

Remarks on RF and Breakdown Modeling of Integrated Diodes

AKB 2023, Corbeil-Essonnes, France

Kund Molnár 10.11.2023

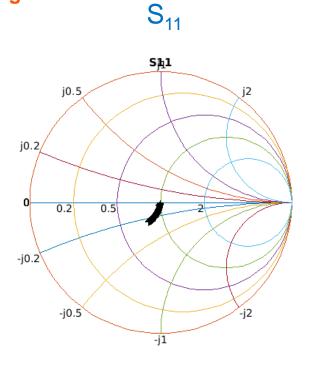
Agenda

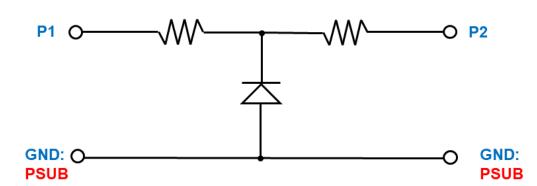
- 1. Level-1 RF Extractions with 1 Internal Node.
- 2. Level-1 RF Extractions with 2 Internal Nodes.
- 3. Diode_CMC Shortcoming in RF Modeling.
- 4. Soft Breakdown Modeling Options.
- 5. Conclusions



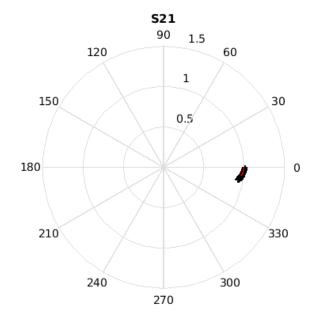
Diode RF Extractions: S-Parameter Setup

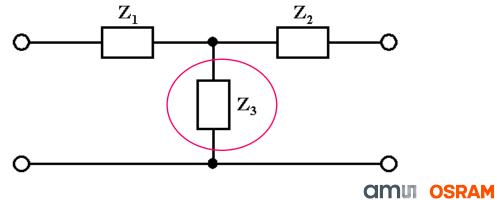
Diode RF Modeling





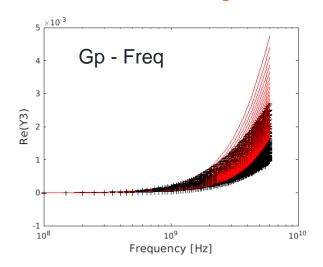


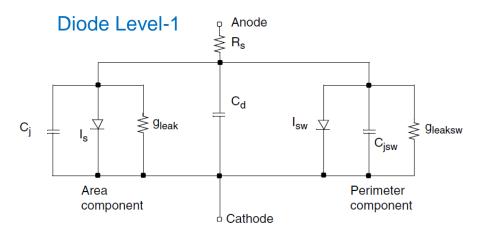


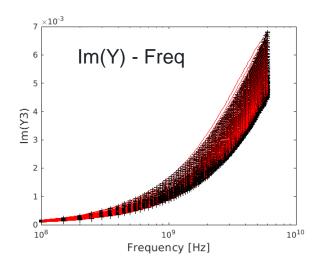


Diode RF Extractions: Cp-Gp from Admittance

Diode RF Modeling

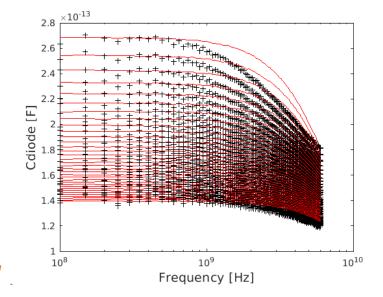




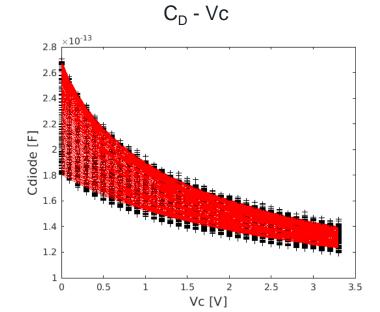


Adjusting Rs seems to be a good solution...

