



Limitations of Bipolar Compact Models for LF Noise PART II – Answers to HICUM WS 2010 remarks

23rd ArbeitsKreis Bipolar (AKB) – Crolles – October 15th 2010

Nicolas DERRIER

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Outline

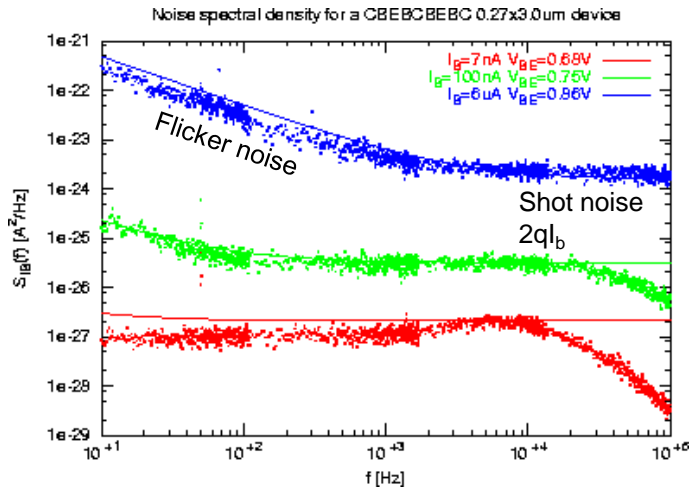
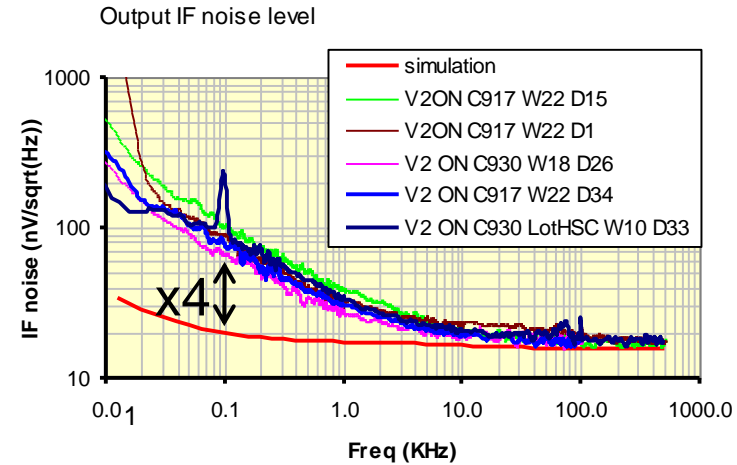
- Summary of the study presented at the HICUM WS 2010
- Remarks and questions opened at the HICUM WS
- Generalities and theory about LF noise measurements setup
- Detailed conditions and setup of our measurements and simulations
- Definitive conclusions

HICUM WS 2010 summary (1/4)

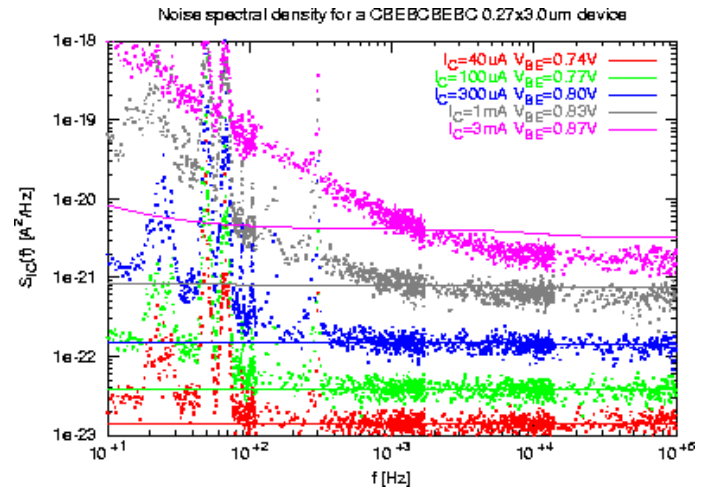


In [Derrier10], we showed that :

- Issue of 1/f Spice simulations at very low frequencies for an application designed with ST BiCMOS 0.13 μ m mmW technology
- Issue coming from the NPN High-Speed SiGe-C simulations
- 1/f simulations vs measurements OK on the Base
- 1/f simulations vs measurements NOT OK on the Collector

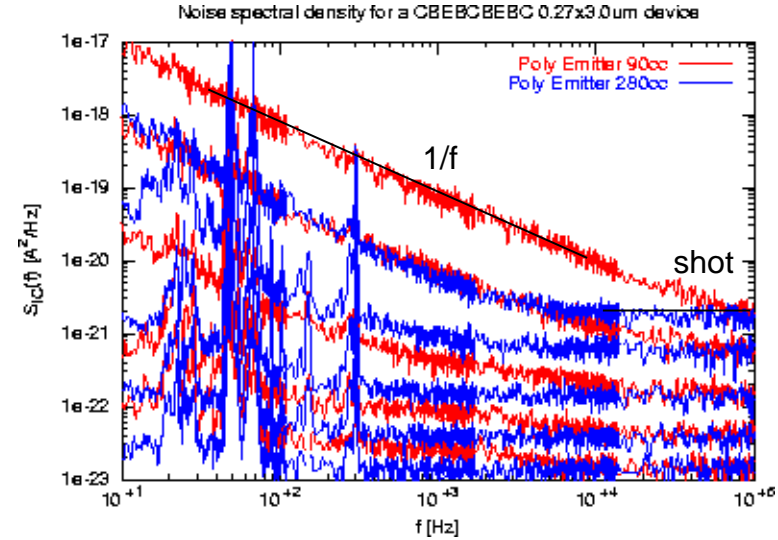
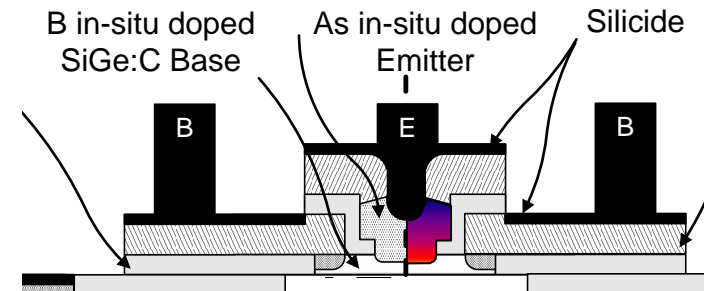


