

Probe Tip Characterization Using Self-Calibration

- equipment
- how to determine probe S-parameter and parasitics (probe specific calibration coefficients) for different probes with only one impedance standard substrate (ISS)
- GSG and GS results
- verification
- conclusions



port1 / reference plane 1
e.g. coaxial 2.92 mm connector

port2 / reference plane 2
coplanar (GSG) or
dual stripline (GS)

problem:determine TwoPort S-Parameter from reference plane 1 to reference plane 2 including "contact parasitics"

available probes at FH

Cascade Microtech
- ACP GSG 150 μm

- ACP GS/SG 150 μm
- HPC GSG 100 μm

Picoprobe

- GGB 40A 150 μm

Suss

- |Z|-Probe 40 N 125 μm

} 5 different
ISS are
necessary
for calibration

measurement equipment

NWA Agilent 8720 D

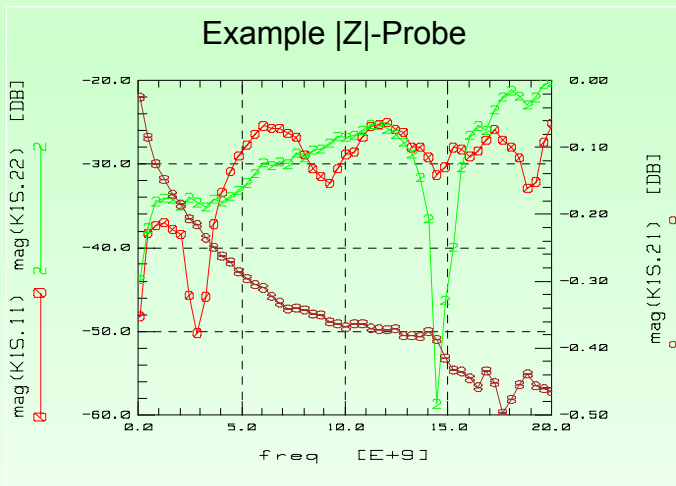
cables Huber+Suhner, 3.5 mm

Agilent 3.5 mm calibration kit (85052D)

Impedance Standard Substrate (ISS)

Cascade 101-190 B

- perform 3.5 mm one port SOL-calibration at port 1
- measure at the calibrated port 1 the on wafer load, open and short Standards connected at port 2
- calculate the probes S-parameters using the shown ICCAP Transform "K1S" (SOL One Port Calibration)



```

UPDATE_MANUAL
n=size(freq)
complex x.22[n]
GL=LO/SOLT.m.11
GO=OL/SOLT.m.11
GS=SL/SOLT.m.11
RL=sc101_190B/l1/S11
RO=sc101_190B/o1/S11
RS=sc101_190B/s1/S11
N12=(GL-GO) / (RL-RO)
N13=(GL-GS) / (RL-RS)
i=0
while i < n
    x.22[i]=(N12[i]-N13[i]) / (RO[i]*N12[i]-RS[i]*N13[i])
    x.12[i]=sqrt(N12[i]*(1-x.22[i]*RL[i])*(1-x.22[i]*RO[i]))
    x.21[i]=x.12[i]
    x.11[i]=GL[i]-RL[i]*x.21[i]**2 / (1-x.22[i]*RL[i])
    i=i+1
end while
return x
    
```

!measured reflection coefficient of the load
 !measured reflection coefficient of the open
 !measured reflection coefficient of the short
 !known reflection coefficient of the load
 !known reflection coefficient of the open
 !known reflection coefficient of the short
 !Source Match Error
 !Source Tracking Error
 !Directivity Error

