

NQS modelling with HiCuM: What works, what doesn't



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Outline

□ Introduction

- NQS basics
- High frequency Y parameters

□ NQS model implementation inside HICUM

- Vertical NQS implementation

□ Modeling results

- NQS modeling with HICUML2
- Scaling of NQS parameters

□ Conclusion

Introduction: NQS basics

□ NQS basics:

As the operating frequency approaches the cutoff frequency, the transistor can no longer follow external excitations instantaneously. (Non Quasi Static effect)

□ Transient variation of electron concentration inside the base

Switch-off

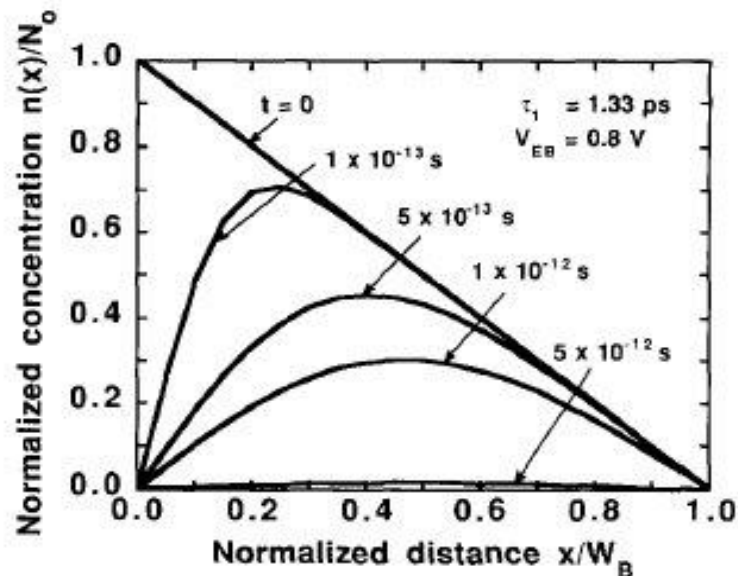


Fig. 5. Transient variation of electron concentration profile when V_{EB} decreases from the switch-on voltage to zero.