SI on the Base



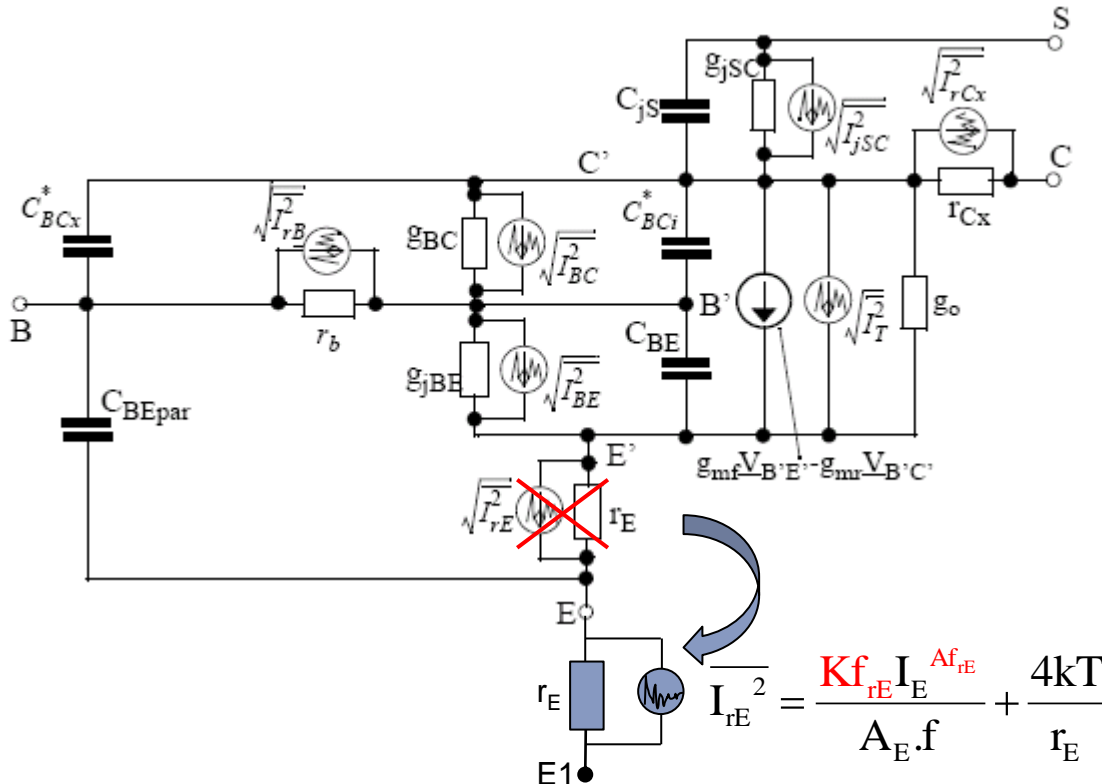
SI on the Collector

- Only 1 $1/f$ noise source in compact models today : $S_{I_{EB}}$
- Other $1/f$ noise sources in the HBT, not taken in account in Spice ?
- A Process DOE shows that the missing $1/f$ noise main contribution comes from the Emitter poly/mono interface (R_E)

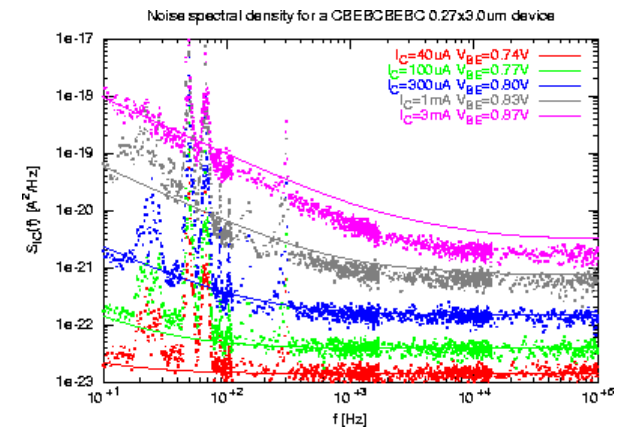
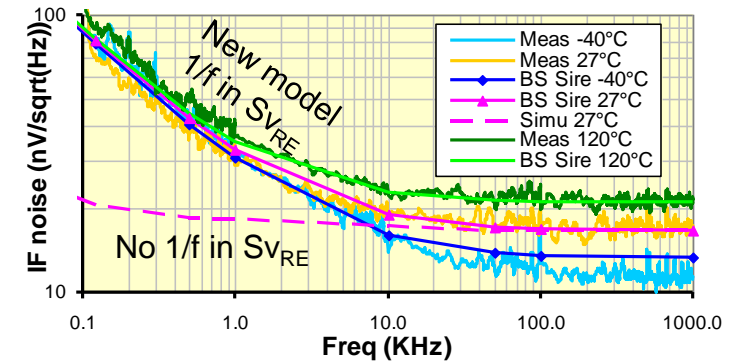