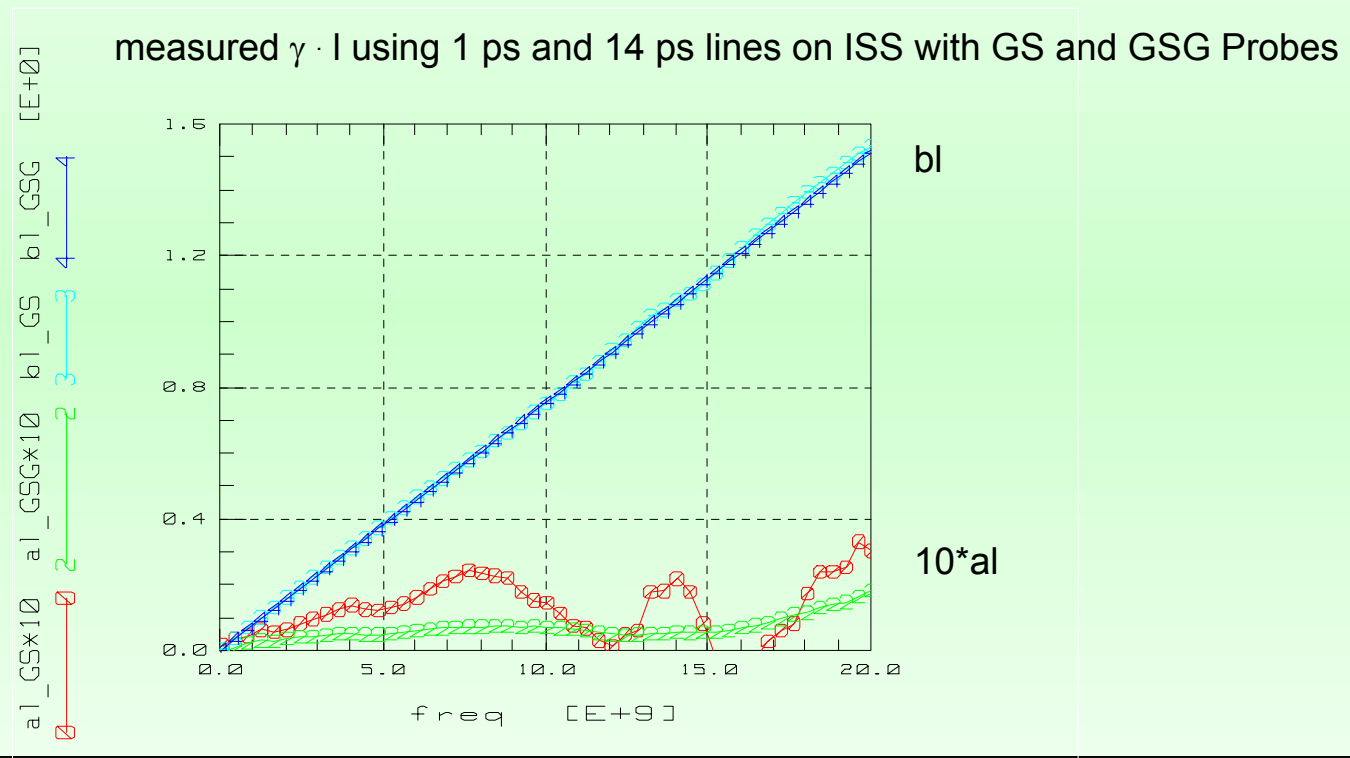
problem1: "in on-wafer probing the electrical behavior of the standards are dependent upon the probe and how it is placed" [1]

problem 2: probe manufacturers deliver probe specific standard definitions only for their own ISS

Standard	Known Parameter	Unknown Parameter-determined by self-calibration
short line ("Thru")	length	line propagation constant
long line ("Line")	length	
on wafer load	DC resistance / Ω	skin effect loss, load inductance, characteristic line impedance
on wafer open ("Reflect")	-	reflection coefficient; open capacitance
on wafer short	-	reflection coefficient; short inductance

result: calculation of the 12 TERM error coefficients

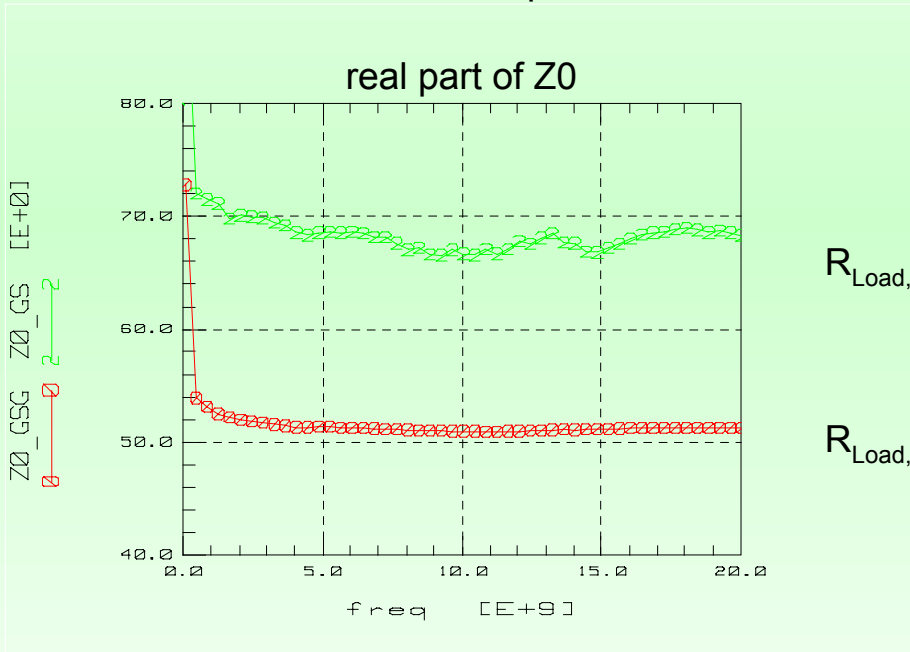
$$\gamma \cdot l = (\alpha + j \cdot \beta) \cdot l = a_l + j \cdot b_l$$



