} < \xi_{C} < x_{C} \quad BGN_{C} \approx 0$$

$$(T_{0})^{1-zetaci} \quad thcs(T) \text{ can only partially abso}$$

$$thcs(T) = thcs(T_0) \cdot \left(\frac{T_0}{T}\right)^{1-zetaci}$$

thcs(T) can only partially absorb the bandgap feature effects

$$thcs(T) = thcs(T_0) \left(\frac{T_0}{T}\right)^{1-zetaci} \cdot \exp\left[-\frac{\Delta V_{GR}}{V_{T0}} \cdot \left(1 - \frac{T_0}{T}\right)\right] \frac{\text{modified } thcs(T) \text{ would allow leaving } htc}{\text{leaving } htc} \cdot \exp\left[-\frac{\Delta V_{GR}}{V_{T0}} \cdot \left(1 - \frac{T_0}{T}\right)\right]$$

Either *hfc* should be scaled as shown or the temperature dependence of *thcs* should be modified (preferred). Needs more experiment or TCAD to decide: no modification in this framework.



Temperature dependence: base resistance

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The quiescent part of the stored *minority charge* with typically $f_{DQr0} < 0$

$$\tilde{Q}_0 = q \cdot A_E \int_{xdEb0}^{xdCb0} N(x) \cdot dx \le q \cdot A_E \int_{xdEb0}^{xdCb0} N_{eff}(x) \cdot dx = Q_{p0} \quad \Longrightarrow \quad \tilde{Q}_0 = (1 + f_{DQr0}) \cdot Q_{p0}$$

The conductivity modulation is to be re-written by the *total* AC junction related charges referenced to the **quiescent** depletion boundaries [5]

$$r_{i} = r_{Bi0} \cdot \frac{\tilde{Q}_{0}}{\tilde{Q}_{0} + \tilde{Q}_{jEtot} + \tilde{Q}_{jCtot} + Q_{f}} \quad Q_{jtot} = \frac{Q_{jnom}}{1 - z} \cdot \left[1 - \left(\frac{v_{tot}}{vd_{0}}\right)^{1 - z}\right]$$

Refines temperature dependence. *rbi* is overaddressed in the model: **change is justified only along with other due updates**

Tetrode structures are not able to detect the bandgap related temperature variation of Q_{p0}



Model implementation: HICUM/L2v2.24G

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$$n_i^2(x_R, T) = n_{i,Si}^2 \cdot \exp\left(\frac{\Delta V_G(x_R)}{V_{T0}} \frac{T_0}{T}\right)$$

The bandgap shrink in the reference point has been tacitly merged in vgb since the introduction of the model thus c_{10} will be kept as it is.

$$I_{T} = c_{10} \cdot \frac{\exp\left(\frac{vbiei}{mcf \cdot V_{T}}\right) - \exp\left(\frac{vbici}{V_{T}}\right)}{Q_{p0}^{G} + h_{jei}^{G} \cdot Q_{p0} \cdot \overline{\Gamma}_{ET} + h_{jci}^{G} \cdot Q_{p0} \cdot \widetilde{\Gamma}_{CT} + Q_{fT}(I_{Tf}, I_{Tr})}$$

$$h_{jei} = \frac{\overline{Q}_{jEnom}}{Q_{p0}} \quad h_{jci} = \frac{\widetilde{Q}_{jCnom}}{Q_{p0}} \qquad h_{jei}^{G} = h_{jei} \cdot \exp\left[\left(1 - \frac{T_{0}}{T}\right) \cdot \zeta_{GE}\right]$$

$$Q_{p0}^{G} = Q_{p0} \cdot \exp\left[\left(1 - \frac{T_0}{T}\right) \cdot \zeta_{G0}\right] \qquad h_{jci}^{G} = h_{jci} \cdot \exp\left[\left(1 - \frac{T_0}{T}\right) \cdot \zeta_{GC}\right]$$

Reverse active mode of operation is usually irrelevant hence the C side has been left AC linked. No other changes than indicated.



HICUM/L2v2.24G: alternative IS parameter

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This model formulation allows for "pseudo-normalization" by specifying *is#0* which case *c10=is*qp0* is internally computed by the code.

nominal temperature:

$$I_{T} = \frac{is \cdot Q_{p0} \cdot \left[\exp\left(\frac{vbiei}{mcf \cdot V_{T}}\right) - \exp\left(\frac{vbici}{V_{T}}\right) \right]}{Q_{p0} + h_{jEi} \cdot Q_{p0} \cdot \overline{\Gamma}_{ET} + h_{jCi} \cdot Q_{p0} \cdot \widetilde{\Gamma}_{CT} + Q_{fT} \left(I_{Tf}, I_{Tr}\right)}$$

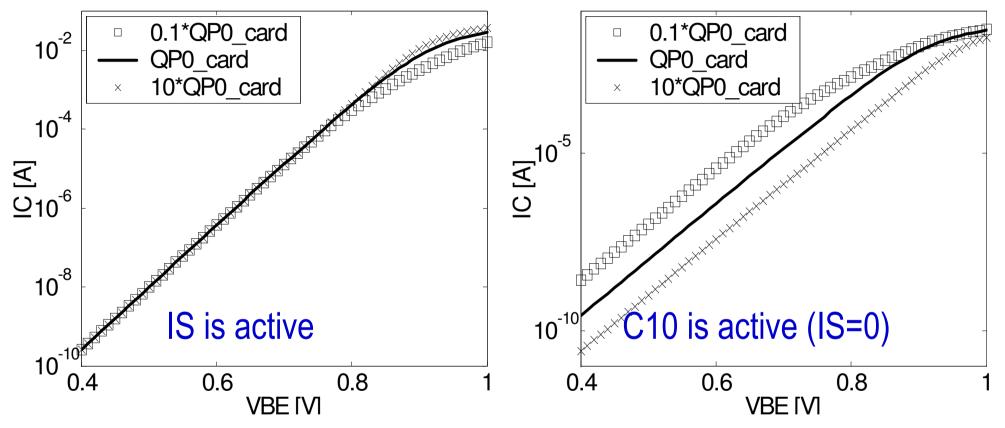
 Q_{fT} vanishes at low currents implying Q_{p0} to cancel out. Extraction of or importing *is*, *hjei* and *hjci* becomes a straightforward procedure. This will be essential for bandgap features parameter extraction.