HICUM WS 2010 summary (3/4)

- Many papers showing also the limitations of actual compact models for 1/f noise, especially for series resistance
- [Ziel86&87], [Haaren98], [Tartarin99], [Borgarino99], [Kirtania96], [Kleinpenning92&94&95], [Benoit05], [Nunez-Perez07]
- Integrating an additional 1/f source on R_E works very fine

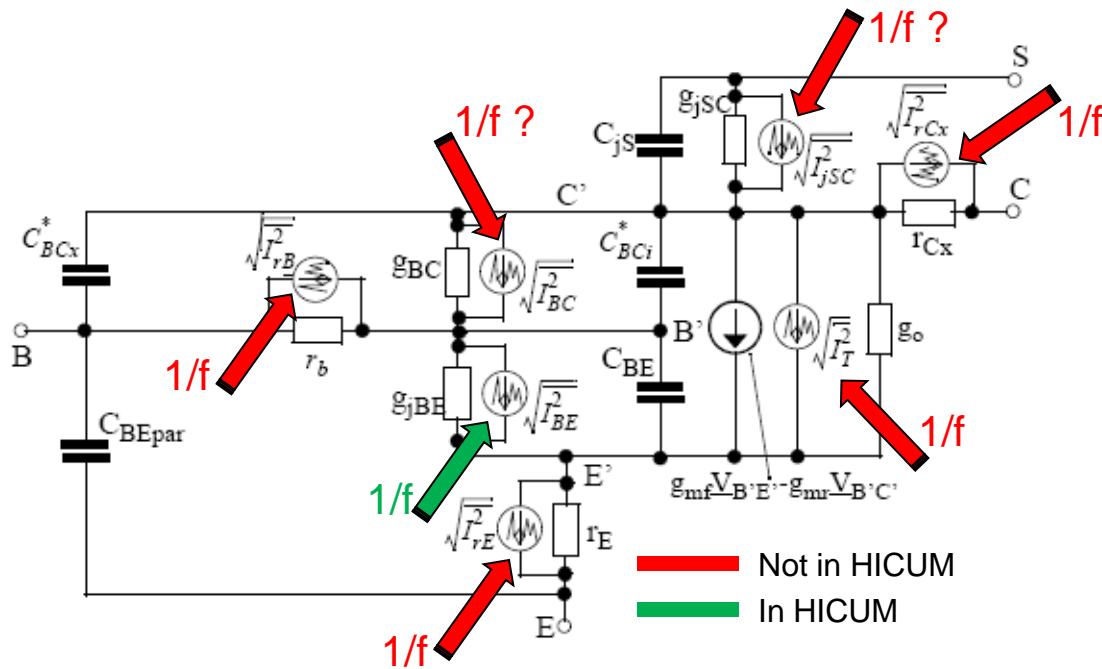


Output noise level LotHSC W10 D27 / LO&RF OFF



HICUM WS 2010 summary (4/4)

- Yes, but only a workaround
- Capacitances partitioning changed, RF model no more valid
- TUD to improve HICUM 1/f noise model, with new parameters ?



$$\overline{I_{r_B}^2} = \frac{4kT}{r_B} + \frac{k_{F_{r_B}} I_B^{a_{r_B}}}{f}$$

$$\overline{I_{r_E}^2} = \frac{4kT}{r_E} + \frac{k_{F_{r_E}} I_E^{a_{r_E}}}{f}$$

$$\overline{I_{r_{C_x}}^2} = \frac{4kT}{r_{C_x}} + \frac{k_{F_{r_{C_x}}} I_C^{a_{r_{C_x}}}}{f}$$

$$\overline{I_T^2} = 2qI_T + \frac{k_{F_{CE}} I_T^{a_{F_{CE}}}}{f}$$

$$\overline{I_{BC}^2} = 2q I_{jBC} + I_{AVL} + \frac{k_{F_{BC}} I_{jBC} + I_{AVL}^{a_{F_{BC}}}}{f}$$