Switch-on

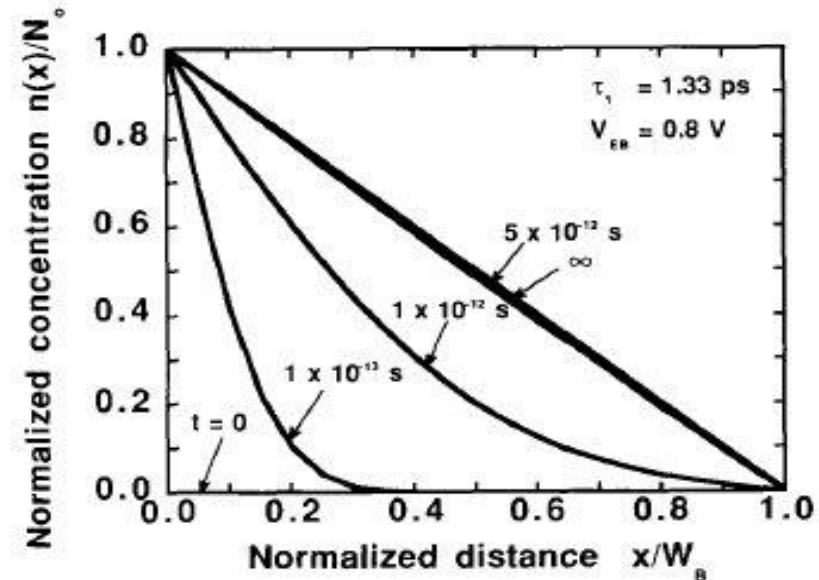
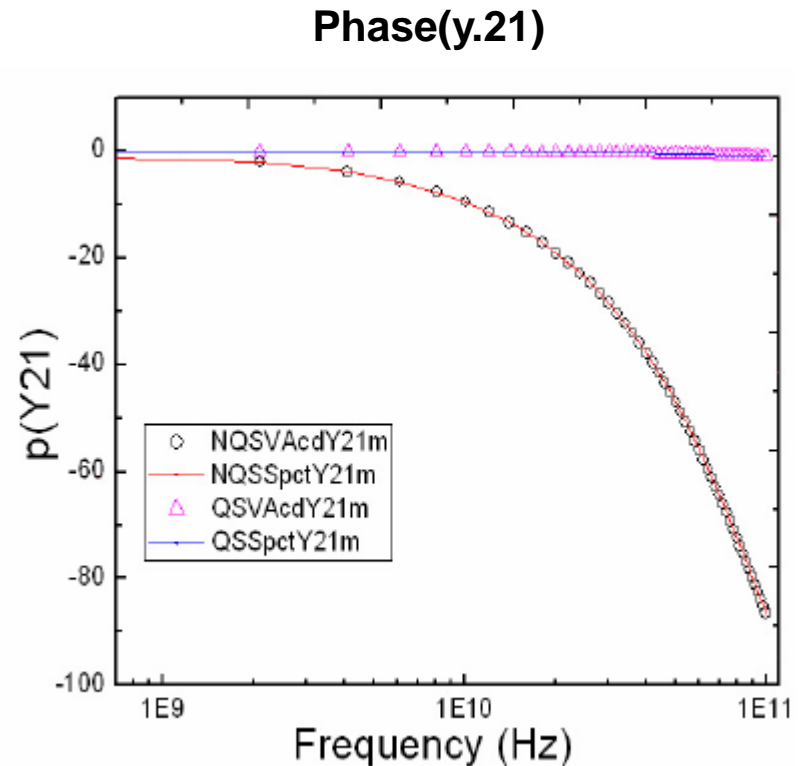
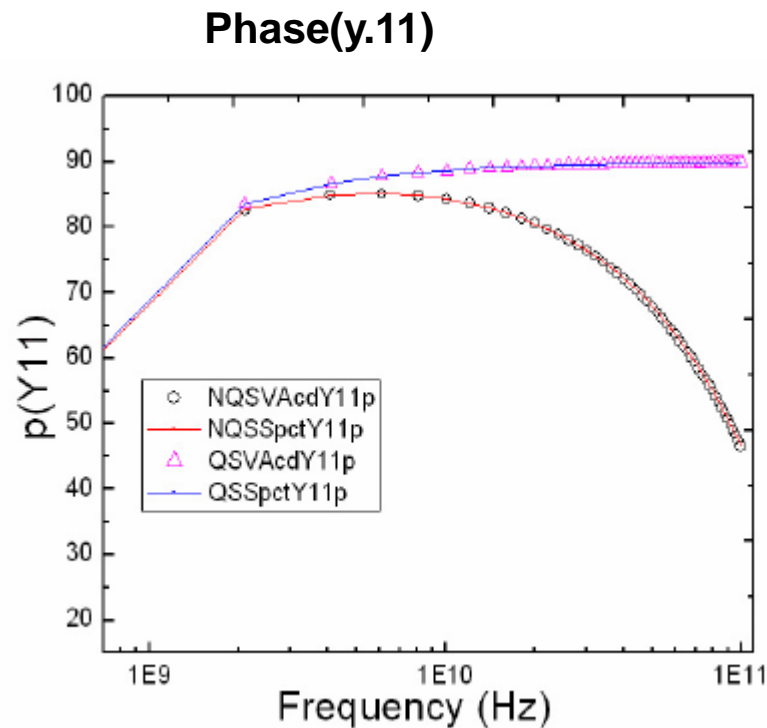


Fig. 8. Transient variation of electron concentration profile when V_{EB} increases from zero to the switch-on voltage.

From Suzuki IEEE TED 1992

Introduction: high frequency Y parameters

- Phase of admittance parameters show NQS effect.

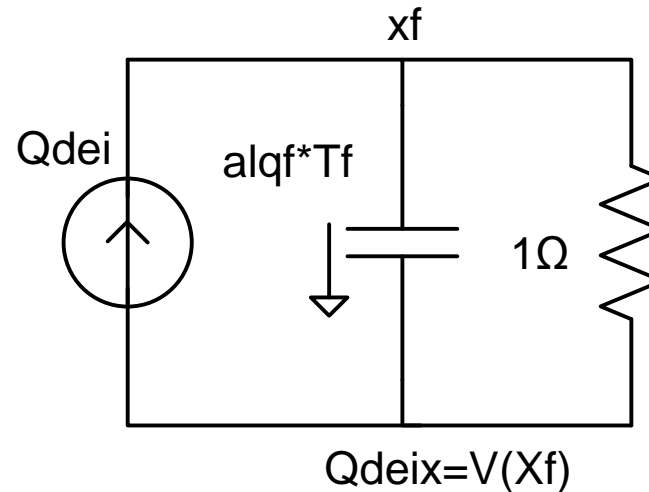


HICUM- Productization and Support Update, Oct 2007.