- measure uncalibrated S-parameter of the load and calculate the reflection coefficient Γ_{Load} using the TRL solution
- measure DC resistance of the load $R_{Load,dc}$
- calculate the line capacitance
- calculate the characteristic impedance

$$C \approx \frac{\gamma \cdot l}{j \cdot \omega \cdot R_{Load,dc}} \cdot \frac{1 + \Gamma_{Load}}{1 - \Gamma_{Load}}$$

$$Z0 \approx \frac{\gamma \cdot l}{j \cdot \omega \cdot C}$$



$$R_{Load,dc,GS} = 100 \Omega$$

$$R_{Load,dc,GSG} = 50 \Omega$$

- measure uncalibrated S-parameter of the load and calculate the reflection coefficient Γ_{Load} using the TRL solution and Z0

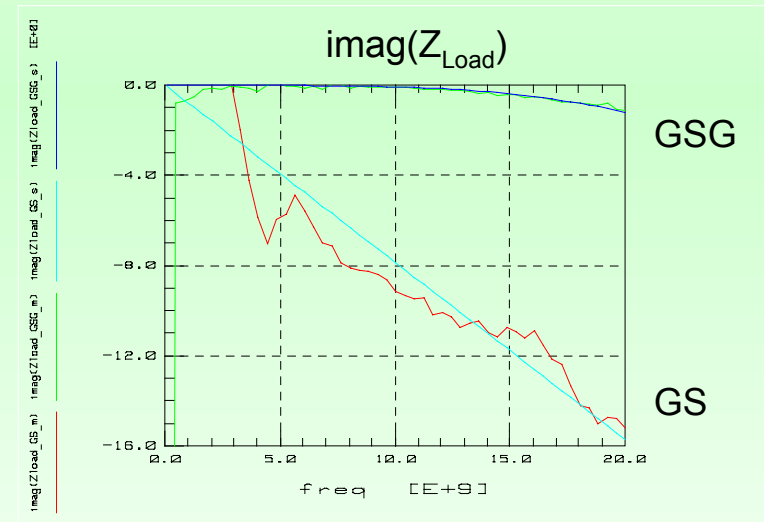
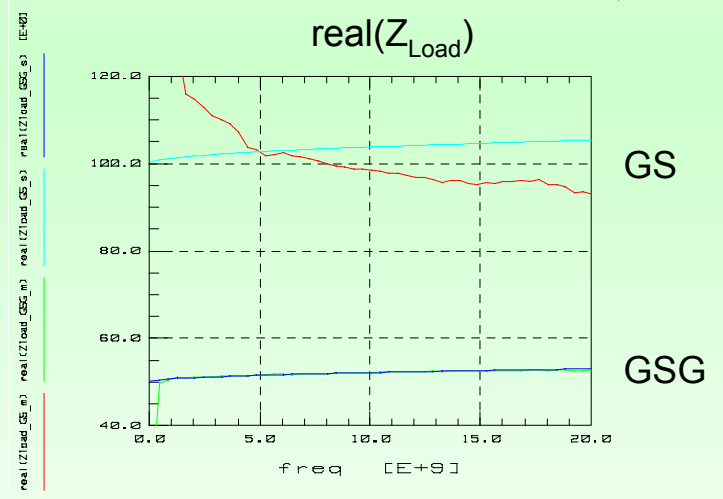
$$Z_{Load} = Z0 \cdot \frac{1 + \Gamma_{Load}}{1 - \Gamma_{Load}} = R \cdot \left(1 + \alpha \cdot \left(\frac{freq}{1GHz} \right)^{0.5} \right) + j \cdot \omega \cdot \left(\frac{L_0}{10^{-12}} + \frac{L_1 \cdot freq}{10^{-24}} + \frac{L_2 \cdot freq^2}{10^{-33}} + \frac{L_3 \cdot freq^3}{10^{-42}} \right)$$

model parameter GSG: $R=50 \Omega, \alpha=0.012, L_0=0 \text{ H}, L_1=0 \text{ H/Hz}, L_2=0 \text{ H/Hz}^2, L_3=-1.2 \text{ H/Hz}^3$

model parameter GS: $R=100 \Omega, \alpha=0.012, L_0=-125 \text{ H}, L_1=0 \text{ H/Hz}, L_2=0 \text{ H/Hz}^2, L_3=0 \text{ H/Hz}^3$