$$\overline{I_{CS}^2} = 2qI_{jCS} + \frac{k_{F_{CS}} I_{jCS}^{a_{F_{CS}}}}{f}$$

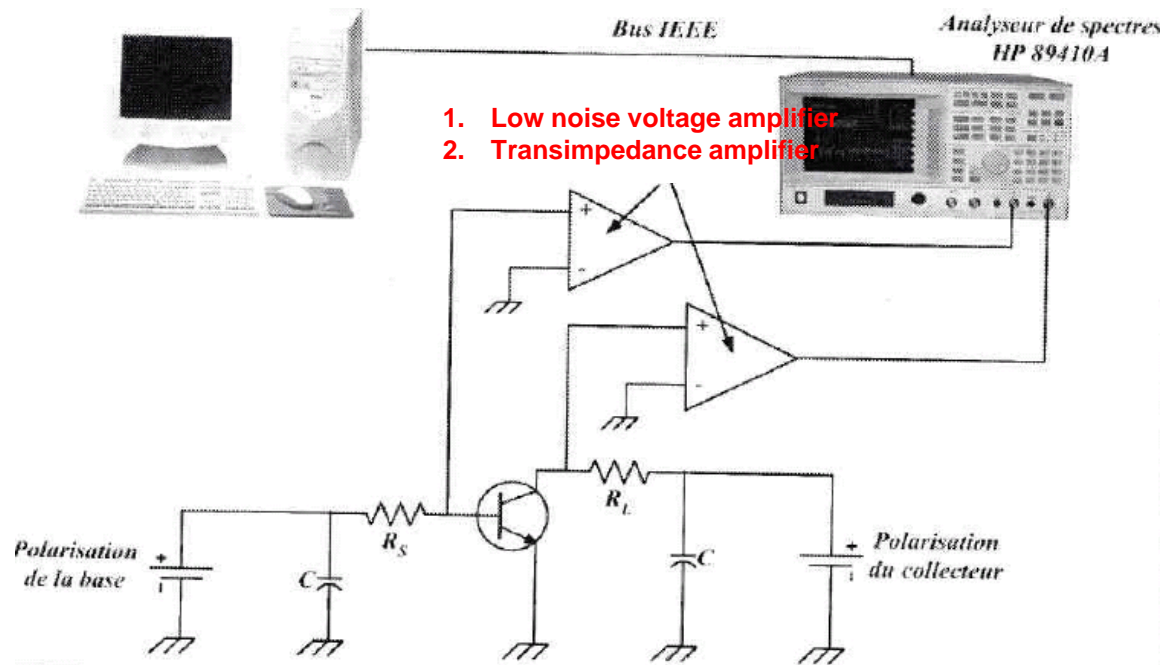
$$\overline{I_{BE}^2} = 2qI_{jBE} + \frac{k_{F_{BE}} I_{jBE}^{a_{F_{BE}}}}{f} \quad (\text{in HL2, } I_{jBE} = I_{jBEi} + I_{jBEp})$$

HICUM WS remarks or questions

- Some people skeptic in front of our results and conclusions
- However, many papers exist showing the same kind of conclusions
- Some questions or remarks about :
 - “What exact setup for measurements and simulations ?”
 - “Are you sure simulations have the same schematic as measurements ?”
 - “Influence of the source resistance ?”

Characterization setup – theory (1/3)

- 2 possible configurations for low frequency noise measurements (common emitter case) :
 1. Use low noise voltage amplifiers : ok for large currents ($I_B > 500\text{nA}$)
 2. Use transimpedance amplifier : allows small currents too
- ST uses configuration 2. to measure all currents values



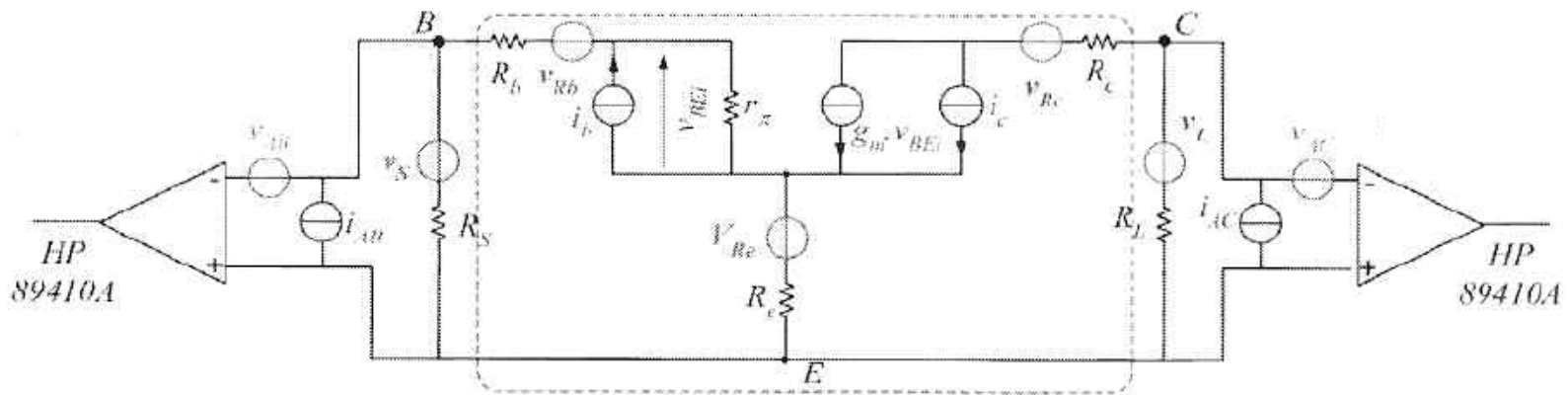
Low noise measurements setup

Characterization setup – theory (2/3)

- The values of the voltage spectral density on the Base and on the Collector are the following [Benoit05] :

$$S_{V_B} = \frac{R_S^2}{(R_S + R_d)^2} \left[4kT \frac{R_d^2}{R_S} + S_{V_{RE}} + S_{V_{RB}} + (r_\pi + R_E h_{fe})^2 S_{i_B} + R_E^2 S_{i_C} \right. \\ \left. + 2R_E (r_\pi + R_E h_{fe}) PR(S_{i_B} i_C) \right]$$

