

# Extraction Method for Thermal Resistance Bipolar-Arbeitskreis Heilbronn, Nov 2004

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- Teststructures for temperature measurement
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# Temperature Measurement

Dependent on the extraction method used different requirements arise:

- DUT and separate sensor
  - some distance from DUT
  - usually requires calibration
  - example : diode
- DUT merged with sensor
  - requires special layout
  - and calibration
  - see example
- DUT as sensor
  - uses regular device
  - best choice



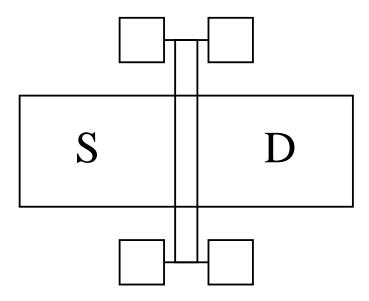


Figure 1: top view of experimental teststructure for temperatue measurement

Principle of teststructure for temperature measurement in SOI MOS-FET's after [1].

The gate is configured for four-point resistance measurement with calibration in off-state of the MOS.

During operation of the MOS this resistance serves as temperature sensor.



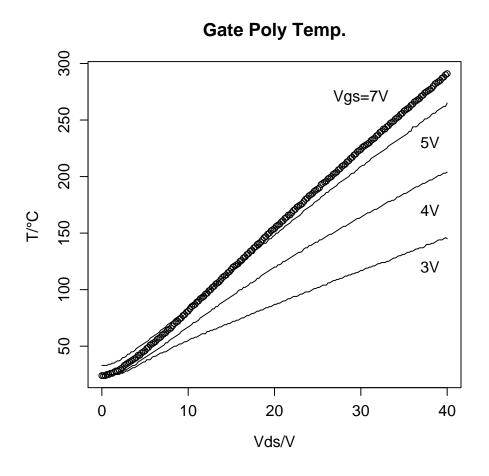


Figure 2: experimental results of SOI NMOS, w=  $320 \mu m$ 

There are several extraction methods suitable for regular devices [2, 3, 4, 5]. Of particular interest is the method from [5], which allows the extraction of thermal resistance in dependence on ambient temperature.

Thermal conductivity changes with temperature ( $\alpha \sim 1.5$ ):

$$\kappa(T) = \kappa_{ref} \left(\frac{T}{T_{ref}}\right)^{-\alpha} \tag{1}$$

Since thermal resistance is inversely proportional to the thermal conductivity, we should expect as claimed in [5]:

$$R_{TH}(T) = R_{TH,ref} \left(\frac{T}{T_{ref}}\right)^{\alpha} \tag{2}$$



Total temperature change caused by ambient and power dissipation [4]:

$$\Delta T = R_{TH}(I_C \Delta V_C + \Delta I_C V_C) + \Delta T_A \tag{3}$$

$$\Delta I_C = TC_F(I_C)I_C\Delta T + \frac{\Delta V_C}{V_A}I_C \tag{4}$$

Keeping  $V_C$  constant :

$$\Delta T|_{\Delta V_C = 0} = R_{TH} V_C \Delta I_C + \Delta T_A \tag{5}$$

$$=\frac{\Delta T_A}{1 - TC_F(I_C)R_{TH}I_CV_C} \tag{6}$$

Keeping  $T_A$  constant :

$$\Delta T|_{\Delta T_A=0} = R_{TH} (I_C \Delta V_C + \Delta I_C V_C \tag{7}$$

$$= \frac{I_C R_{TH} (1 + \frac{V_C}{V_A}) \Delta V_C}{1 - T C_F (I_C) R_{TH} I_C V_C}$$
 (8)

### Variation of $I_C$ :

$$\Delta I_C = \frac{\partial I_C}{\partial T} \Delta T \tag{9}$$

$$= TC_F(I_C)I_C\Delta T \tag{10}$$

$$= TC_F(I_C)I_C(\Delta T|_{\Delta V_C=0} + \Delta T|_{\Delta T_A=0})$$

$$\tag{11}$$

$$= \frac{TC_F(I_C)I_C}{1 - TC_F(I_C)R_{TH}I_CV_C} (\Delta T_A + I_C R_{TH}(1 + \frac{V_C}{V_A})\Delta V_C)$$
 (12)

### Extraction of $R_{TH}$ :

$$\frac{\frac{\Delta I_C}{\Delta V_C}|_{\Delta T_A=0}}{\frac{\Delta I_C}{\Delta T_A}|_{\Delta V_C=0}} = I_C R_{TH} (1 + \frac{V_C}{V_A}) \tag{13}$$

$$R_{TH} = \frac{I_C(V_C + \Delta V_C, T_A) - I_C(V_C - \Delta V_C, T_A)}{I_C(V_C, T_A + \Delta T_A) - I_C(V_C, T_A - \Delta T_A)} \frac{\Delta T_A}{\Delta V_C} \frac{1}{I_C(1 + \frac{V_C}{V_A})}$$
(14)

This is eq. 3 of [5] using  $I_C$  instead of  $I_B$  as sensored signal.