NQS model implementation : vertical NQS

- Phase shift network with excess charge.

HBT HICUML2V24

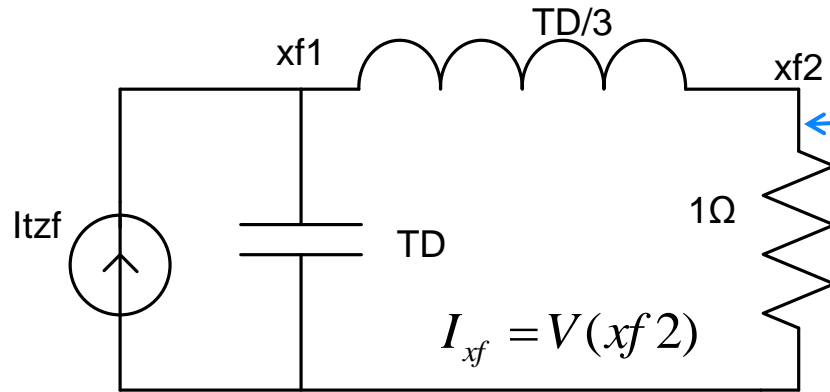


$$Q_{deix} = \frac{Q_{dei}}{1 + sT_D}$$

$$T_D = alqf \cdot T_f$$

NQS model implementation: vertical NQS

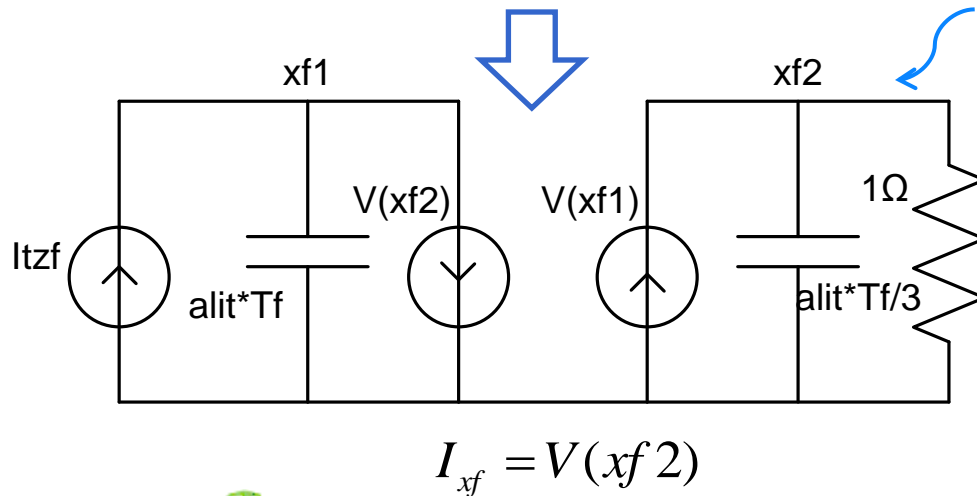
Phase shift network with transfer current (HICUML2).



Weil-McNamee formulation

$$I_{xf} = \frac{I_{tzf}}{1 + sT_D + (sT'_D)^2}$$

Gyrator equivalent formulation (L. Lemaitre)