 $C_D = \frac{Im(Y)}{Im(Y)}$



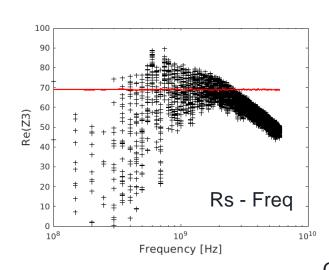
C_D - Freq

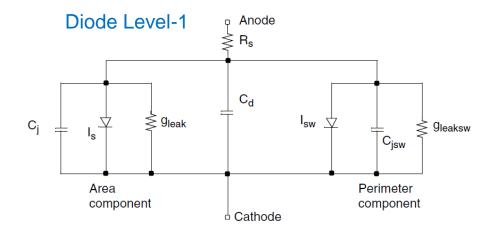


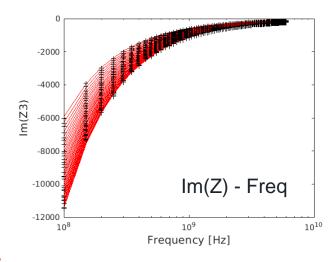


Diode RF Extractions: Cs-Rs from Impedance

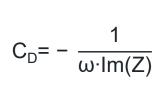
Diode RF Modeling

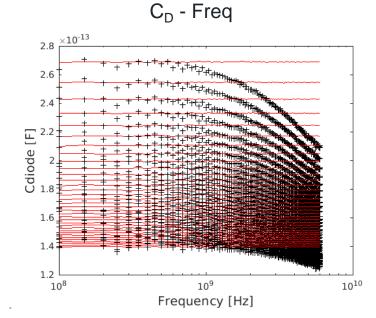


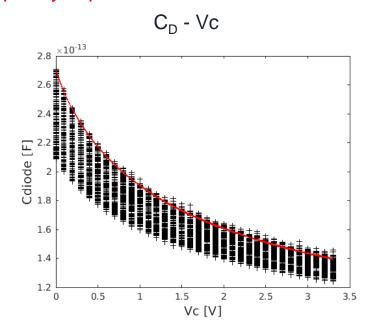




Not sufficient to reproduce the frequency dependence.



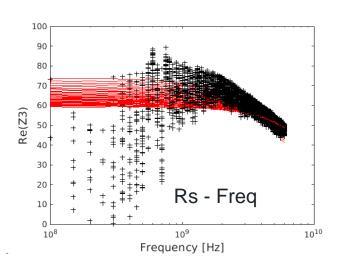


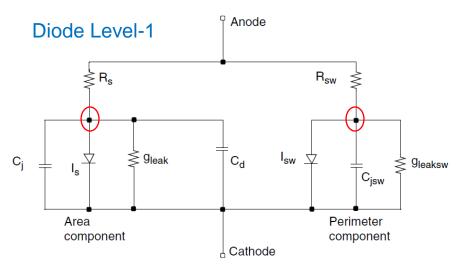


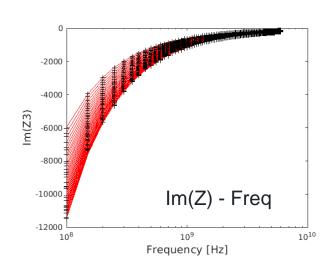


Diode RF Extractions: Cs-Rs from Impedance with 2 Internal Nodes

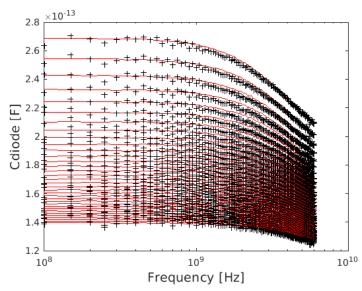
Diode RF Modeling



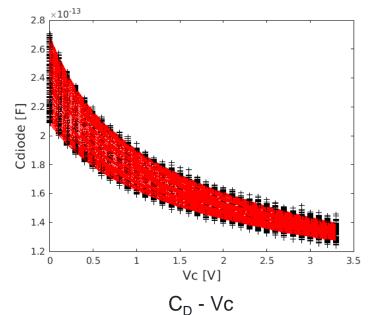




2 internal nodes are needed to produce this behavior!