Cascade ACP GS $R=50 \Omega, L=57.8 \text{ pH}$

Cascade HPC GSG $R=50 \Omega, L=-3.5 \text{ pH}$



- measure uncalibrated S-parameter of the open and calculate the reflection coefficient Γ_{Open} using the TRL solution and Z_0

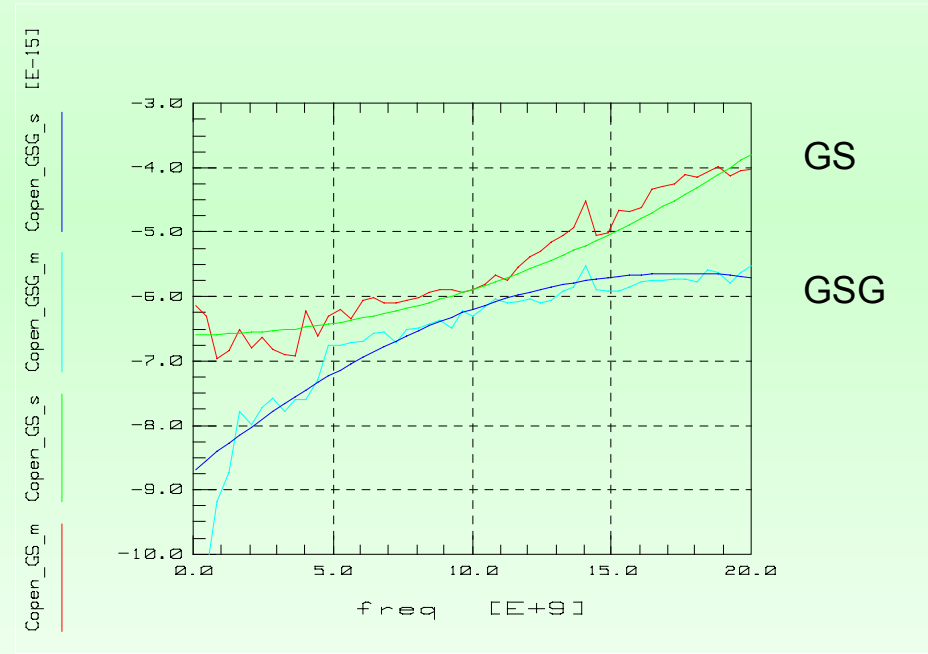
$$Y_{Open} = \frac{1}{Z_0} \cdot \frac{1 - \Gamma_{Open}}{1 + \Gamma_{Open}} = j \cdot \omega \cdot \left(\frac{C_0}{10^{-15}} + \frac{C_1 \cdot freq}{10^{-27}} + \frac{C_2 \cdot freq^2}{10^{-36}} + \frac{C_3 \cdot freq^3}{10^{-45}} \right)$$

model parameter GSG: $C_0 = -8.7 \text{ F}, C_1 = 350 \text{ F/Hz},$
 $C_2 = -10 \text{ F/Hz}^2, C_3 = 0 \text{ F/Hz}^3$

model parameter GS: $C_0 = -7 \text{ F}, C_1 = 0 \text{ F/Hz},$
 $C_2 = 7 \text{ F/Hz}^2, C_3 = 0 \text{ F/Hz}^3$