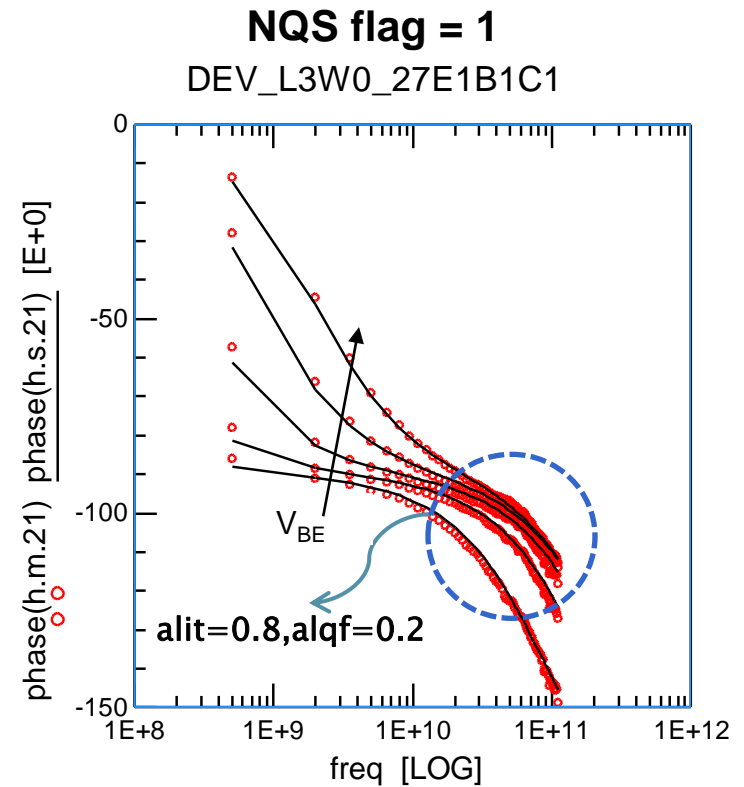
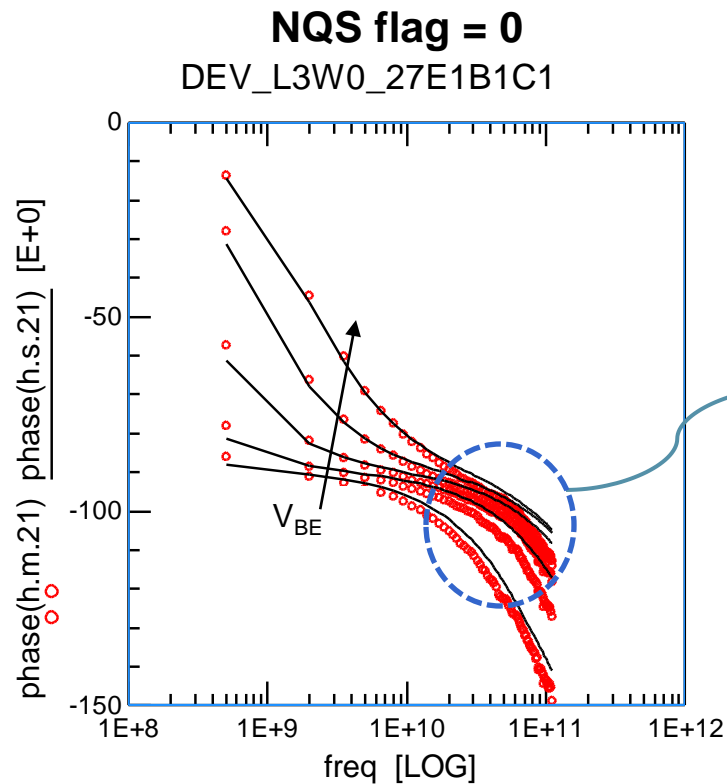
$$I_{tzf} = \frac{d}{dt} (alit * T_f * V(xf1)) + V(xf2)$$