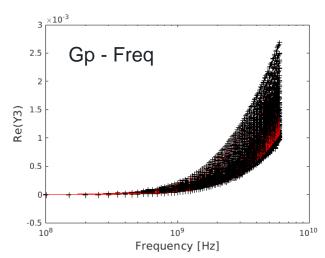
C_D - Freq

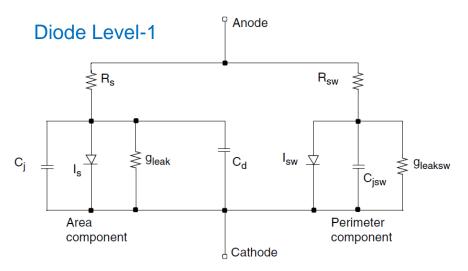


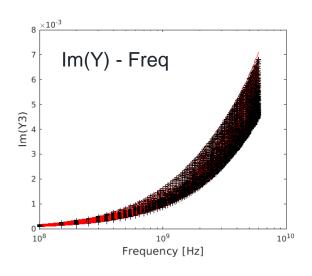


Diode RF Extractions: Cp-Gp from Admittance with 2 Internal Nodes

Diode RF Modeling





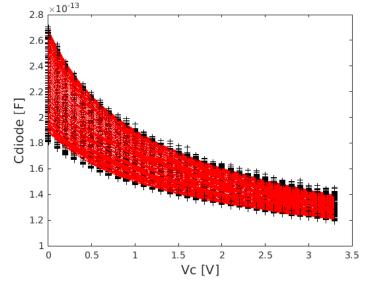


Can be produced by simple Rs – Cs equivalent circuit too.

$$G_{p} = \frac{\omega^{2}C_{s}^{2}R_{s}}{1 + (\omega C_{s}R_{s})^{2}}$$

$$U_{p} = \frac{2.8 \times 10^{-13}}{1 + (\omega C_{s}R_{s})^{2}}$$

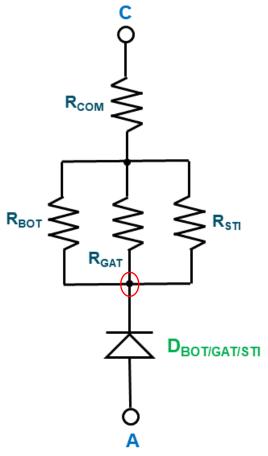
$$U_{p} = \frac{2$$



C_D - Vc

Diode_CMC Shortcoming

Diode RF Modeling



Resistance: one component with 4 possible scaling rules:

Area (RSBOT), Gate-edge (RSGAT), Perimeter (RSSTI) and fixed (RCOM)

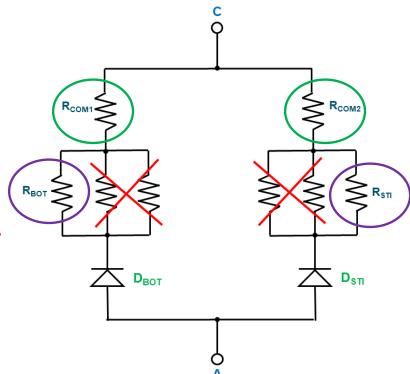
$$RS_{nom} = \frac{1}{\frac{AB}{RSBOT} + \frac{LG}{RSGAT} + \frac{LS}{RSSTI}} + RCOM$$

Capacitance: fully compatible with Level-1

- Single branch implementation with one internal node.
- Not suggestiv of RF modeling.
- > Subcircuit of 2 Diode_CMC diodes needed.