Cascade ACP GS $C = -11 \text{ fF}$

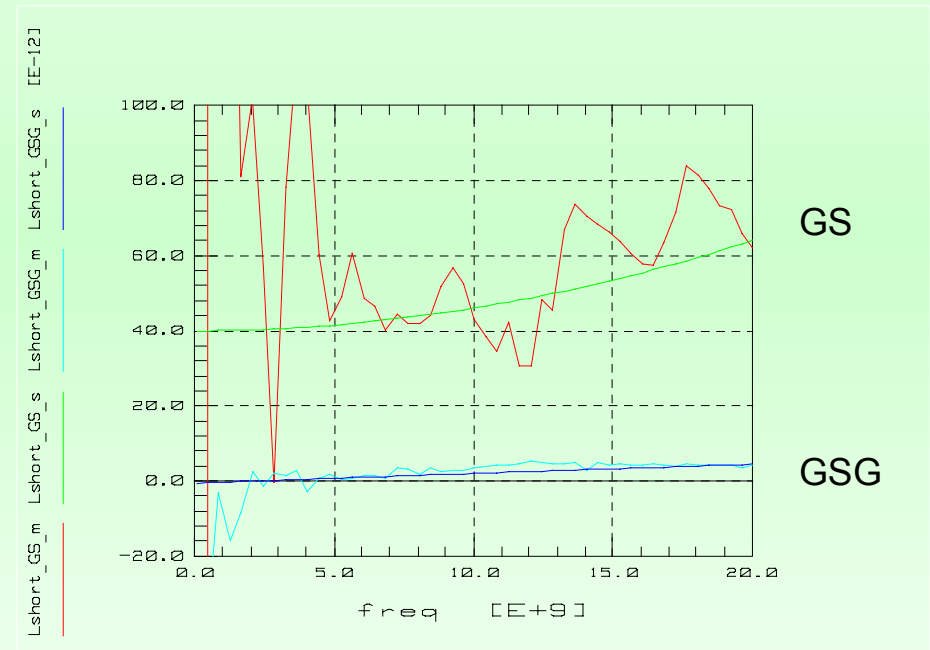
Cascade HPC GSG $C = -9.3 \text{ fF}$

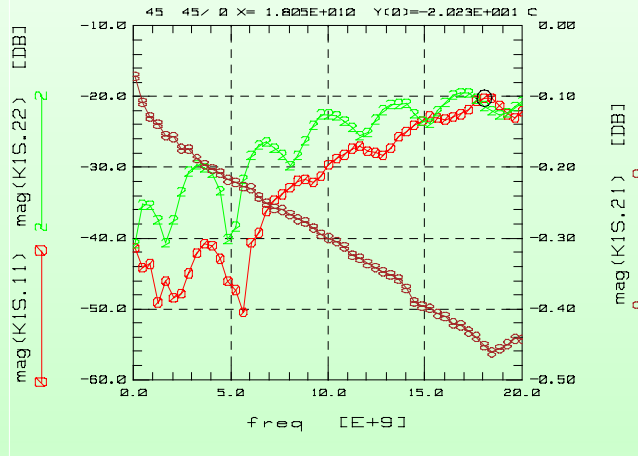


- measure uncalibrated S-parameter of the open and calculate the reflection coefficient Γ_{Short} using the TRL solution and Z_0

$$Z_{Short} = Z_0 \cdot \frac{1 + \Gamma_{Short}}{1 - \Gamma_{Short}} = j \cdot \omega \cdot \left(\frac{L_0}{10^{-12}} + \frac{L_1 \cdot freq}{10^{-24}} + \frac{L_2 \cdot freq^2}{10^{-33}} + \frac{L_3 \cdot freq^3}{10^{-42}} \right)$$

- model parameter GSG: $L_0 = -0.5 \text{ H}, L_1 = 250 \text{ H/Hz},$
 $L_2 = 0 \text{ H/Hz}^2, L_3 = 0 \text{ H/Hz}^3$
- model parameter GS: $L_0 = 40 \text{ H}, L_1 = 0 \text{ H/Hz},$
 $L_2 = 60 \text{ H/Hz}^2, L_3 = 0 \text{ H/Hz}^3$
- Cascade ACP GS $L = 49.8 \text{ pH}$
- Cascade HPC GSG $L = 2.4 \text{ pH}$





Kalibriersubstrat C101-190B
Pitch 150 μm
SN:434907



Messung vom 30.07.2008

data provided by Cascade:

Copen = -9.7 fF

Lshort = 4.8 pH

Lload = 0.2 pH

delay = 1 ps

$Copen_0 = -11.5\text{fF}$

$Copen_1 = 267 \cdot 10^{-27} \text{ F/Hz}$

$Copen_2 = -6.5 \cdot 10^{-36} \text{ F/Hz}^2$

$Lshort_0 = 7 \text{ pH}$

$Lshort_1 = 350 \cdot 10^{-24} \text{ H/Hz}$

$Rload = 50.0\Omega$

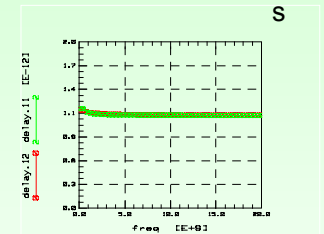
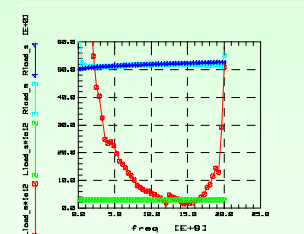
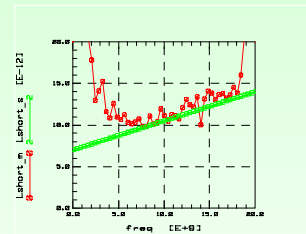
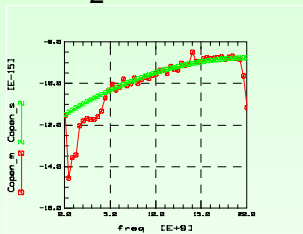
$\alpha = 0.012$