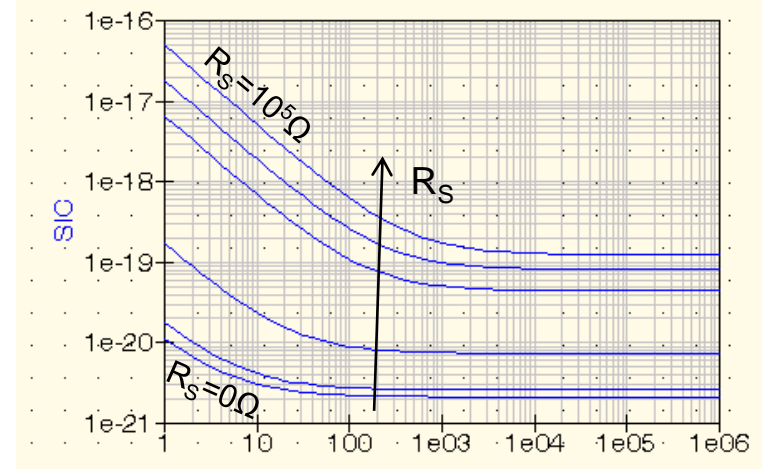
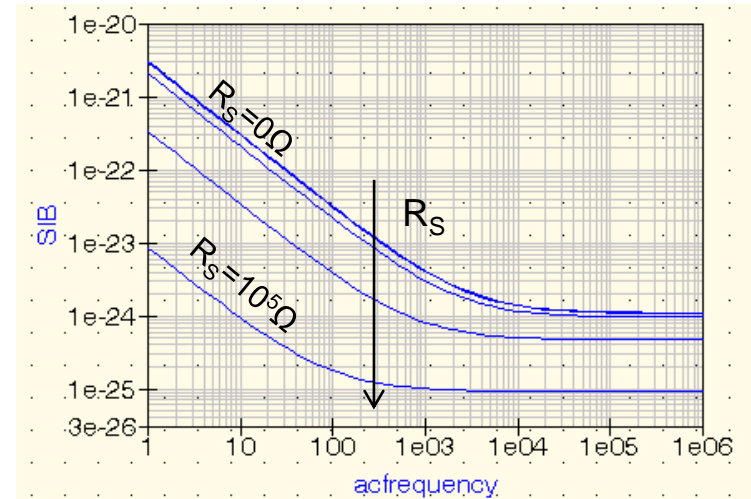
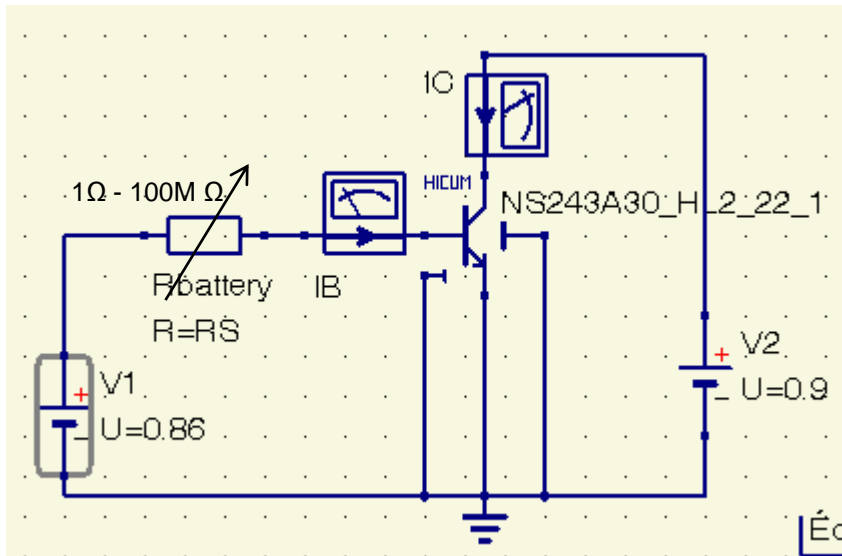
$$S_{V_C} = 4kTR_L + \frac{h_{fe}^2 R_L^2}{(R_S + R_d)^2} \left[4kTR_S + S_{V_{RE}} + S_{V_{RB}} + (R_S + R_B + R_E)^2 S_{i_B} + \frac{(R_S + R_B + R_E + r_\pi)^2}{h_{fe}^2} S_{i_C} \right. \\ \left. + \frac{2(R_S + R_B + R_E)(R_S + R_B + R_E + r_\pi)}{h_{fe}} PR(S_{i_B} i_C) \right]$$