$$V(xf1) = \frac{d}{dt} (alit * T_f / 3 * V(xf2)) + V(xf2)$$

$$I_{xf} = \frac{I_{tzf}}{1 + sT_D + (sT'_D)^2}$$

Modeling results: NQS modeling with HICUM L2

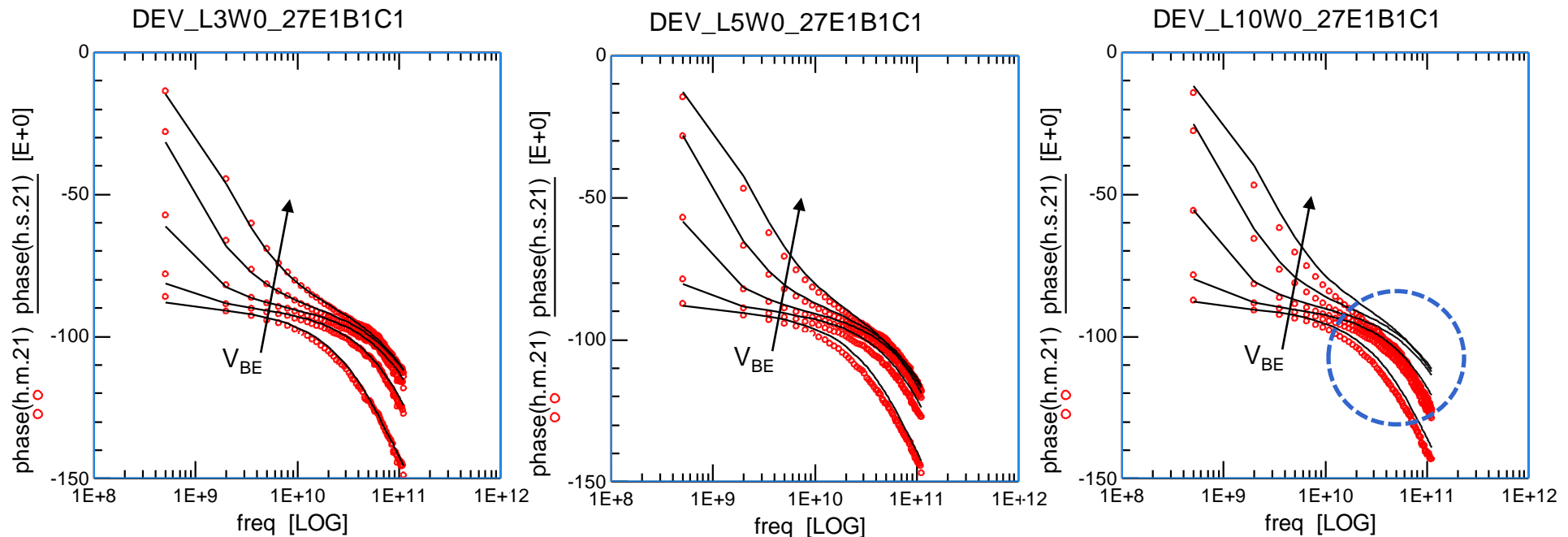
- Compact modeling of transistor having B3T technology and BiCMOS9MW layout. (f_T peak =240GHz)
- h_{21} : phase (current gain) $V_{BE}=0.76V \rightarrow 0.92V$, $V_{CE}=1.2V$



NQS effect in h_{21} phase and modeled using HICUM L2V24 model in a transistor of $L_E = 3\mu m$ and $W_E = 0.27\mu m$.

Modeling results: NQS modeling with HICUM L2

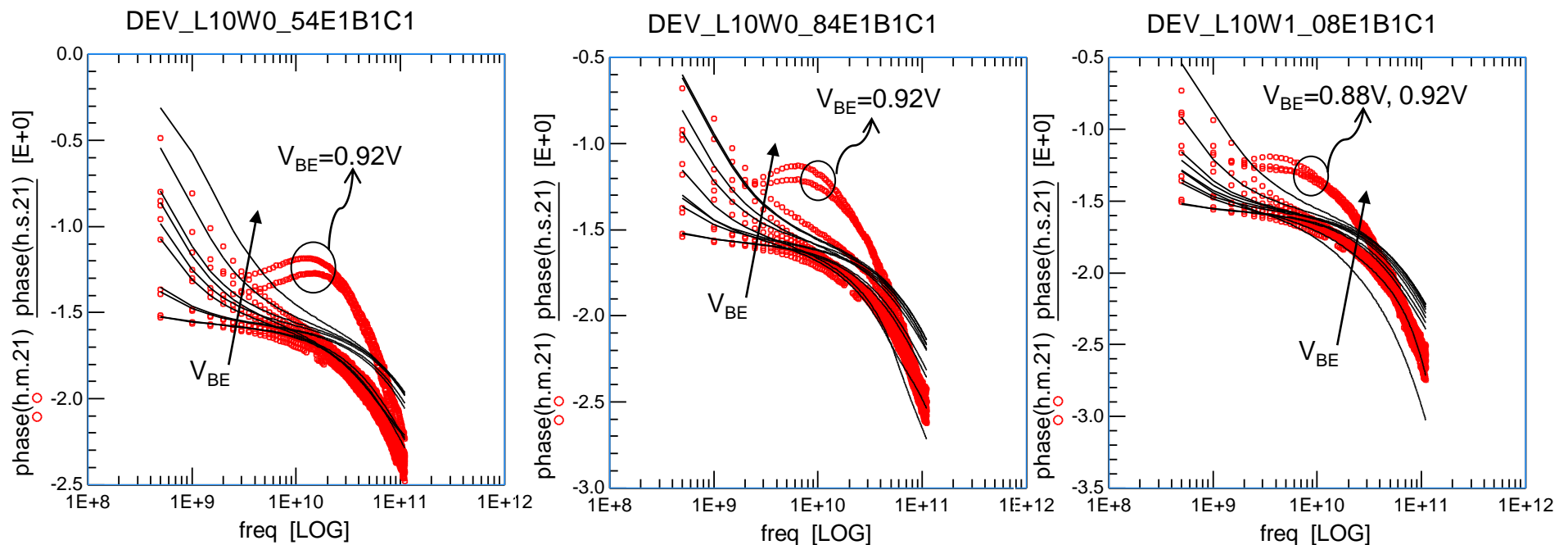
□ Parameter optimization in different devices $V_{BE}=0.76V \rightarrow 0.92V$, $V_{CE}=1.2V$