$Lload_0 = 3 \text{ pH}$

Line Z0 = 50 Ω

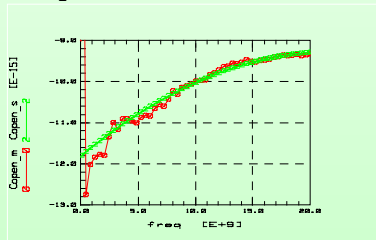
delay = 1.105ps

loss@1GHz = 29 $\frac{\text{G}\Omega}{\text{s}}$



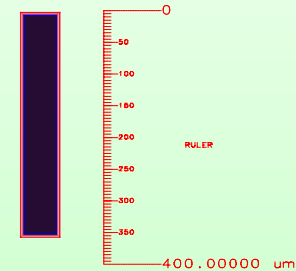
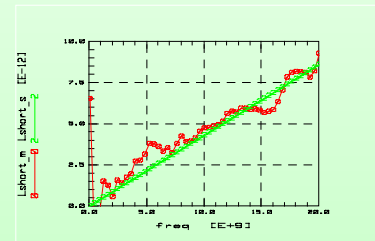
ACP40 GSG Pitch 150 μm SN:434907 Messung vom 30.07.2008

$Copen_0 = -11.5 \text{ fF}$
 $Copen_1 = 267 \cdot 10^{-27} \text{ F / Hz}$
 $Copen_2 = -6.5 \cdot 10^{-36} \text{ F / Hz}^2$

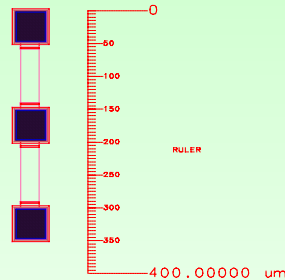
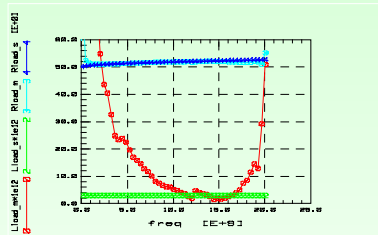


probe in air

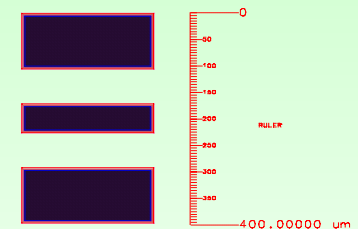
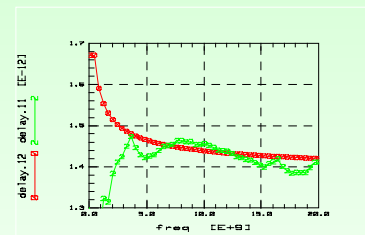
$Lshort_0 = 0 \text{ pH}$
 $Lshort_1 = 430 \cdot 10^{-24} \text{ H / Hz}$



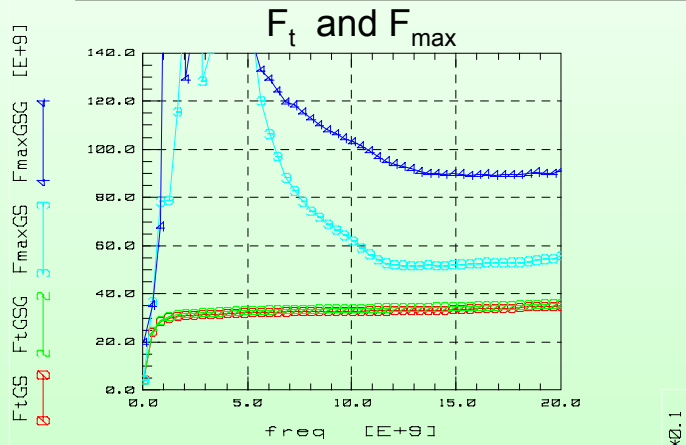
$Rload = 50.55 \Omega$
 $alpha = 0.0115$
 $Lload_0 = 0 \text{ pH}$
 $Lload_1 = -300 \cdot 10^{-24} \text{ H / Hz}$