Small signal scheme of the measurements chain

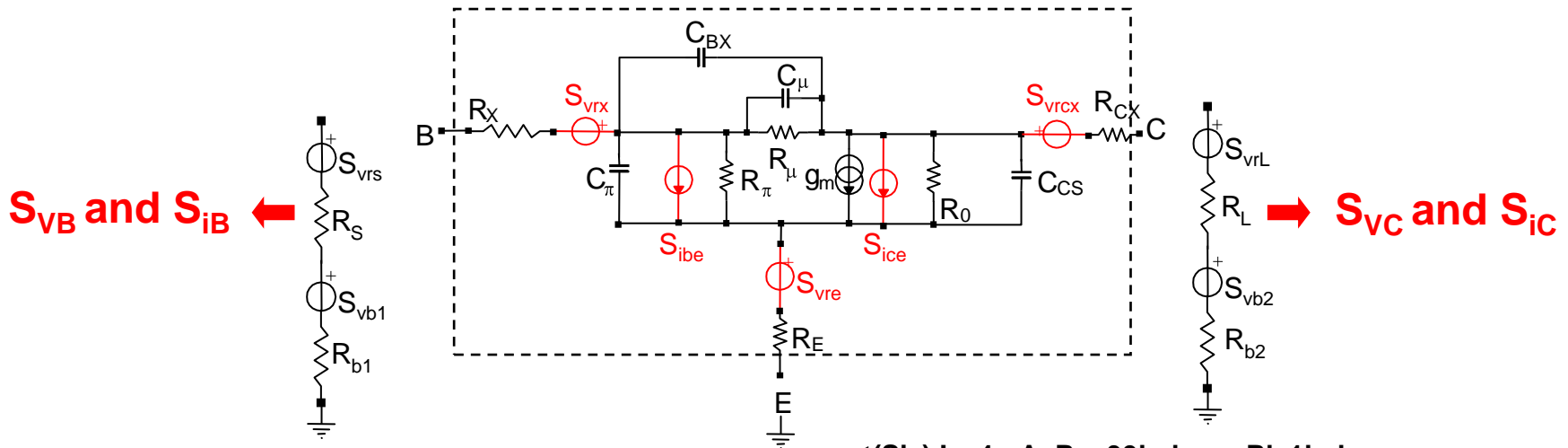
Characterization setup – theory (3/3)

- The source resistance R_S is very important and plays on the spectral density on the Base and on the Collector
- 2 cases are often considered :
 - High impedance setup ($R_S \gg R_d$)
 - Null impedance setup ($R_S=0\Omega$)



Characterization setup – practice (1/4)

- 1st case : high impedance setup, $R_s=39k\Omega$
- Simulations vs measurements S_{vC}/S_{iC} OK on the Collector



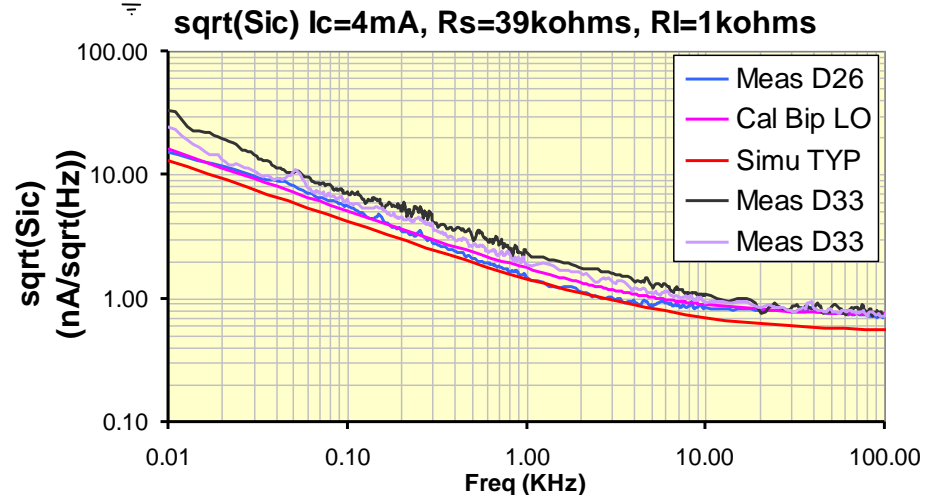
• Conditions :

$R_s=39k\Omega$; $R_L=1k\Omega$

$R_{b1}=149\Omega$; $R_{b2}=407\Omega$

• Measurements :

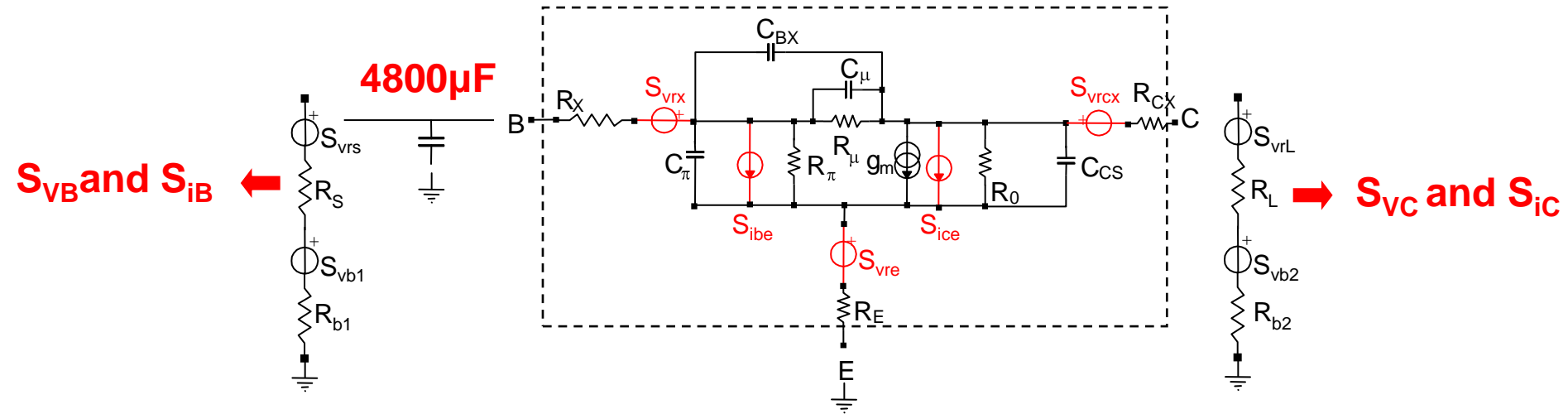
$$S_{ic} \approx \beta_{noise}^2 * S_{ibe} \text{ [Benoit05]}$$