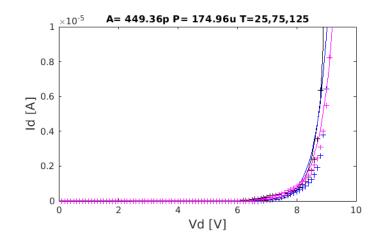


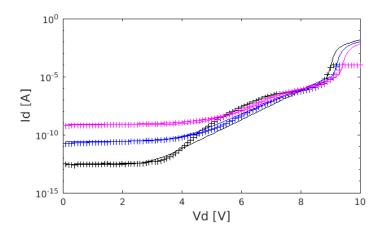
- 1. Setting R_{COM1} & R_{COM2} (unscaled), $R_{sbot}=R_{ssti}=R_{sgat}=0$
- 2. Setting R_{SBOT1} & AB_1 and R_{SSTI2} & LS_2 , $R_{COM1} = R_{COM2} = 0$, $R_{SBOT2} = R_{SSTI1} = R_{SGAT} = 1e9$



Diode Soft Breakdown I

Diode Breakdown Modeling





Level1:

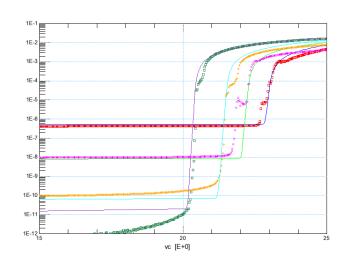
- > Trap-assisted tunneling current used for modeling the soft breakdown region:
 - JTUN, JTUNSW (tunneling saturation currents)
 - NTUN (emission coeff., set to a large value e.g. ntun>10)
- ➤ Hard breakdown: BV, NZ=1
- Area/Perimeter components (JS, JSW, CJ0, CJSW etc) as usual.

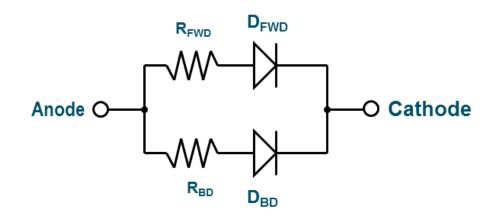
Diode Soft Breakdown II

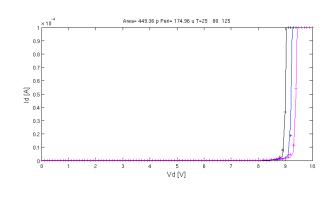
Diode Breakdown Modeling

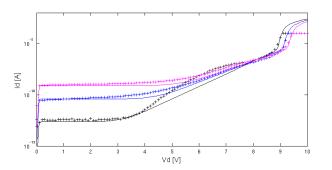
Diode Subcircuit:

- ➤ Composed of two Level1 diodes in parallel with series resistance.
- ▶ D_{FWD}: Forward and reverse DC including soft breakdown.
- ➤ D_{FWD}: Soft breakdown modeling using IBV, BV and NZ (slope).
- ➤ D_{BD}: Hard breakdown only, JS=JSW=CJ0=CJSW=0.
- D_{BD}: Resistance after BD can be adjusted with R_{BD} (fwd bias is not affected).









T=25, 80, 125 C



Conclusions

Diode RF & Breakdown Modeling

- Level-1: Cs-Rs extractions from impedance indicates best whether both internal nodes need to be activated.
- > Diode_CMC: Should be extended to 2 internal nodes to make it more suitable for RF modeling.
- Two different approaches have been discussed with Level-1 for modeling complex BD behavior:
 - Utilizing the trap-assisted tunneling current.
 - Sharing the BD effect between two diodes. No advanced effects are needed. Resistance after BD.



Thank you