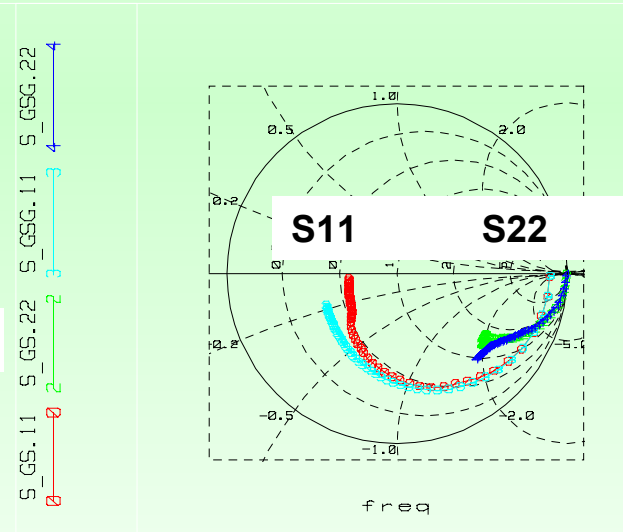
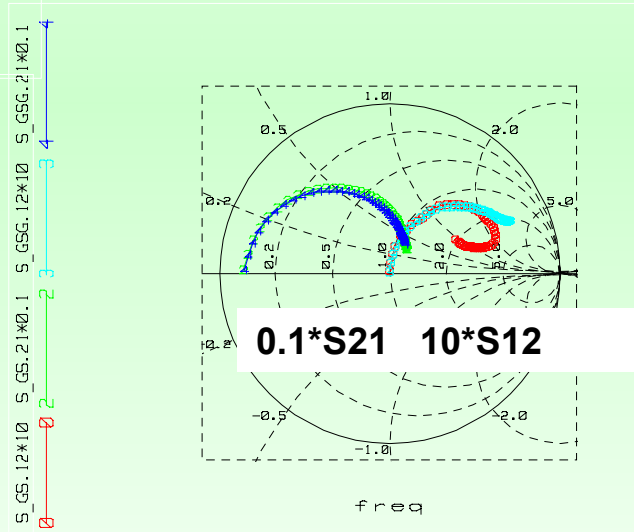
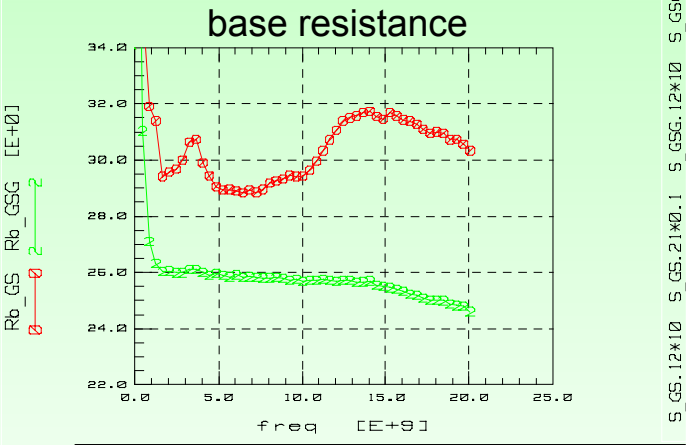
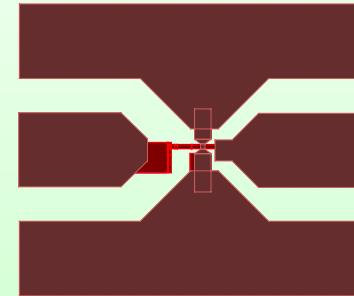
$Line1 \ Z0 = 61 \Omega$
 $delay = 1.35 \text{ ps}$
 $loss@1GHz = 110 \frac{G\Omega}{s}$