The correlation nightmare among *c10*, *qp0*, *hjei*, *hjci* existing in the stdandard Hicum/L2 formulations at low bias has been removed, making the model uniquely extractable for these parameters



HICUM/L2v2.24G: pseudo-normalization

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Vbe<= 0.7 interval is invariant to range of order qp0 variations. qp0 is a medium-current parameter: its extraction by low bias methods like using tetrode structures might be error prone



Additional model parameters, HICUM/L2v2.24G a leap ahead in analog

no	name	description	default	range	unit	factor
1	delte	coefficient of the zero bias emitter MRG charge temperature variation	0	[0:Inf)	-	-
2	deltc	coefficient of the zero bias collector MRG charge temperature variation	0	[0:Inf)	-	-
3	zetag0	bandgap parameter of the quiescent hole charge	0	[-10:10]	-	-
4	zetage	bandgap parameter of the emitter MRG charge	0	[-10:10]	-	-
5	zetagc	bandgap parameter of the collector MRG charge	0	[-10:10]	-	-
6	zedc	grading factor of the emitter MRG charge	0.9	[0:1)	-	-
7	vdedc	built in potential of the emitter MRG charge	0.9	(0:10]	V	-
8	ajedc	ratio maximum to zero bias value of the B-E MRG charge voltage gradient	10	[0:Inf)	-	-
9	is	saturation current (alternative to c10)	0	[0:1]	Α	M



Model implementation, HICUM/L0v1.2G

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$$ver = \frac{Q_{p0}}{\overline{Q}_{jEnom}} = \frac{1}{hjei} \quad ver^{G} = ver \cdot \exp\left[\left(1 - \frac{T_{0}}{T}\right) \cdot \zeta_{VER}\right] \quad \zeta_{VER} = \zeta_{G0} - \zeta_{GE}$$

$$vef = \frac{Q_{p0}}{\widetilde{Q}_{jCnom}} = \frac{1}{hjci} \quad vef^{G} = vef \cdot \exp\left[\left(1 - \frac{T_{0}}{T}\right) \cdot \zeta_{VEF}\right] \quad \zeta_{VEF} = \zeta_{G0} - \zeta_{GC}$$

$$\zeta_{IQF} = \zeta_{G0} \quad \exp_{T} = \exp\left[\left(1 - \frac{T_{0}}{T}\right) \cdot \zeta_{IQF}\right]$$

$$iqf^{G} = iqf \cdot \exp_{T} \quad iqr^{G} = iqr \cdot \exp_{T} \quad iqfh^{G} = iqfh \cdot \exp_{T} \quad iqfe^{G} = iqfe \cdot \exp_{T} \quad (iqfe = iqfh / tfh)$$

$$is^{G}(T) = is(T_{0}) \cdot \left(\frac{T}{T_{0}}\right)^{\zeta_{CT}} \cdot \exp\left[\left(\frac{vgb}{V_{T0}} - \zeta_{IQF}\right) \cdot \left(1 - \frac{T_{0}}{T}\right)\right] I_{T} = is^{G} \cdot \frac{\exp\left(\frac{vbiei}{mcf \cdot V_{T}}\right) - \exp\left(\frac{vbici}{mcr \cdot V_{T}}\right)}{q_{pT}^{G}}$$

This looks fairly complicated: normalization smears the bandgap effects all over the model. Theoretical advantage of the provident omission of normalization in HICUM/L2 is evident.



Additional model parameters, HICUM/L0v1.2G

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no	name	description	default	range	unit	factor
1	delte	coefficient of the zero bias emitter MRG charge temperature variation	0	[0:Inf)	1	-
2	deltc	coefficient of the zero bias collector MRG charge temperature variation	0	[0:Inf)	ı	-
3	zetaver	bandgap parameter of the reverse Early coefficient	0	[-10:10]	ı	-
4	zetavef	bandgap parameter of the forward Early coefficient	0	[-10:10]	ı	-
5	ibhrec	specific BC barrier recombination current	0	[0:1]	Α	M
	iqfe	roll-off current of the emitter charge component, 0 is Inf (replaces tfh by iqfe=iqfh/tfh)	0	[0:1]	Α	М



Summary

- a principal low bias charge expression has been derived relying on the widely accepted depletion approximation
- the AC linked approach has been shown to be only an option with the risk of implying model misfit at advanced SiGe devices
- a generic approach has been suggested for the MRG charges
- novel quiescent low bias charge concept and consistent bandgap feature temperature model have been proposed
- model was implemented in MATLAB and benchmarked to the modified VA codes Hicum/L2v2.24G and Hicum/L0v1.2G



Appreciation

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It is honourable mention the vision of the Hicum originators

Prof. Dr.-Ing. H. M. Rein and Prof. Dr.-Ing. M. Schröter

for having proposed a model such that it seamlessly allows for extensions not relevant at the time of its native formulation

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