## Measurement conditions:

- DUT : SiGe2\_power npn, 4x 19.7  $\mu m$  x 1.3  $\mu m$
- $V_C = 2V, \Delta V_C = 0.2V$
- $I_C \sim 1.5 mA$  at room temperature
- $\Delta T_A = 10^{\circ} C$
- change base drive with temperature to keep power disspation constant



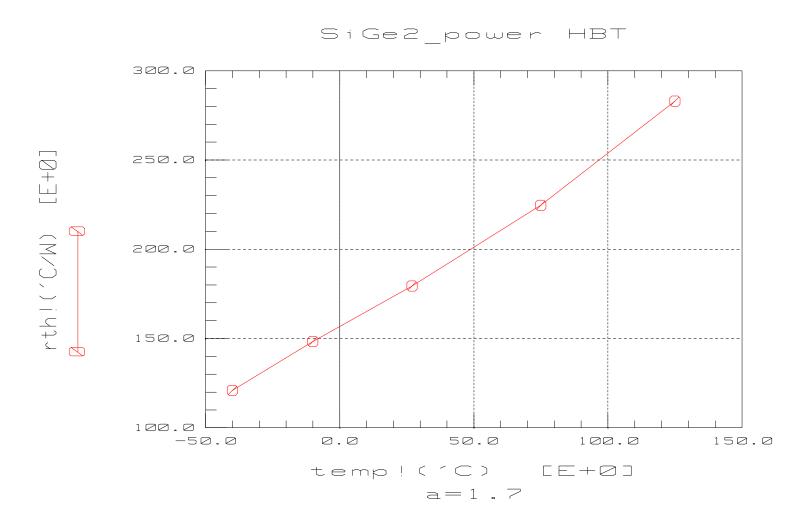


Figure 3: extracted dependence of RTH on ambient temperature





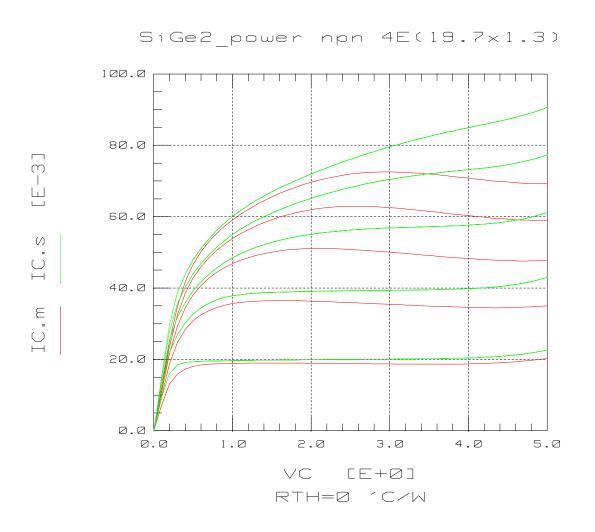


Figure 4: output characteristics of HBT w/o. selfheating



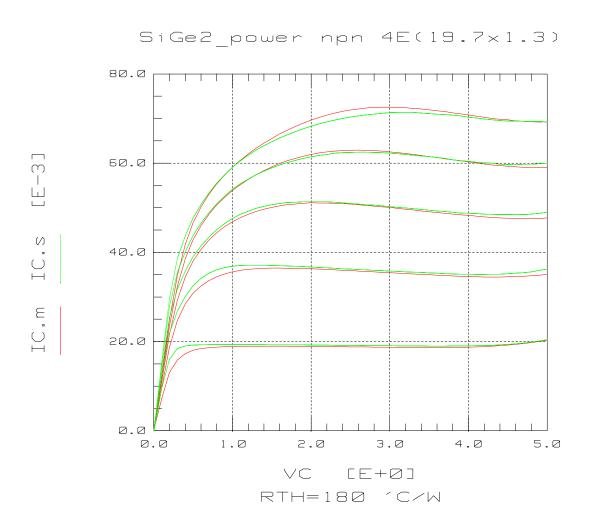


Figure 5: output characteristics of HBT including selfheating



# **Conclusions**

- Most recent method for R\_TH extraction has been verified
- Thermal resistance depends on temperature
- Therefore selfheating (SH) becomes nonlinear
- Can be handled by Kirchhoff-Transformation [7]
- Models having SH should take that into account





### References

- [1] L.T. Su et.al., T-ED, Jan. 1994
- [2] H. Tran et.al., BCTM, 1997
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- [4] D.T. Zweidinger et.al., BCTM, 1995
- [5] J.C.T. Passchens et.al., BCTM, 2004
- [6] M. Reisch , Solid-State Electronics, pp. 677-679, 1992
- [7] K. Poulton et.al., JSSC, Oct 1992