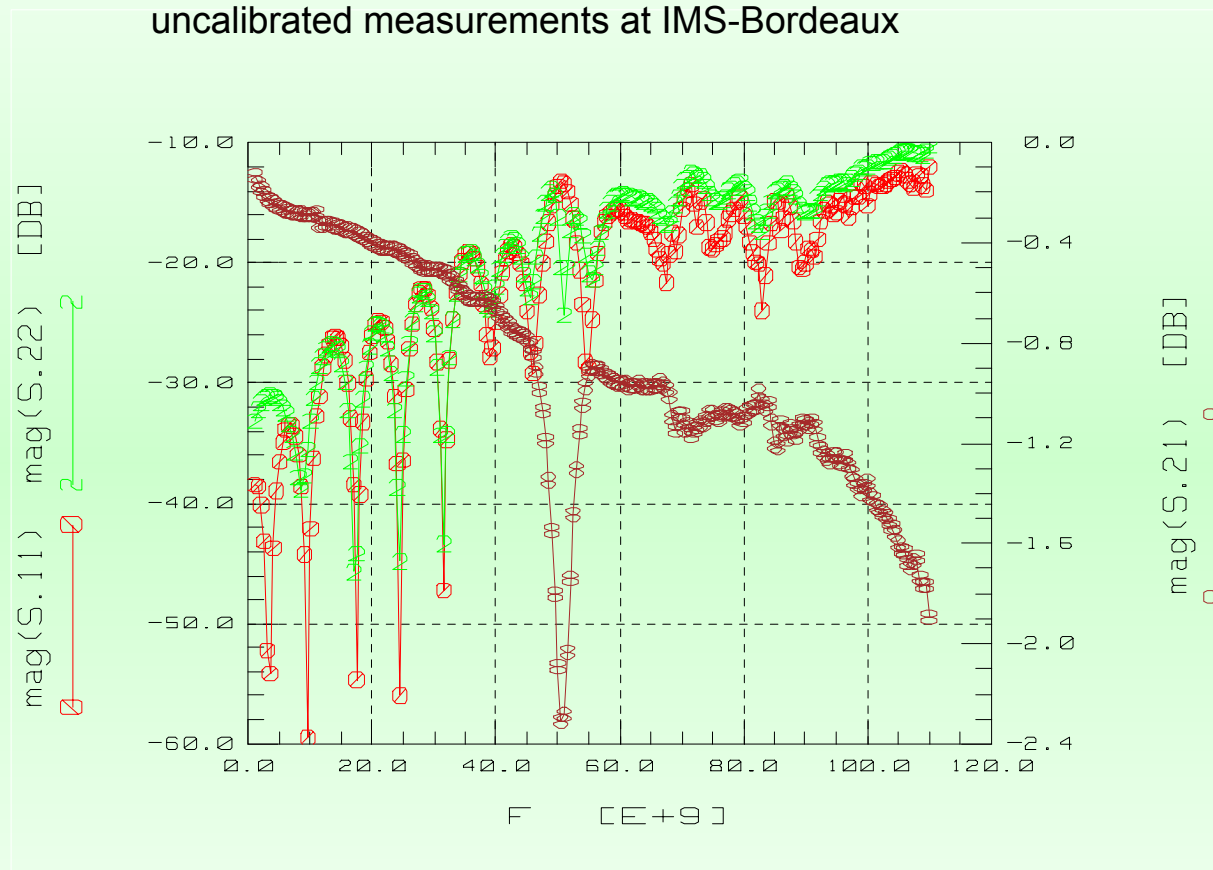


Probe	Pitch/ μm	max(S11)/dB	max(S22)/dB	min(S21)/dB
Suss Z-Probe 40 N GSG Ni	125	-25	-20.5	-0.5
GGB EDPW 40 GSG W	150	-25	-22	-0.65
ACP40 GSG W	150	-18	-20	-0.5
ACP40 GSG W	150	-21	-26	-0.38
HPC40 GSG BeCu	100	-24	-20	-0.3
HPC40 GSG BeCu	100	-26	-26	-0.34
ACP40 GS W	150	-19	-13	-0.76
ACP40 SG W	150	-18	-14	-0.76
Suss Z-Probe GSG Ni 1mm	100	-25	-25	-0.39



GaAs HBT $2 \times 10 \mu\text{m}^2$
 $V_{ce} = 3 \text{ V}; I_c = 10 \text{ mA}$
 Frequency 0.05...20.05 GHz





a method for self calibration of on wafer probe parasitics is shown using TRL

this is necessary if a ISS with unknown standard definitions is used (e.g. 101-190B with GS probes)

known standard parameters are length of two transmission lines with equal characteristic line impedance and dc resistance of a load

the method can be further improved using NIST multiline TRL algorithm

with the TRL solution the S-parameters of short, open, load and thru standards are measured and modeled using the parameters available in e.g. 8510 NWA to achieve a broadband solution without ripple in SOLT/SOLR calibrations

the load model is extended by inclusion of the skin effect

it is possible to characterize a self built on wafer calibration kit

using the probe tip contact parasitics for the available calibration standards the probes S-parameter are calculated with the SOL algorithm

GS and GSG measurements deliver different results for transistor modeling (e.g. F_{max} and R_b)

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- [1] Cascade Microtech Inc. Impedance Standard Substrate description of P/N 101-190
 - [2] G.F.Engen, C.A. Hoer, "*Thru-Reflect-Line*": *An Improved Technique for Calibrating the Dual Six-Port Automatic Network Analyzer*, MTT-27, No12, Dec 1979, pp.987-993.
 - [3] D.F. Williams and R.B. Marks, *Transmission Line Capacitance Measurement*, IEEE MICROWAVE AND GUIDED WAVE LETTERS, VOL 1, NO. 9, SEP 1991,pp.243-245.

Thank You for Your attention