Slide with courtesy of L. Moquillon

Characterization setup – practice (2/4)

- 2nd case : null impedance $R_s=0\Omega$



• Conditions :

$R_s=0\Omega$; $R_L=1k\Omega$

$R_{b1}=149\Omega$; $R_{b2}=407\Omega$

Impedance=0 on the base means coupling capa needed to avoid the battery R_{b1}

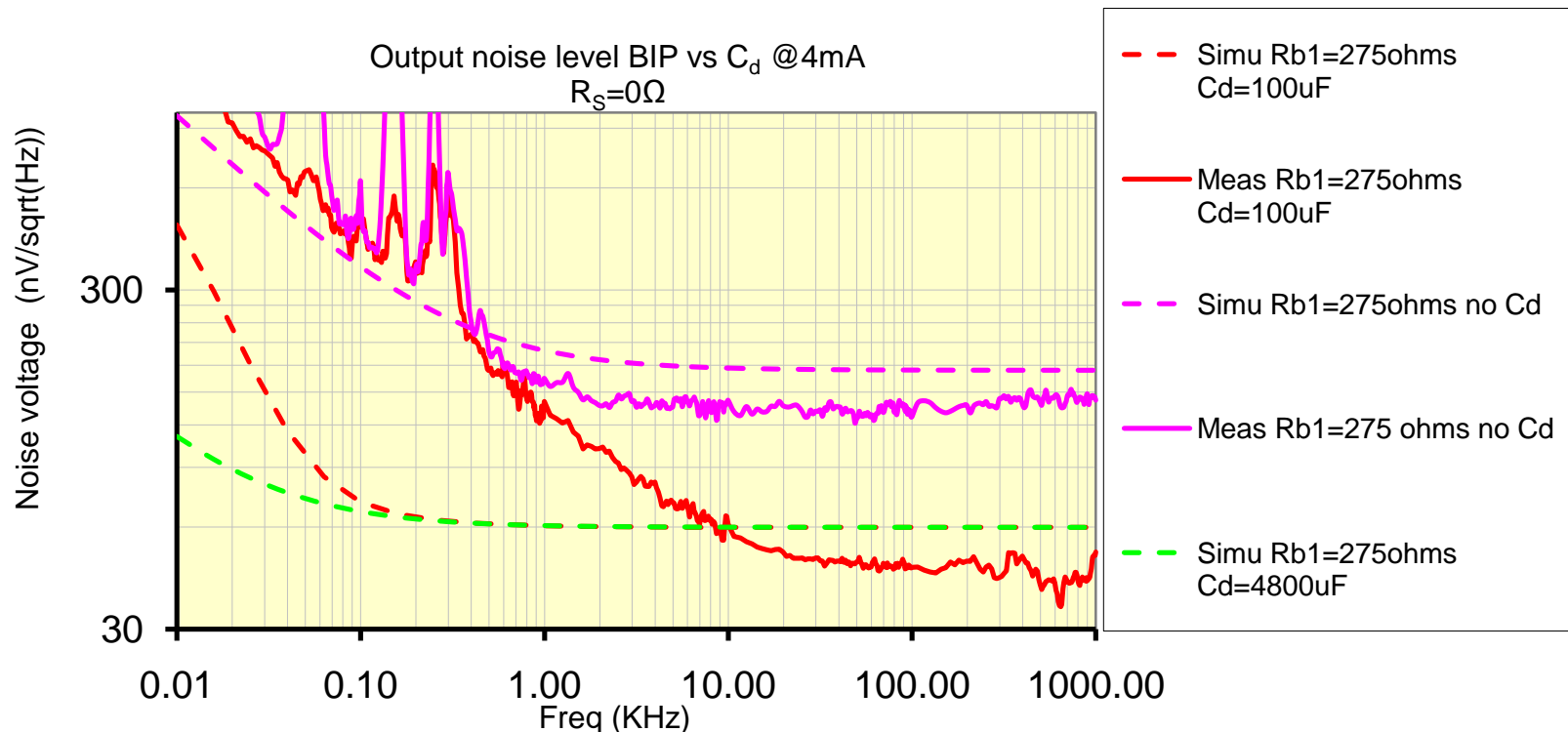
$C=4800\mu F \rightarrow f_c=0.22Hz$

• Measurements : $S_{vc} \approx R_L \left(S_{ibe} \beta_{noise} + \frac{1}{(1 + g_m R_E)^2} (S_{ice} + g_m^2 (S_{vre} + S_{vrb})) \right)$ [Benoit05]

β_{noise} small

Characterization setup – practice (3/4)