NQS effect in h_{21} phase and modeled using HICUM L2.24 model for transistors having emitter Lengths (L_E) = 3 μm , 5 μm , 10 μm and emitter width (W_E) = 0.27 μm .

Modeling results: NQS modeling with HICUM L2

- Large transistors show abnormality in h_{21} phase at high injection level

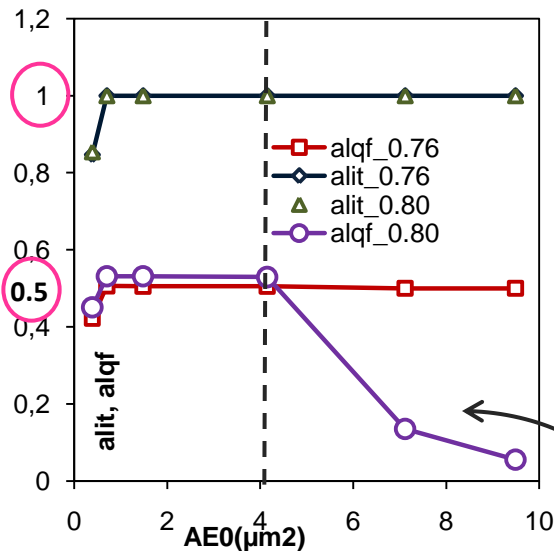


h_{21} phase for different W_E (0.54, 0.84, 1.08 μm) with same L_E (10 μm) at $V_{BE}=0.76V \rightarrow$ 0.92V, $V_{CE}= 0.8V, 1.2V$

Modeling results: : Scaling of NQS parameters

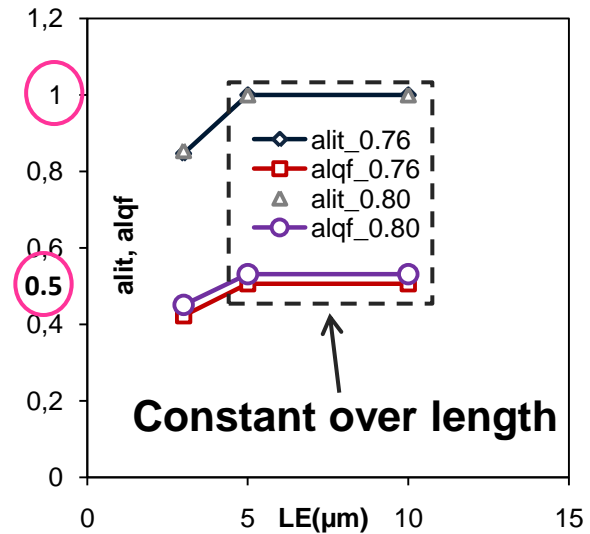
- NQS parameters are optimized for low injection level and plotted against emitter length (L_E) and real emitter area (considering the spacer width of 55 nm).

NQS parameters at $V_{BE}=0.76V$ and $0.80V$ and $V_{CE}=1.2V$



alit= 1.0
alqf= 0.5

alqf decreases at higher emitter area



Constant over length

Conclusion

- NQS basic is presented is presented while showing the HICUM implementation.
- HBT Modeling results with HICUM L2V24 model are presented.
- HICUM provides good results but at high frequency and long devices there is scope for modification (essderc paper 2011, submitted).
- Transistors with higher emitter widths show abnormality in h_{21} phase at high injection.
- Scaling of two NQS parameters show that constant value of NQS parameters ($al_{it}=1$, $al_{qf}=0.5$) provide a good modeling accuracy.

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