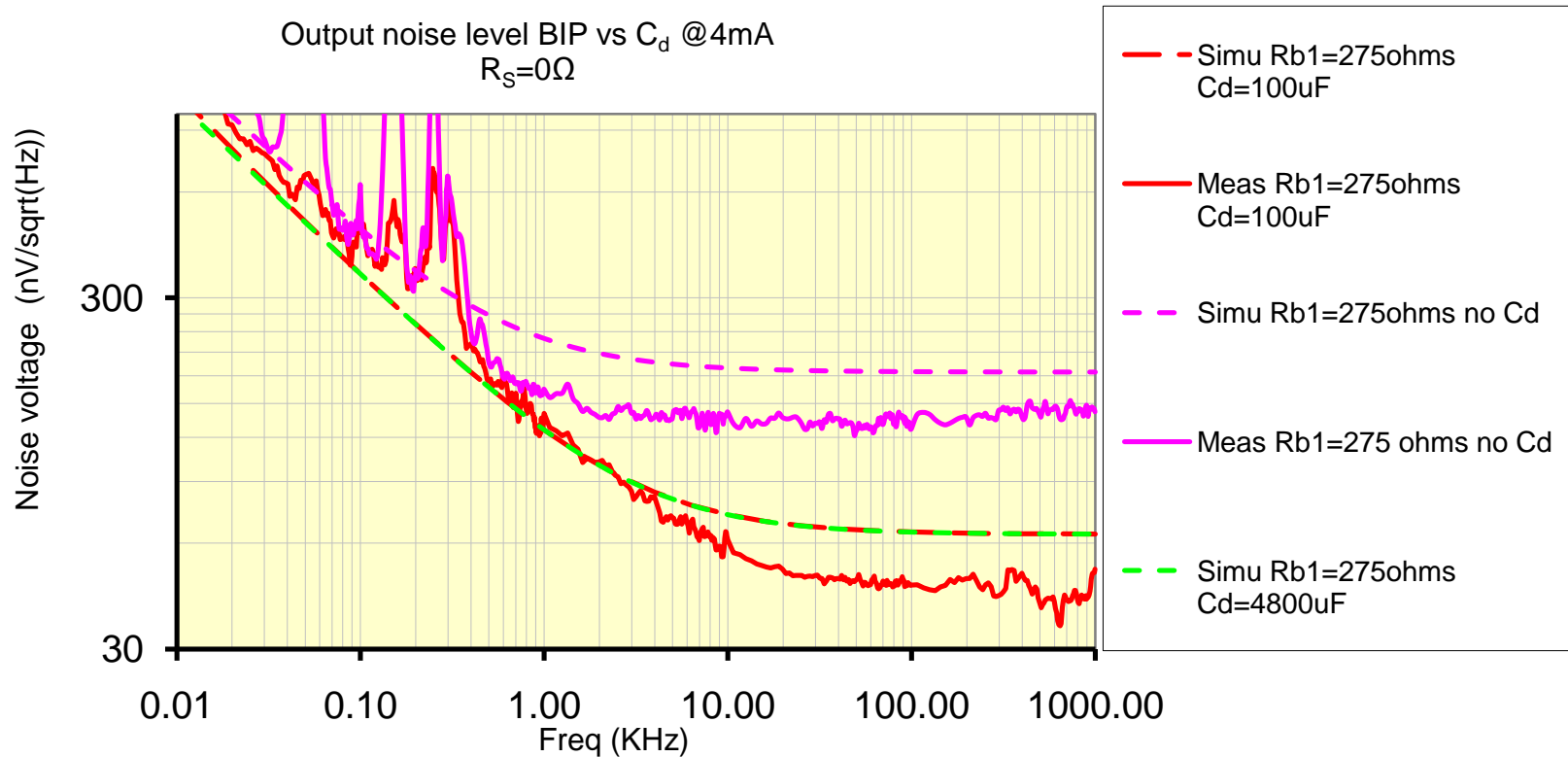
- Results for $R_s=0\Omega$
 - There is still the resistance coming from the battery on the Base R_{b1}
 - A decoupling capacitance C_d is needed to filter R_{b1}
 - C_d has no influence on measurements $< 1\text{kHz}$!!!
 - Confirms the existence of preponderant $1/f$ noise source other than SI_{EB}



Slide with courtesy of L. Moquillon

Characterization setup – practice (4/4)

- Results for $R_s=0\Omega$
 - The same simulations with our additional $1/f$ noise source in subcircuit R_E works fine !



Conclusions

- Summarized the HICUM WS 2010 presentation about ST 1/f noise issue in compact models
- Shown that our LF noise simulations and measurements on HBT are consistent

- R_S influence and variation taken in account in both measurements and simulations
- Investigated both cases $R_S=0\Omega$ and large $R_S (>> R_d)$
 - $R_S >> R_d$ works fine, since LF noise measured on the collector comes mainly for Si_{EB} , which is correctly modeled
 - Our issue comes from the $R_S=0\Omega$ configuration, where we clearly see that another 1/f source than Si_{EB} is visible on the device output

- ST definitive conclusions are the same as in HICUM WS :
 - Other 1/f noise source than Si_{EB} (BE junction) is present in ST BiCMOS 130nm mmw HBT High-Speed
 - Seems to be located in the emitter poly/mono interface (on R_E in small signal scheme)
 - Actual compact models are too much simplified to correctly match 1/f noise simulations vs measurements on the Collector for this advanced HBT
 - Critical issue for some advanced mmw designs and applications
 - Possibility for TUD to add 1/f components in the existing HICUM noise sources ?

Special thanks to...

- Zoltan Huszka (AustriaMS) for his email exchanges on that topic.
- L. Moquillon & P. Garcia (ST) for their design expertise.
- S. Haendler (ST) for his characterizations.

- [**Derrier10**] : Nicolas DERRIER, “Limitations of Bipolar Compact Models for Low Frequency Noise – Application to HICUM”, HICUM Workshop 2010, Dresden
- [**Nunez-Perez07**] : Jose Cruz NUNEZ PEREZ, “Contribution à la Conception de Système de RadioCommunication : de la Modélisation de Transistors Bipolaires à l’Evaluation des Performances d’un Système d’Emission-Reception”, Thèse de Doctorat soutenue le 03 Décembre 2007, INSA Lyon
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